

AM Broadcast Antenna Engineering

Problem Report

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LIST OF FIGURES

Figure 1. Ground Conductivity Map	23
Figure A1. Sample Input Deck	36
Figure A2. Sample Output	37
Figure B1. Sample Input Deck	50
Figure B2. Sample Output	51
Figure D1. Results for WGR and WFRB	89
Figure D2. Results for WFIL	90
Figure D3. Results for WCKL	91
Figure D4. Results for WHLM	92
Figure D5. Results for WSYR	93
Figure D6. Attempted Service Area	94
Figure D7. Protected Contour Points	95
Figure D8. Maximum Pattern Envelope	96

CONTENTS

LIST OF FIGURES	ii
CHAPTER I. INTRODUCTION	
A. BACKGROUND	1
B. OBJECTIVES	7
C. APPROACH	8
CHAPTER II. OVERVIEW OF DESIGN PROCEDURE	11
CHAPTER III. INTERFERENCE ANALYSIS TECHNIQUES	
A. DETERMINATION OF PROTECTED SERVICE AREAS	14
B. SITE FEASIBILITY AND GUIDES FOR PATTERN SHAPE AND SIZE	22
CHAPTER IV. EXAMPLE PROBLEMS	
A. FEASIBILITY STUDY FOR SMETHPORT, PENNSYLVANIA ON 560 KHZ	28
B. POTENTIAL SERVICE AREAS ON 560 KHZ IN PENNSYLVANIA	29
CHAPTER V. SUMMARY	
A. RESULTS	32
B. CONCLUSIONS	32
C. RECOMMENDATIONS	33
BIBLIOGRAPHY	34
APPENDIX A. STDPATRN SAMPLE INPUTS, OUTPUTS, AND PROGRAM LISTING	35
APPENDIX B. PCMLL2 SAMPLE INPUTS, OUTPUTS, AND PROGRAM LISTING	48
APPENDIX C. FCC GROUND WAVE CHARTS	67
APPENDIX D. EXAMPLE PROBLEM	88
APPROVAL OF EXAMINING COMMITTEE	97

CHAPTER I. INTRODUCTION

A. BACKGROUND

Since the beginning days of radio broadcasting, the problem of providing interference-free reception has been of major concern. In 1921, the broadcast band consisted of two frequencies (750 kHz and 833 kHz). Anyone wealthy enough to afford the equipment and operating costs could broadcast on one of those two channels. At the outset, little thought was given to the problem of interference between competing broadcasters. As the number of broadcasters grew, however, interference became a severe problem. By 1923, the problem was completely out of hand, and a solution was urgently needed. Thus, in that year, the Secretary of Commerce decided to assign separate channels to each station to help alleviate the interference problem. In 1924, the broadcast band from 550 kHz to 1500 kHz with channel spacing of 10 kHz was established. The establishment of this broadcast band, and the desire to provide interference free broadcasts to consumers, forced upon us the problems of frequency allocation and interference protection that still remain today.

Although in the 1920's the methods of reducing interference were not as sophisticated as today, the concept of separating stations using the same frequency by a considerable distance or by limiting the hours of operation (time sharing) are still evident. The problem of providing interference protection to a neighboring station, and still having a sufficiently large coverage area was one of the most difficult problems faced by early radio engineers. This problem was compounded by the fact that the antenna systems of the day were essentially omnidirectional. Thus, any attempt to reduce

radiation for the sake of interference in one direction would automatically reduce it in all directions, hence, reducing the coverage area and the market audience available to the station. This dilemma was solved in 1932 by the construction of the first directional antenna for broadcast use¹. A simple two tower array was used successfully by WFLA-WSUN in Clearwater, Florida to prevent interference to WTMJ in Milwaukee, Wisconsin. Thereafter, directional array (DA) antennas have been used extensively to limit interference and efficiently cover desired market areas.

To design antennas to meet specifications of interference level and coverage area, knowledge of how to compute the radiation from an antenna, and how this radiation is attenuated as it travels over the surface of the earth are of vital importance. For the engineers of fifty years ago, methods were available for predicting the fields radiated by single towers and arrays of towers. However, the solution to the attenuation problem was computationally intractable, and even the theoretical solutions were riddled with errors.

The problem of how the electromagnetic field is attenuated over a lossy surface was first addressed in a theoretical fashion in 1909 by Arnold Sommerfeld. Sommerfeld's original solution contained an error and was the cause of considerable confusion. It took until 1930 before the mistake was even recognized, and until 1936 that an acceptable solution was found. Here, the work of K. A. Norton was used for a solution that was valid for short distances. Later, in 1940, the Federal Communication Commission (FCC) Groundwave Charts were published. These charts utilized Norton's solution for short distances, and the work of Blath van der Pol and H. Bremmer for larger distances. The solution for middle distances was computationally intractable at

the time, so the FCC fit the curve between the largest distance where the Norton solution was valid and the shortest distance where the other solution was considered valid. In addition to the obvious error incurred by this process, there were also systematic errors in the drafting process, whereby the values at the larger distances were shifted upward. These curves have been updated several times since 1940, but these problems were not corrected. In 1979, the FCC attempted to recompute the curves using a computer program. Again, the middle distance area was not in full agreement with those of the International Radio Consultative Committee (CCIR) and other theoretical results. Finally, in 1986, the FCC published a computer program that made available a reliable and accurate means of computing the attenuation of the groundwave field strength over a lossy spherical earth².

Despite the availability of computers and computerized solutions to many of the problems of antenna radiation and radio wave propagation, there are still many problems to be surmounted by the broadcast engineers of the 1980's. A brief mention of some of these may be useful to set the stage for the objective of this report.

One subject that is always of great interest is obtaining a larger coverage area for a smaller amount of input power. This problem is usually approached by designing antennas whose directional properties are such that the largest portion of their radiation is along the ground. Historically, this was done by building taller and taller towers. This, however, led to other difficult problems, not the least of which was the cost of a the tower, and the fact that it is sometimes impossible to build an exceedingly tall tower in some locales due to both local zoning laws, and Federal Aviation Administration (FAA) rules.

Other topics that have been of interest are the bandwidth of the antenna, and the antenna feed elements^{3,4,5,6,7,8,9}. The bandwidth is of some concern for stations broadcasting monaural program material and using DAs, since cases arise where the carrier frequency is greatly attenuated with respect to the sideband information. This will result in over-modulation distortion in receivers utilizing diode detectors. The effects of this type of interference on AM stereo broadcasts is a troublesome problem that is not yet fully resolved. Thus, suffice it to say, that a broadband antenna system is usually a goal of the antenna designer.

Finally, one topic that is of interest to virtually all broadcasters is the effect of reradiation of energy from parasitic elements (eg. tall buildings, electric utility poles, and other broadcast towers) ^{10,11}. The constant modernization and urbanization of once rural areas has forced this problem on broadcasters who purposely erected broadcast towers far from city structures. Fortunately, over the years, there have been numerous techniques developed to detune these parasitic structures, and thus minimize the reradiation of energy in undesired directions, although this still remains a problem.

In addition to the countless technical problems to be surmounted by the modern day consulting engineer, there are also other problems with which to be reckoned. One of the largest problems is obtaining up-to-the-minute information concerning the licensing status of all the stations that may have to be considered in an interference study. This can be overcome if the designer has the funds to obtain a data base of information covering the entire broadcast band and the time to keep the information organized. Otherwise, one is forced to make numerous visits to the public information reading room at the FCC headquarters in Washington, D.C. The engineer is also

responsible for being up to date on all of the latest FCC rulings and recommendations. New rulings are constantly appearing and old ones are being updated. The new rulings and updates are available on a subscription service, so that being current is only a matter of reading the material pertinent to the particular area of consulting with which one is involved.

Finally, even with the data bases and recent rules in hand, there are very few references available that outline the technical procedures required to design broadcast antennas in accordance with FCC rules and regulations. This, at first glance, may seem to be a minor problem in comparison with some of those mentioned previously; however, for the uninitiated, this problem can seem immense. Upon inspection, it does not take long to realize that the "legalese" contained in the FCC rules and regulations ¹² governing the broadcast services is somewhat confusing. In addition, even though some of the technical matters seem to be completely specified, others are not, leaving some question as to the correct procedures. For the broadcaster who wants to have a radio station, the solution to getting a suitable antenna design is to contract a consulting engineer to do the job. It seems however, that perhaps due to intense competition, there is a reluctance for the knowledgeable engineers to publicly reveal all of their methods in designing antennas. Short of trying to gain employment with these experienced engineers, until now there has been no way to learn these engineering techniques. This shortage of information on actual specification techniques for AM broadcast antenna patterns will be the focus of this report.

Before jumping in over one's head, it is useful to discuss the extent and nature of this documentation shortage. At one time, prior to the availability of the latest FCC computer program (See reference 2), there was a desperate

need for a new method for predicting ground wave field strengths. Not only were the old graphs in error, but they were completely unmanageable for serious design endeavors in the modern computer age. Moreover, attempts to digitize the curves for use in automated design seemed ridiculous in light of their already apparent shortcomings, but was done none-the-less to speed the process. A well-documented computer program was needed that would eliminate the errors contained in the old graphs and put the FCC methods on firm theoretical ground. Prior to beginning work on this problem report, it was thought that a major reworking of the FCC Ground Wave Charts would be the eventual focus of the report. A few phone calls to the FCC offices in Washington, D.C. revealed, however, the FCC had already developed a new program and documented it in a report (See reference 2) published only a few months prior to the author's inquires. It should be noted that the new program is excellent. It is written in a very portable version of FORTRAN, and requires no specialized mathematical functions, other than those common functions built into all current FORTRAN languages. The accompanying report is also excellent. It includes a complete history of the problems associated with the methods the FCC has used in estimating ground wave field strength, along with all of the flow diagrams, and a glossary of the variables used in the program and their symbolic counterparts used in the original works of Sommerfeld and Norton. The program is included in the listing of program PCMLL2 in Appendix A.

The next part of the documentation problem was determining how to design the antenna system. The FCC has ruled that the usual assumptions of a sinusoidal current distribution on the tower be a common starting point for all computations. This has eliminated the use of the modern method-of-moments

techniques currently employed for most difficult antenna problems in this frequency range. It therefore seemed useless to rehash the standard array theory seen in every undergraduate text on antennas. Moreover, the 1949 text by Smith¹³ on the directional antenna system is specifically directed to the AM broadcast antenna engineer. This superb work is where the uninitiated must look for discussions of the types of antenna computations necessary for AM antenna engineering.

Although antenna and groundwave propagation theories are well documented, explanations of how to use these theories as tools to perform the interference studies and antenna synthesis are not as well documented. These explanations are essential for those engineers attempting to file construction permits with the FCC for AM broadcast stations.

B. OBJECTIVES

The objectives of this project were :

1. Provide documentation on the use of wave propagation and antenna theories as tools in performing an interference analysis for AM broadcast stations operating during daylight hours. The discussion is limited to operation during daylight hours since the propagation theory for ionospheric radio paths is not as well understood or as computationally tractable as the groundwave path, nor is it necessary for this specific project.
2. Illustrate the use of these wave propagation and antenna theories as tools in an actual design problem.

C. APPROACH

In an attempt to meet the objectives, several intermediate steps were performed. First, a literature search of pertinent subject material was initiated. This provided the background necessary to undertake the ensuing design and documentation project. As well as searching the literature for the theoretical aspects of the problem, several persons knowledgeable in the area were contacted either by phone, or in person. These persons include the student's advisor, Dr. James F. Corum, an experienced radio consulting engineer; Mr. Wayne Fried, an applications reviewer for the FCC; Mr. Victor Tawil, an engineer with the FCC Office of Engineering Technology; and Mrs. Virginia Cannon, manager of the Downtown Copy Center (DCC), Washington, D.C. (DCC provides support service to consulting engineers for data base and rule changes). With pertinent literature and first hand information, some of the techniques for designing the antennas became clear. The next step was the development of the computer software necessary to rapidly implement the antenna and wave propagation theories. The computerization of these design tools is necessary due to the iterative nature of the interference study.

Simultaneously, with the development of the software, the acquisition of a data base of antenna information for stations on several frequencies was initiated. This data base is necessary to perform an interference study pursuant to the FCC Rules and Regulations, Part 73. As discussed with Dr. Corum, there were initially five different frequencies and corresponding locations suggested by employees of station WPLW in Pittsburg, Pennsylvania, for use as example problems and as potential profit making ventures. The WPLW engineers were interested in the feasibility of locating a station at these

sites and operating on their specified frequencies. (For confidentiality, these locations and frequencies, except for the example problem contained in this report, will not be disclosed.) Unfortunately, due to lack of funds, only a data base of stations on one of the suggested frequencies was obtained.

Throughout the duration of the project, the computer programs were constantly modified, tested against known examples, and debugged until accurate and reliable results were obtained for the purpose of proceeding with an actual design problem. The programs used to analyze antennas had to be written from scratch, even though excellent documentation existed (See reference 13). The program for prediction of field strengths, though available from the FCC, still had to be keyed in and modified slightly to work on the West Virginia University IBM mainframe computer. This program was subsequently enhanced to provide other features necessary for the interference analysis.

After some unsuccessful attempts at trying to reverse the FCC's equivalent distance method, so as to simplify the antenna synthesis part of the example problem, the real world example problems were attempted. Unfortunately, it was not possible to locate an antenna in the area suggested by the clients in Pittsburgh, due to interference from an adjacent channel station in Buffalo, New York. Thus, the synthesis problem was not fully illustrated. Subsequently, in an effort to demonstrate the actual synthesis of an antenna system, a possible service area was determined on the same frequency some 50 miles southeast of the original proposed site. At this point, it may have been possible to proceed with the synthesis had it not been for lack of time and resources.

Finally, a report describing the methods and tools required to perform an

interference study, and synthesis of an antenna system for use in the AM broadcast band was written.

CHAPTER II. OVERVIEW OF DESIGN PROCEEDURE

Before embarking on a discussion of numerous theoretical and practical details, it seems appropriate to first provide an overview of the major steps involved in designing an AM broadcast antenna.

The primary step in the design of an antenna system is to determine a suitable service area and frequency of operation. Sometimes, as this report will illustrate, the location and frequency are given by the broadcaster, who, after making a market study, has decided that it might be profitable to serve some particular area. More often, it will probably be the engineer who is responsible, at least in part, for narrowing the choices of which areas can be serviced, and on which frequencies this service can be established. Thus, it is necessary that the engineer be able to determine the protected service area (the area within the protected field strength contour) of existing stations on the frequency of interest (cochannel), as well as those on adjacent channels (stations on the next higher or lower frequencies).

Once the service areas of the existing stations are established, it is possible to determine which areas, if any, can be serviced by the new station. Many times, it is impossible to provide service to an area on a particular frequency, and several frequencies must be examined for interference. With luck, a desirable service area and suitable frequency can be identified that will allow a new station to operate without causing objectionable interference to either cochannel or adjacent channel stations.

Once the service area and frequency have been established, the location of the antenna system must be determined. This is a very difficult task which demands a great deal of attention. The search for the best location must be

guided by experience and practice. A very limited list of factors to consider when deciding where to locate the antenna system would certainly include: the principle and secondary service areas, and their relationship to the protected service areas of other stations, the blanket area of the station, the conductivity of the ground at the site, the accessibility of the site, and whether the site is obtainable for the purpose of installing a broadcast antenna according to local zoning laws and FAA regulations. The location of the antenna is a very important and difficult assignment.

Once the engineer and client reach an agreement as to the location of the transmitter site, it is the engineer's responsibility to determine what type of antenna and what power level is required to provide adequate service to the desired areas, yet not produce interference in the protected service areas of other stations. The objective is to find the least expensive means of satisfying the above criteria on service area and interference. Sometimes, the designer will be fortunate in that a single tower will provide adequate coverage and not produce interference. More often than not the design will require at least two, usually three, and sometimes 6, 8, or as many as 12 towers arrayed together. These directional array antennas are used to produce a service area of the required shape or to limit interference to one or more stations in various directions from the transmitter site. Usually, the most economical design is one with the fewest towers of the smallest height.

As one proceeds with the design of the antenna to meet the above criteria, one quickly realizes the value of a computer. The process is essentially one of trial and error guided by experience. There are some guides to learning how the various parameters of the array affect the pattern for simpler arrays, but for an 8 tower array, the designer is probably on his

own. One reference that is essential for beginning estimates of the array parameters is the work by Carl Smith¹⁴, entitled Directional Antenna Patterns. This book contains, in a systematic way, over 15,000 array patterns with their associated parameters for two and three tower arrays. The parameters include the orientation, spacing, height, sectionalization, loading, and the magnitude and phase of the current on the towers. Even with this reference, the antenna design is a very tedious and exhausting process, where experience and fast computers pay big dividends in time.

Once the antenna design is completed, the engineer proceeds to prepare a construction permit and submit it to the FCC. The construction permit indicates all of the essential information about the proposed location, frequency, antenna parameters and pattern, service areas, etc., as well as a supporting analysis showing that the proposed service area does not produce interference to other stations.

CHAPTER III. INTERFERENCE ANALYSIS TECHNIQUES

With the overview of the process in mind, some of the particulars will be addressed. The discussion will begin with details about how to determine the protected service area of a broadcast station using the FCC ground wave field strength prediction program (hereafter referred to as FCCGW). Then, an illustration of determining the feasibility of placing an antenna in a given location will be presented. At this point, assuming there exists a feasible location, a technique for estimating the maximum radiation in the azimuthal plane of the new station at the proposed location will be discussed. This maximum envelope will enable the techniques and patterns presented in the two works by Smith (See references 13 and 14) to be used to their fullest extent in the synthesis of the antenna.

A. DETERMINATION OF PROTECTED SERVICE AREAS

The protected service area of a station depends upon the class of the station and the class of the potential interfering station as set forth in the FCC Rules Part 73. Further, it is also dependent on the power radiated, the directional properties of the antenna, the frequency of operation, and the electrical properties of the ground surrounding the antenna.

As far as the influences of the class of the station and the interfering station are concerned, nothing difficult is involved. The Code of Federal Regulations (CFR) Title 47, Part 73.182 specifies what field strength contour is considered protected for the various classes of stations. The maximum allowable interfering signal permitted within the protected service area of the

station is also defined.

The directional characteristics of the antenna are also important in establishing the area within the protected contour. There are two general types of antennas: omnidirectional and directional. For omnidirectional antennas (i.e., single vertical towers), the field strength is equal in all directions in the azimuthal plane. For a directional antenna, the field strength varies as a function of the azimuth angle. The radiation also varies as a function of elevation angle for both types of antennas, but this variation is not usually important in the daytime interference study, where the ground wave is the principle mode of propagation. For both cases, information must be obtained on the unattenuated radiated field strength at one kilometer (i.e., the antenna pattern).

In the case of the omnidirectional antenna, the pattern is always presumed circular for the case of interference analysis, even if it might not be perfectly so in the real world. On file with the FCC is the expected value of field strength at one kilometer per kilowatt of input power. This value, along with the licensed power of the station, are necessary for establishing the protected service area for an omnidirectional antenna.

In the case of the directional antenna, there are several different types of patterns encountered in interference studies performed in conjunction with the FCC. The first pattern, the theoretical pattern, is described in detail in both the FCC rules Part 73.150, and in Smith's book. Essentially, this is the pattern given by superposition of the field radiated by each of the towers in the array. The field includes the effects of FCC specified losses in the tower and assumes a sinusoidal current distribution on all towers. The theoretical pattern is never used in computing interference. It is only used as a

reference point for establishing what is referred to as the standard pattern. This pattern is essentially the same as the theoretical pattern with the quadrature addition of some terms which the FCC requires to make the computed pattern more closely approximate the field strength the actual antennas tend to produce in the area of pattern nulls. The addition of these terms also provides some padding so that the field strength predicted should be larger in all directions than that actually produced by the antenna. This helps ensure that even for "tight fit" computed contours there will be not objectionable interference when the actual antenna is installed. Finally, there is the case of the augmented pattern. A station must file a pattern of this type when the actual measured radiation is significantly different from the computed standard pattern filed with the construction permit. This is only acceptable if the deviations from the computed standard pattern are not significant enough to cause objectionable interference to another station's service area. The methods for establishing the augmented pattern from the standard pattern, and actual measurements are set forth in FCC Rules Part 73.151 and 73.152. The program STDPATRN, used for computation of standard patterns is listed in Appendix B.

The distance from the transmitter to the protected contour and hence the protected service area of a station is dependent on the frequency of operation and the ground parameters, both at the location of the antenna and some distance from the antenna. The distance to the protected contour increases as the frequency decreases for a given field strength and set of ground parameters. It might be noted that only antennas which produce vertical polarization are used in the AM broadcast band due to the very high attenuation of the horizontal field component in this frequency range. As such,

the computer program FCCGW only considers vertically polarized fields.

At this point, it is useful to discuss the capabilities of the program FCCGW, and the modifications necessary to predict the distance from the transmitter site to the protected contour. First, the capabilities of FCCGW will be discussed. This program requires as inputs the following information:

- The unattenuated field strength in mV/m at a distance of 1 kilometer from the antenna. This information is tabulated and on file for all licensed stations at the FCC.
- The ground conductivity in mS/m over which the signal is to travel.
- The ground permittivity relative to free space over which the signal is to travel. Unless proven differently with measurements by the consultant, this value is always assumed to be 15 for the purposes of interference computations.
- The frequency in MHz.
- The distance over which the signal is to travel.

With these parameters, the FCCGW program can very accurately compute the field strength at a given distance from the transmitter. This is very useful, but not quite what is required to perform an interference analysis. What is desired, is the distance from the transmitter to some specific value of field strength (e.g., the 500 μ V/m contour). Thus, additions to the FCCGW program are required to solve this problem.

The simplest means of finding the distance to a given contour is to use an iterative Newton Method of guessing the distance and computing the field strength. When the computed field strength is very close to the desired contour level, the distance used in the computation is the desired result. The subroutine CONTUR (Listed as part of PCMLL2 in Appendix A) was developed by the author for this purpose, and the program FCCGW was used as a subroutine to evaluate the field strengths.

It should be pointed out that there are several difficulties encountered with the iterative technique applied to the field strength as a function of distance relationship computed by FCCGW. The first of these is the fact that Newton's method will tend in some cases to compute as negative numbers as new guesses for the distance. This is a problem, since all distances entered into the FCCGW subroutine must be positive to avoid fatal program errors. The difficulties can be circumvented by simply checking for to see if the distance is negative. If it is, then simply divide the old guess by two. This is appropriate since a negative distance is indicating that the next guess must be smaller. Another problem arises due to the fact that FCCGW uses two different solution techniques, dependent on the distance from the transmitter and the frequency. If the distance is less than $(80.0 / \text{frequency})^{1/3}$, then the program implements Sommerfeld's solution for flat earth with some curved earth correction terms. For greater distances, a more accurate residue sum solution is used. The two solutions, do not yield exactly the same results in the limit as the distance approaches $(80.0 / \text{frequency})^{1/3}$. This produces a discontinuity in the field strength function, which in turn will cause the program to loop indefinitely between two different guesses for the distance, one guess on one side of the discontinuity and one guess on the other. This condition must be detected, and the more accurate residue sum solution used until the method converges to a value within the desired field strength tolerance. Both of these problems are already accounted for in the program listed in Appendix B.

The method discussed above can not account for the case where the ground conductivity changes as one moves away from the transmitter site. The effects due to changes in the conductivity must be included in the analysis

for accurate results. The FCC suggests that reasonably accurate results can be obtained by using the equivalent distance technique. This is described in Part 73.183 of the rules and illustrations of its use with the FCC Ground Wave Charts are provided.

For several years now, the FCC Ground Wave Charts have not been provided with copies of the Rules and Regulations in the CFR format, despite the fact that they are referred to in illustrations of interference computations and the equivalent distance technique! They are available, but at additional cost. Copies of older versions of these charts with distances in miles have been provided for convenience in Appendix C. These charts, being photocopies of the originals, are not sufficient for use in an engineering study, but are useful when trying to follow the examples contained in the FCC rules and regulations. They can also provide a guide to the reasonableness of results obtained using the FCCGW, CONTOUR, and PCMLL2 programs.

The equivalent distance technique can be used when the unattenuated field strength at one kilometer (or one mile if using charts in Appendix C), the ground conductivity, and location of discontinuities in the conductivity are known. The method presumes that the field strength is the same on each side of the discontinuity, but the equivalent distance to the transmitter changes abruptly as the wave propagates over the discontinuity. Thus, the location of the imaginary transmitter can be effectively closer or further away, depending upon the values of the conductivity.

For example, consider the case of a station with a field in a given direction of 100 mV/m at 1 mile , operating on a frequency of 610 kHz. The conductivity is 10 mS/m for the first 10 miles, then abruptly changes to a value of 5 mS/m for the next 10 miles, at this point it changes again to a

value of 15 mS/m. It is desired to find the distance to the 0.5 mV/m contour of the station.

If the distance to the first discontinuity is greater than the distance to the 0.5 mV/m contour using the first value of conductivity, the problem can be solved using a single value for the conductivity. Using Graph 3 of Appendix C, one can see that the distance to the 0.5 mV/m contour using a conductivity of 10 mS/m is approximately 69 miles, much greater than the 10 miles to the first discontinuity. Since the equivalent distance technique is necessary, the field strength must be computed at the first discontinuity using a conductivity of 10 mS/m. This field strength is 8.4 mV/m. Now, the equivalent distance to the 8.4 mV/m contour must be established using a sigma of 5 mS/m. This distance can be read from the curve as about 8.5 miles. The imaginary transmitter appears to be 1.5 miles from the actual transmitter. Thus, the imaginary transmitter is closer to subsequent discontinuities in the conductivity. These distances must then be adjusted by -1.5 miles.

Proceeding as before, one must ascertain whether the distance to the 0.5 mV/m contour is less than the distance to the next discontinuity. The distance to this discontinuity must be adjusted by the -1.5 miles found above. The distance to the 0.5 mV/m contour is about 47 miles, much greater than the effective distance to the next contour of $(20 - 1.5) = 18.5$ miles. The field strength at 18.5 miles must now be found using a sigma of 5 mS/m. This is about 2.9 mV/m. Again, the equivalent distance to this contour must be found using a sigma of 15 mS/m. This distance is approximately 29 miles, effectively 11.5 miles more distant than the previous imaginary transmitter. The distance to all other discontinuities must be adjusted by +11.5 miles. Since there are no further changes in the conductivity, the next step is to find the distance to

the 0.5 mV/m contour using a sigma of 15. This distance is 84 miles. Now, bear in mind that this is the distance from an imaginary transmitter that is actually $- (11.5 - 1.5) = - 10.0$ miles from the actual transmitter site. Thus, the distance to the 0.5 mV/m contour from the actual transmitter site is 74 miles.

To simplify the computation of the distance to a given field strength contour when the equivalent distance technique must be used, a computer program was written to perform the necessary computations. This program, has been named PCMLL2, and is listed in Appendix A. It uses the CONTUR and FCCGW subroutines to compute the distances and field strengths necessary in the equivalent distance technique.

With a computerized method for predicting the distance to a field strength contour it is possible to determine the protected service area of a station. The conductivity profile as a function of the radial distance from the transmitter site must be obtained on 35 different bearings (every 10 degrees beginning at true North). This information, along with the field obtained from the standard or augmented pattern, is sufficient to compute the distance to the protected contour. By plotting these protected contour points on a suitable map, the service area can be identified as the area within the envelope of these plotted points. For convenience, the program PCMLL2 will accept as input the location (latitude and longitude) of the transmitter site, and compute the latitude and longitude of each of the contour points. Plotting of the points using their latitude and longitude is sometimes necessary, since plotting by the use of the bearing and radial distance method can incur errors due to the inherent distortions of flat maps.

Determination of the ground conductivity is to be made with the use of FCC map M-3, which shows the conductivity of the soil projected on a map of

the United States. A copy of Map R-3, a less detailed replica of map M-3, is shown in Figure 1. Values of conductivity used in analyses submitted to the FCC should use values from M-3. There have been attempts to digitize Map M-3, (See reference 15). The FCC allows the use of programs that can supply soil conductivity information from M-3, and this would be the preferred method for a determining the conductivity profile along the radials.

B. SITE FEASIBILITY AND GUIDES FOR PATTERN SHAPE AND SIZE

With the capability to determine the protected contour of one station, it is just a matter of repetition of the process for many stations to evaluate whether a particular location is suitable for service on a specific frequency. To find out what stations are operating on a given frequency, one needs to obtain a copy of the listing of stations from the FCC in Washington D.C., alternatively, the list can be obtained by contacting either the designated FCC copying firm or the Downtown Copy Center. Using this list, and some common sense, it is possible to eliminate quickly 80 to 90 percent of the stations from consideration in an interference study. For example, if one desires to locate a station in Pennsylvania, there is no need consider a 1 kW Class IV station in California. It will be necessary to examine the listing carefully on the frequency of interest and adjacent frequencies up to 30 kHz away. Once the culling is complete, the standard / augmented pattern data will have to be obtained for stations using DAs from the FCC files. Sufficient information on omnidirectional antennas is provided in the data base listing itself. By examining the shape of the pattern of the most distant of the stations using DAs, sometimes a few more computations can be eliminated. For example, if the

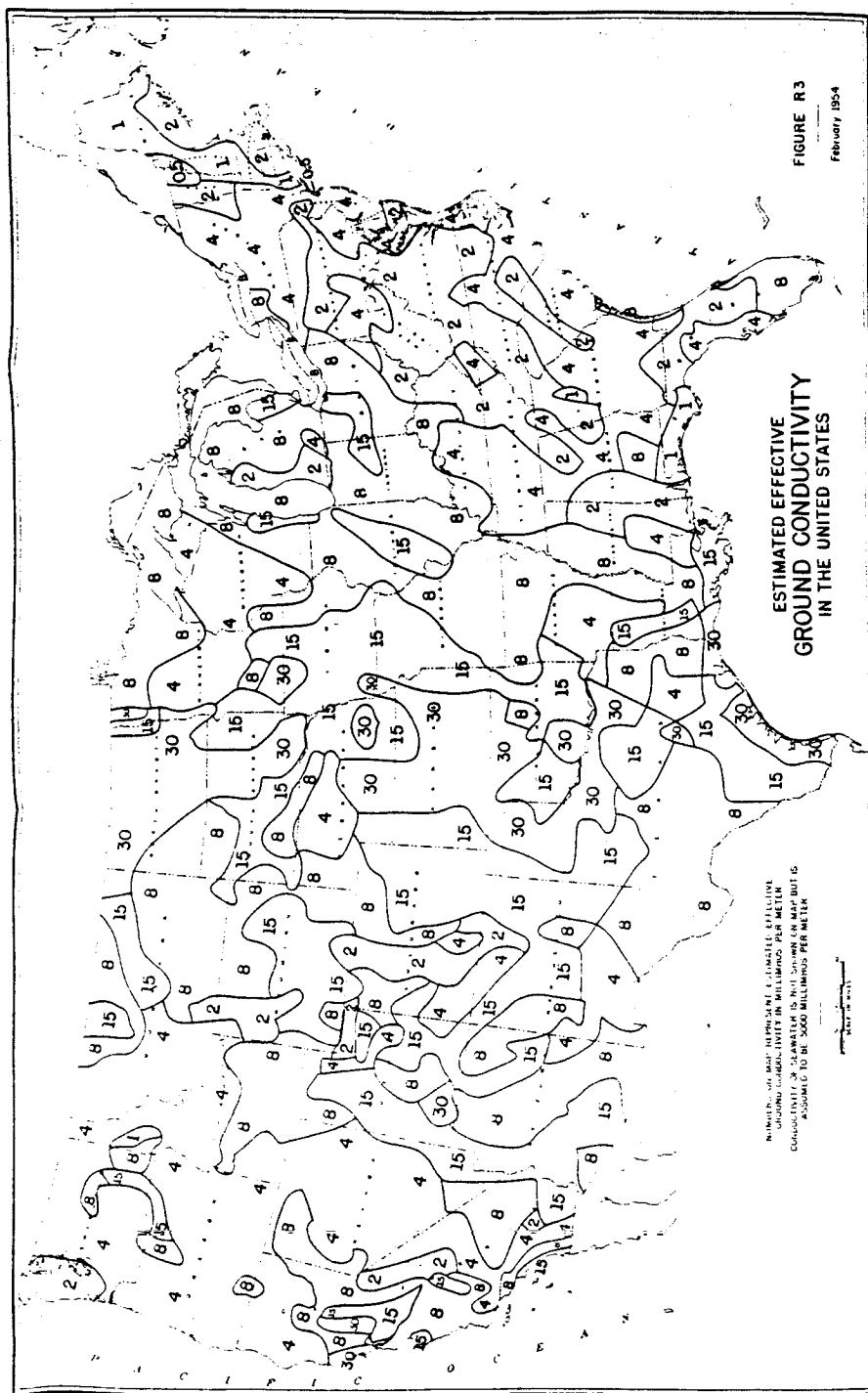


Figure 1. Ground Conductivity Map

radiation in the direction of the location of interest from station A is small, or on the order of that produced by another station that lies between station A and the desired location, then station A can usually be eliminated from further consideration.

After plotting the protected contour points of stations not culled on a suitable map, it becomes apparent where it is possible to locate a new station. If the desired market area is contained within the protected service area of any station using the desired frequency or adjacent frequencies, then it is impossible to locate a station at that site, and a new location or frequency must be chosen. If the protected contours of other stations are some distance from the desired site, then it may be possible to locate a station at the site. This is the point in the procedure where a specific location must be chosen for the antenna.

Once the precise location of the transmitter site is established, the engineer can attempt to design an antenna such that the desired service area provides a signal free from interference and fading and will not produce interference to other stations. At this point, there exists some disagreement between the work of Smith and the author on how to proceed with the synthesis. Smith states that " the pattern of each station is arranged so that a minimum of energy is directed toward other stations on the same channel. In addition, the arrays are located in a position with respect to their primary service area so that the main lobe of energy covers this primary area as completely as possible." The latter of these two statements is a well known technique of locating the antenna, and the author is in full agreement on this point. The first statement, however, must be clarified. When designing an antenna for the purpose of limiting interference, it is not necessarily in the

direction of the other transmitters where minimum energy should be directed. Instead, minimum energy should be directed toward protected service areas of the other stations.

What is generally done is to examine the jigsaw-like shape produced by the overlapping protected contours of other stations around the proposed site. Based on the shape of this area and the desired service area, the antenna parameters are estimated which are required to produce a radiation pattern that will provide service to as much of this jigsaw-like area as possible, yet not interfere with the other stations. Once the parameters are estimated and the standard pattern computed, the distance out to the maximum allowable field strength at the other station's protected contour is computed. If the point corresponding to this field strength level is not within the protected service area of another station, there is no objectionable interference.

At the beginning of the design, as outlined above, the engineer judges the parameters based on the shape of the unserved area. In other words, the shape of the antenna pattern is assumed to approximate the shape of the unserved area. This assumption can be very misleading, since the field strength at a constant radius about the antenna (the pattern) is not linearly related to the distance to the protected contour of the other stations. Moreover, when the conductivity changes greatly from one bearing to the next, the shape of the unserved area has almost nothing in common with the shape of the antenna pattern required to provide coverage to it.

Instead of computing the distance on each radial to the maximum allowable interfering field strength and then ascertaining if objectionable interference will result for every intermediate antenna change, or using the shape of the unserved area as a guide to the pattern shape, the author proposes a

different technique. This method, if possible, will eliminate unnecessary computations and provide the antenna designer with an actual pattern envelope with which to proceed with design. In addition, the method will allow the techniques presented in Smith's books to be used more effectively, and hopefully lead to better and faster solutions to antenna synthesis problems.

The main objective of the method is to determine the unattenuated field strength at one kilometer from the proposed site that will produce the maximum allowable field strength at the protected contour points of the other stations. The maximum allowable field strength at the protected contour points is given in the FCC rules, and as stated previously, is dependent on the class of the station and whether the stations are operating on the same or adjacent channels. Typically, the interfering signal should be a factor of 1/20 the desired signal for cochannel stations. For adjacent channel stations, the level of the interfering signal can be equal to the level of the desired signal.

Even though the method would be useful, the exact methodology for solving the problem has not yet been solved for all cases. The method will work is when the soil conductivity between the transmitter site and the protected contour point is homogeneous. One must determine the attenuation resulting from propagation over the homogeneous path using FCCGW, then divide the maximum allowable field strength at the protected contour point by this attenuation. When the conductivity between transmitter and contour point is not homogeneous, the reverse of the FCC equivalent distance method must be employed. The author has spent considerable time looking at this problem, but as of yet has not found a solution. It seems that even though an exact solution is not available, it would still be better to estimate the pattern size and shape using the homogeneous case rather than guessing from the shape

of the service area. One could use the maximum value of conductivity found in a given path over the whole path, thus providing some margin of protection from interference in that particular direction.

CHAPTER IV. EXAMPLE PROBLEMS

A. FEASIBILITY STUDY FOR SMETHPORT, PENNSYLVANIA ON 560 KHZ

Due to the costs of obtaining the directional antenna information required for the studies, only one location and corresponding frequency was analyzed. The data base listing for stations on frequencies from 550 kHz to 570 kHz was obtained, along with all standard / augmented pattern information for stations using DAs on these frequencies. A feasibility study was performed for operation in the Smethport market area on 560 KHz.

The culling process described in the previous section was applied to the stations in the data base listing until the number of stations involved was small enough to begin actual computational work. At this point, a guess was made on which station would most likely prevent the Smethport station from becoming reality. The guess was WGR, a Class III station in Buffalo, New York. Class III stations are protected to the 0.5 mV/m contour. This station uses an omnidirectional antenna during daylight hours with a power of 5 kW. From the data base listing, the unattenuated field strength at one kilometer for WGR is 294.51 mV/m/kW, thus the total unattenuated field strength is 1472.55 mV/m at one kilometer. With the field strength established, the next step was to determine the conductivity profile along the various radials from the WGR's transmitter site. Since a copy of Map M-3 was not available, conductivity values were read from Map R-3. In the area of Buffalo, the conductivity is given as 8 mS/m which abruptly changes to 4 mS/m in a very short distance as one proceeds southeast toward Smethport from Buffalo. For the purposes estimating the feasibility of the location, a sigma of 4 was used

for the entire path. This data was entered into the PCMLL2 program, and the resulting distance to the 0.5 mV/m contour of WGR was computed to be 186 kilometers. The distance between the the WGR transmitter site, and the Smethport location is only about 115 kilometers. Thus, it is impossible to install a new station in the Smethport area on 560 kHz.

B. POTENTIAL SERVICE AREAS ON 560 kHz IN PENNSYLVANIA

Since the Smethport location was deemed unsuitable, and data were not available for the other frequencies of interest, it was decided to determine if there was any location in Pennsylvania that might be suitable for a broadcast station on 560 Khz. Since the culling process for the Pennsylvania region had already been performed, the next step was to determine the protected contour points of these stations and plot them on a suitable map. The stations remaining for consideration after the culling process were:

1. WGR, Buffalo, New York. 5 kW omnidirectional on 550 kHz.
2. WHLM, Bloomsburg, Pennsylvania. 1 kW direcional on 550 kHz
3. WFRB, Frostburg, Maryland. 5 kW omnidirectional on 560 kHz.
4. WCKL, Catskill, New York. 1 kW direcional on 560 kHz.
5. WFIL, Philadelphia, Pennsylvania. 5 kW directional on 560 kHz.
7. WYSR, Syracuse, New York. 1 kW directional on 570 kHz.

The protected contours of each of these stations was computed using the PCMLL2 program. The PCMLL2 program provides the bearing and distance to the contour point, the lattitude and longitude of the point, and the bearing and distance of these points from the proposed site. After several station contours had been established, the area around Williamsport, Pennsylvania

seemed promising as a service area. The program output showing the location of the protected contour points of each of the stations listed above, and the bearing and distance of each of these contour points to a possible transmitter site in the Williamsport area are shown in Appendix D. By looking at the distance of the contour point to the proposed site, one can find the points that will be of significance in establishing a maximum pattern envelope. For example, on a given radial from the proposed site, there may be several protected contour points that fall within a few degrees of the bearing. There are always at least two for each station, one corresponding to the side of the protected contour nearest the site, and the other the side farthest from the site. Occasionally, there may even be more than two from a single station, and there are generally be some from other stations as well.

The question arises which of the many contour points are significant in the interference analysis, and in establishing the maximum pattern envelope. In the case of points from the same station, the significant contour point is the one nearest to the proposed site. When there is more than one station involved, this is not necessarily the case. The one that will limit the maximum field strength at one kilometer to the smallest value is the one that is most significant. In other words, if there is a protected contour point of an adjacent channel 40 miles from the proposed site, and a cochannel one 50 miles distant, the most significant one is probably the cochannel one at 50 miles since the field must be $1/20^{\text{th}}$ the protected contour at this point, and it can equal the protected contour at 40 miles.

Some of the protected contour points in the area around Williamsport are shown in Figure A1. The maps used for guiding this sketch were from the United States Geological Survey 1:500,000 series. Maps titled State of

Pennsylvania, and State of New York were necessary.

Again, due to the fact that a copy of Map M-3 was unavailable, R-3 values were used. To further simplify the analysis, a conductivity of 4 mS/m was used to compute the protected contours of all stations except WHLM, Bloomsburg, Pennsylvania. Examination of Map R-3 will reveal that most of Pennsylvania has a conductivity of either 2 or 4 mS/m. Thus, by assuming 4, distances to the protected contour points for paths that traverse through areas where the conductivity is really 2, will be greater than those computed if the equivalent distance was precisely applied. This error is of little significance for a feasibility study of this type. A value of 2 mS/m was used for computation of the protected contour of WHLM since it was centrally located in a large area with that conductivity value.

CHAPTER V. SUMMARY

A. RESULTS

The objectives of the project have been met. The use of antenna and wave propagation theory as tools for performing interference analyses used in AM broadcasting has been documented. The objective was not to be all inclusive, but rather as informative as possible within the confines of the problem report setting. The illustration of the use of the tools of antenna and propagation theory in a practical design problem has been illustrated to the extent possible given time and budget constraints. Clearly, by the analyses in Section III, it is not possible to locate a broadcast station in the Smethport, Pennsylvania area using a frequency of 560 kHz. Despite the null result, the problem illustrates a practical example of a real world problem. Secondly, even though time constraints did not allow the synthesis of a suitable antenna, the fact that a possible service area has been identified in the Williamsport, Pennsylvania area on a frequency of 560 kHz is still another demonstration of the use of antenna and propagation theory for interference analyses and possible antenna synthesis.

B. CONCLUSIONS

The project has served several useful purposes. It should be possible, with this document, a copy of the CFR Title 47 Parts 73., and the two books by Smith (See references 13 and 14), for a senior level electrical engineering undergraduate to become familiar and productive with the tools of the trade in

AM broadcast antenna engineering for daytime service.

The solutions to the practical problems, even though not as immediately profitable as one might like, are demonstrations of the tools of the trade, and constitute useful information to the WPLW personnel.

C. RECOMMENDATIONS

One suggestion, would be to extend the tutorial to the analysis of skywave interference. This is a much more involved problem and would be the next logical step in the work. Also, determination of a solution to the problem of working the equivalent distance method backwards would be of use in the engineering community. It might be interesting to find out if there is an analog to this problem in the case of the skywave propagation. Finally, it is recommended that the other four suggested locations and frequencies be evaluated for their potential development for broadcast use.

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APPENDIX A. STDPATRN SAMPLE INPUTS, OUTPUTS, AND PROGRAM LISTING

The program STDPATRN computes the standard and theoretical patterns of directional array antennas pursuant to the FCC Rules and Regulations. Use is made of the techniques presented by Smith (See reference 14) in computing the pattern size, and the driving point impedances when possible. For the case of antennas with tower heights not equal to 90 electrical degrees, information on the radius of the tower must be provided, if it is not, the program computes a pattern without including losses, and no impedance information is available. The same is true when the towers used are either top loaded or sectionalized, since analytical formulations are not generally available for evaluating the base and radiation impedances of these structures.

The program uses a simple card deck for input which can easily be expanded and modified according to individual needs. The deck should be set up in the following format.

- CM - Indicates that the next card will contain the number of comment lines. These lines will follow immediately there after.
- TR - Indicates that the next card will contain the number of towers in the array. Immediately following this, will be two cards for each tower in the array. The first of these cards will contain:
 - spacing degrees, bearing degrees, field ratio, phasing degreesThe second of these cards will contain:
 - tower type, A-sect degrees, B-sect degrees, C-sect degrees, D-sect degrees, radius degrees
- PW - Indicates the next card will contain the power in kiloWatts.
- PT - Indicates the next card will contain the information on pattern cuts. This card will contain the following:
 - theta start, theta step, number theta steps, phi start, phi step, number phi steps
- EN - Indicates end of data set.

A sample data set is shown in Figure A1, and the resulting output is shown in Figure A2. Following this is the actual program listing.

```

CM
2
WCKL 560 KHZ, CATSKILL, NY DAYTIME PATTERN
3 ELEMENT DIRECTIONAL ARRAY
PW
1.00
TR
3
0.0, 0.0, 1.0, -149.
1, 90.0, 0.0, 0.0, 0.0, 0.0
60.0, 140.0, 1.96, 0.0
1, 90.0, 0.0, 0.0, 0.0, 0.0
120.0, 140.0, 1.0, 149.0
1, 90.0, 0.0, 0.0, 0.0, 0.0
PT
0.0, 1.0, 0, 0.0, 10.0, 35
EN

```

Figure A1. Sample Input Deck

WCKL 560 MHZ, CATSKILL, NY DAYTIME PATTERN
3 ELEMENT DIRECTIONAL ARRAY

ARRAY INPUT POWER : 1.00 KW

ARRAY SPECIFICATIONS :

TWR #	SPACING	BEARING	RATIO	PHASE
1	0.00	0.00	1.0000	-149.00
2	60.00	140.00	1.9600	0.00
3	120.00	140.00	1.0000	149.00

TOWER SPECIFICATIONS :

TWR #	A-SECT	B-SECT	C-SECT	D-SECT	RADIUS
1	90.00	0.00	0.00	0.00	0.00
2	90.00	0.00	0.00	0.00	0.00
3	90.00	0.00	0.00	0.00	0.00

CALCULATED TOWER BASE IMPEDANCE

TOWER # 1	37.53981	21.25775
TOWER # 2	37.53981	21.25775
TOWER # 3	37.53981	21.25775

MUTUAL IMPEDANCE BETWEEN TOWERS :

TOWER # 1 TO 2	28.91697	-5.71276
TOWER # 1 TO 3	10.70145	-18.38202
TOWER # 2 TO 1	28.91697	-5.71276
TOWER # 3 TO 1	10.70145	-18.38202

DRIVING POINT IMPEDANCE

TOWER # 1	-10.31348	41.86661
TOWER # 2	12.13476	26.28417
TOWER # 3	4.61341	2.58416

REFERENCE CURRENT = 5.29360112
POWER RADIATED IS 0.836301982
K FACTOR = 316.568604
THEORETICAL ERMS = 316.679199 MV/M @ 1KM

PATTERN FIELD INTENSITY VALUES
(MV/M AT 1 KM)

THETA	PHI	THEORETICAL	STANDARD
0.00	0.00	477.646484	501.528809
0.00	10.00	399.440674	419.412598
0.00	20.00	313.522949	329.198975
0.00	30.00	226.520218	237.846268
0.00	40.00	145.588965	152.973434
0.00	50.00	77.769882	81.658371
0.00	60.00	27.716765	29.134094
0.00	70.00	2.133583	2.208762
0.00	80.00	12.566341	13.194660
0.00	90.00	7.138972	7.506421
0.00	100.00	8.804175	9.244386
0.00	110.00	29.737442	30.699310
0.00	120.00	48.351111	50.873703
0.00	130.00	61.906479	65.001801
0.00	140.00	66.710903	70.055801
0.00	150.00	61.906479	65.001801
0.00	160.00	48.351111	50.873703
0.00	170.00	29.737442	30.699310
0.00	180.00	8.804175	9.244386
0.00	190.00	7.138972	7.506421
0.00	200.00	12.566341	13.194660
0.00	210.00	2.133583	2.208762
0.00	220.00	27.716765	29.134094
0.00	230.00	77.769882	81.658371
0.00	240.00	145.588965	152.973434
0.00	250.00	226.520218	237.846268
0.00	260.00	313.522949	329.198975
0.00	270.00	399.440674	419.412598
0.00	280.00	477.646484	501.528809
0.00	290.00	512.891602	570.036133
0.00	300.00	591.549072	621.126465
0.00	310.00	621.451116	652.523926
0.00	320.00	631.523447	663.099609
0.00	330.00	621.451116	652.523926
0.00	340.00	591.549072	621.126465
0.00	350.00	512.891602	570.036133

Figure A2. Sample Output

```

*****
C *****
C STD PATRN - THIS PROGRAM COMPUTES THE STANDARD PATTERN FOR
C AM DIRECTIONAL BROADCAST ANTENNAS ACCORDING TO RULES SET FORTH
C IN THE FEDERAL COMMUNICATIONS COMMISSION RULES AND REGULATIONS
C PART 73.150.
C
C IMPLICIT REAL*4 (K), COMPLEX (Z)
C
C COMMON / ARRAY / S(10), HR(10), FR(10), PH(10)
C COMMON / PARAM / NT, PKW, PRKW, K
C COMMON / TOWER / ISECT(10), A(10), R(10), C(10), D(10), P(10)
C COMMON / IMPED / Z(10,10), ZS(10), ZSI(10), ZR(10), ZRI(10)
C COMMON / PAI / THETA, DTHETA, MTHETA, PHI, DPHI, NPHI
C COMMON / INOUT / IN, IO
C COMMON / CONST / PI, DTR
C
C INITIALIZE INPUT OUTPUT DEVICES.
C
C IN = 5
C IO = 6
C
C INITIALIZE CONSTANTS
C
C PI = 3.1415927
C DTR = PI / 180.
C
C READ ARRAY PARAMETERS FROM DATA FILE
C
C CALL INPUT
C
C IF TOWER HEIGHT NOT ODD MULTIPLE OF 90.0, AND RADIUS = 0.0
C THEN CANT INCLUDE LOSSES. THUS, ASSUME EFFICIENCY = 100%.
C
C DO 5, J = 1, NT
C IF ( R(J) .EQ. 0.0 .AND. A(J) .NE. 90.0 ) THEN
C PRKW = PKW
C GOTO 50
C
C 5 ENDOF
C CONTINUE
C
C COMPUTE RADIATION AND MUTUAL IMPEDANCES OF ARRAY ELEMENTS
C
C CALL ZCALC
C
C COMPUTE BASE SELF IMPEDANCES OF EACH TOWER
C
C CALL ZBASE
C
C SOLVE FOR CURRENT IN REFERENCE TOWER ASSUMING LOSSES
C
C SSV = 0.0
C
C DO 26, J = 1, NT
C 26 SSV = SSV + FR(J)*FP(J) * REAL ( ZRI(J) )
C
C REF = SQRT ( PKW * 1000. / SSV )
C WRITE ( IO, * ) 'REFERENCE CURRENT = ', REF
C
C COMPUTE POWER RADIATED BY ARRAY
C
C PRAD = 0.0
C DO 27, J = 1, NT
C 27 PRAD = PRAD + REAL( ZR(J) ) * ( FR(J) * REF ) ** 2
C PRKW = PRAD / 1000.
C WRITE ( IO, * ) 'POWER RADIATED IS ', PRKW
C
C COMPUTE K FACTOR FOR ARRAY.
C
C 50 CALL KFACTOR ( THRM50 )
C
C WRITE ( IO, * ) 'K FACTOR = ', K
C
C COMPUTE THEORETICAL ERMS IN HORIZONTAL PLANE
C
C THERMS = K * THRM50
C
C WRITE ( IO, * ) 'THEORETICAL ERMS = ', THERMS, 'mV/M @ 1KM'
C
C COMPUTE PATTERN VALUES FOR SPECIFIED THETA AND PHI VALUES

```

```

STD00010
STD00020
STD00030
STD00040
STD00050
STD00060
STD00070
STD00080
STD00090
STD00100
STD00110
STD00120
STD00130
STD00140
STD00150
STD00160
STD00170
STD00180
STD00190
STD00200
STD00210
STD00220
STD00230
STD00240
STD00250
STD00260
STD00270
STD00280
STD00290
STD00300
STD00310
STD00320
STD00330
STD00340
STD00350
STD00360
STD00370
STD00380
STD00390
STD00400
STD00410
STD00420
STD00430
STD00440
STD00450
STD00460
STD00470
STD00480
STD00490
STD00500
STD00510
STD00520
STD00530
STD00540
STD00550
STD00560
STD00570
STD00580
STD00590
STD00600
STD00610
STD00620
STD00630
STD00640
STD00650
STD00660
STD00670
STD00680
STD00690
STD00700
STD00710
STD00720
STD00730
STD00740
STD00750
STD00760
STD00770
STD00780
STD00790
STD00800

```

```

C      CALL PATTN
C      END
C*****
C      INPUT - THIS SUBROUTINE READS THE CARDS CONTAINING THE INFORMATION
C      REQUIRED FOR SIMPATM AND PRINTS OUT THE INFO FOR REFERENCE.
C      SUBROUTINE INPUT
C      IMPLICIT REAL (A), COMPLEX (Z)
C      CHARACTER * 80 CM(4)
C      CHARACTER * 2 CARD
C      COMMON / ARKAY / S(10), BR(10), FR(10), PH(10)
C      COMMON / PAKK / NT, PKW, PRKW, K
C      COMMON / TOWN / ISECT(10), A(10), B(10), C(10), D(10), P(10)
C      COMMON / AUG / NAUG, PHIP(30), SPAN(30), EAUGO(30)
C      COMMON / PAT / THETA0, DTHETA, NTHETA, PHIO, DPHI, NPHI
C      COMMON / INOUT / JI, IO
C
10  READ ( IN, 1000 ) CARD
    IF ( CARD.EQ. 'CA' ) THEN
        READ ( IN, * ) NC
        DO 20, J = 1, NC
20     READ ( IN, 1001 ) CM(J)
        ENDIF
        IF ( CARD.EQ. 'TR' ) THEN
            READ ( IN, * ) NT
            DO 30, J = 1, NT
30     READ ( IN, * ) S(J), BR(J), FR(J), PH(J)
            READ ( IN, * ) ISECT(J), A(J), B(J), C(J), D(J), P(J)
            ENDIF
            IF ( CARD.EQ. 'PW' ) THEN
                READ ( IN, * ) PKW
            ENDIF
            IF ( CARD.EQ. 'AG' ) THEN
                READ ( IN, * ) NAUG
                DO 40, J = 1, NAUG
40     READ ( IN, * ) PHIP(J), SPAN(J), EAUGO(J)
            ENDIF
            IF ( CARD.EQ. 'PT' ) THEN
                READ ( IN, * ) THETA0, DTHETA, NTHETA, PHIO, DPHI, NPHI
            ENDIF
            IF ( CARD.EQ. 'EN' ) THEN
                GOTO 35
            ELSE
                GOTO 10
            ENDIF
C
C      PRINT OUT INFO READ FROM DATA FILE
C
35  DO 50, J = 1, NC
50  WRITE ( IO, 1001 ) CM(J)
        WRITE ( IO, 2000 ) PKW
        WRITE ( IO, 2001 )
C
        DO 60, J = 1, NT
60  WRITE ( IO, 2002 ) J, S(J), BR(J), FR(J), PH(J)
        WRITE ( IO, 2003 )
C
        DO 70, J = 1, NAUG
70  WRITE ( IO, 2004 ) J, A(J), B(J), C(J), D(J), P(J)
        IF ( NAUG.EQ. 0 ) GOTO 100
        WRITE ( IO, 2005 )
C
        DO 80, J = 1, NAUG
80  WRITE ( IO, 2006 ) J, PHIP(J), SPAN(J), EAUGO(J)
C
100 RETURN
C
1000 FORMAT (A2)

```

```

STD00810
STD00820
STD00830
STD00840
STD00850
STD00860
STD00870
STD00880
STD00890
STD00900
STD00910
STD00920
STD00930
STD00940
STD00950
STD00960
STD00970
STD00980
STD00990
STD01000
STD01010
STD01020
STD01030
STD01040
STD01050
STD01060
STD01070
STD01080
STD01090
STD01100
STD01110
STD01120
STD01130
STD01140
STD01150
STD01160
STD01170
STD01180
STD01190
STD01200
STD01210
STD01220
STD01230
STD01240
STD01250
STD01260
STD01270
STD01280
STD01290
STD01300
STD01310
STD01320
STD01330
STD01340
STD01350
STD01360
STD01370
STD01380
STD01390
STD01400
STD01410
STD01420
STD01430
STD01440
STD01450
STD01460
STD01470
STD01480
STD01490
STD01500
STD01510
STD01520
STD01530
STD01540
STD01550
STD01560
STD01570
STD01580
STD01590
STD01600

```

```

1001 FORMAT (A0)
2000 FORMAT ( /, ' ARRAY INPUT POWER : ', F6.2, ' KW ', / )
2001 FORMAT ( /, ' ARRAY SPECIFICATIONS : ', /,
& ' TWR # SPACING BEARING RATIO PHASE', / )
2002 FORMAT ( 3X, 12, 6X, F7.2, 6X, F7.2, 6X, F7.4, 6X, F7.2 )
2003 FORMAT ( /, ' TOWER SPECIFICATIONS : ', /,
& ' TWR # A-SECT B-SECT C-SECT DSECT
& ' RADIUS', / )
2004 FORMAT ( 3X, 12, 5 ( 6X, F7.2 ) )
2005 FORMAT ( /, ' PATTERN AUGMENTATION DATA : ', /,
& ' AUG # PHI SPAN FAUG', / )
2006 FORMAT ( 3X, 12, 3 ( 6X, F7.2 ) )
C
END
C
*****
C
E - THIS SUBROUTINE COMPUTES THE THEORETICAL INVERSE DISTANCE FIELD
C AT 1 KILOMETER FOR THE GIVEN AZIMUTH AND ELEVATION.
C
C SUBROUTINE E ( THETA, PHI, ETH )
C
C IMPLICIT REAL (K)
C COMPLEX ET
C
C COMMON / ARRAY / S(10), RP(10), FR(10), PH(10)
C COMMON / PARAM / NT, PKW, PRKW, K
C COMMON / TOWER / ISECT(10), A(10), B(10), C(10), D(10), R(10)
C COMMON / CONST / PI, DTR
C
C INITIALIZE VARIABLES
C
C ET = ( 0.0, 0.0 )
C
C DO VECTOR SUM OF CONTRIBUTION FROM EACH TOWER
C DO 10, I = 1, NT
C
C GET THE VERTICAL PLANE RADIATION CHARACTERISTIC F( THETA )
C CALL FELV ( THETA, I, F )
C
C ET = ET + FR(I) * F * EXP ( ( 0.1 ) * DTR * ( S(I) *
& COS ( DTR * THETA ) * COS ( DTR * ( RP(I) - PHI ) ) +
& PH(I) ) )
C
C 10 CONTINUE
C
C TAKE MAGNITUDE OF COMPLEX FIELD
C ETH = ABS ( ET )
C
C RETURN
C
C END
C
*****
C
FELV - COMPUTES THE ELEVATION PATTERN FACTOR FOR USE IN FIELD
C CALCULATIONS.
C
C SUBROUTINE FELV ( THETA, I, F )
C
C COMMON / TOWER / ISECT(10), A(10), B(10), C(10), D(10), R(10)
C COMMON / CONST / PI, DTR
C
C CONVERT ALL LENGTHS AND ANGLES FROM DEGREES TO RADIANS
C
C TH = DTR * THETA
C A1 = DTR * A(I)
C B1 = DTR * B(I)
C C1 = DTR * C(I)
C D1 = DTR * D(I)
C G = A1 + B1
C H = C1 + D1
C DEL = H - A1
C
C FOR STANDARD TOWER
C IF ( ISECT(1) .EQ. 1 ) THEN

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C      F = ( COS ( A1 * SIN ( TH ) ) - COS ( A1 ) ) /
C      ( ( 1.0 - COS ( A1 ) ) * COS ( TH ) )
C      ENDIF
C      FOR TOP LOADED TOWER
C      IF ( ISECT(I) .EQ. 2 ) THEN
C      FNUM = COS ( R1 ) * COS ( A1 * SIN ( TH ) ) -
C      SIN ( TH ) * SIN ( B1 ) * SIN ( A1 * SIN ( TH ) ) -
C      COS ( A1 + R1 )
C      FDEN = COS ( TH ) * ( COS ( R1 ) - COS ( A1 + R1 ) )
C      F = FNUM / FDEN
C      ENDIF
C      FOR A SECTIONALIZED TOWER WITH TOP LOADING
C      IF ( ISECT(I) .EQ. 3 ) THEN
C      FNUM = ( SIN ( DEL ) * ( COS ( R1 ) * COS ( A1 * SIN ( TH ) )
C      - COS ( G ) ) +
C      SIN ( R1 ) * ( COS ( D1 ) * COS ( C1 * SIN ( TH ) ) -
C      SIN ( TH ) * SIN ( D1 ) * SIN ( C1 * SIN ( TH ) ) -
C      COS ( DEL ) * COS ( A1 * SIN ( TH ) ) ) )
C      FDEN = COS ( TH ) * ( SIN ( DEL ) * ( COS ( R1 ) - COS ( C1 ) )
C      + SIN ( R1 ) * ( COS ( D1 ) - COS ( DEL ) ) )
C      F = FNUM / FDEN
C      ENDIF
C      RETURN
C      END
C      *****
C      KFACTOR - THIS SUBROUTINE COMPUTES THE K FACTOR THAT DETERMINES THE
C      ARRAY SIZE. POWER FLOW INTEGRATION IS USED TO GET THE RMS FIELD
C      INTENSITY.
C      SUBROUTINE KFACTOR ( THRMSO )
C      IMPLICIT REAL (K)
C      COMMON / PARM / NT, PKW, PRKW, F
C      COMMON / CONST / PI, DTR
C      INITIALIZE VARIABLES
C      E2P = 0.0
C      E2T = 0.0
C      SET THETA AND PHI INCREMENTS TO 1.0 DEGREE
C      DTHETA = 1.0
C      DPHI = 1.0
C      INTEGRATE OVER UPPER HEMISPHERE.
C      DO 20, I = 0, 89
C      THETA = I * DTHETA
C      F2P = 0.0
C      DO 10, J = 0, 359
C      PHI = J * DPHI
C      CALL F ( THETA, PHI, ETH )
C      E2P = E2P + ETH * * 2
C      10 CONTINUE
C      IF ( I .EQ. 0 ) THEN

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C GET FACTOR FOR HORIZONTAL PLANE THEORETICAL RMS FIELD
C THRSO = SQRT ( E2P / 360.)
C
C F2T = E2P / 2
C ELSE
C F2T = E2T + COS ( DTR * THETA ) * F2P
C ENDIF
C 20 CONTINUE
C
C FP = SQRT ( ( PI / ( 180. * 360. ) ) * E2T )
C
C K = 244.73 * SQRT ( PRKW ) / FP
C
C RETURN
C
C END
C *****
C ZCALC - THIS SUBROUTINE COMPUTES THE RADIATION AND MUTUAL IMPEDANCE
C OF THE ARRAY ELEMENTS. ANALYTICAL EXPRESSIONS ARE EVALUATED FOR
C STANDARD TOWERS NOT USING TOP LOADING OR SECTIONALIZATION.
C FOR TOP LOADED OR SECTIONALIZED TOWERS.
C
C SUBROUTINE ZCALC
C
C IMPLICIT REAL ( K ), COMPLEX ( Z )
C
C COMPLEX RAD1, MUT1
C
C EXTERNAL RAD1, MUT1
C
C COMMON / ARRAY / S(10),BR(10),FP(10),PH(10)
C COMMON / PAWM / NT, PKW, PRKW, K
C COMMON / TOWER / ISECT(10), A(10), B(10), C(10), D(10), R(10)
C COMMON / IMPED / Z(10,10), ZS(10), ZSL(10), ZB(10), ZBL(10)
C COMMON / CONST / PI, DTR
C
C COMPUTE IMPEDANCE OF EACH TOWER TO ITSELF AND EVERY OTHER TOWER
C
C DO 20, J = 1, NT
C DO 10, I = 1, J
C
C IF ( ( I.EQ. J ) .AND. ( ISECT( I ) .EQ. 1 ) ) THEN
C COMPUTE RADIATION IMPEDANCE OF ELEMENT I
C
C Z( I, I ) = RAD1 ( A(I) )
C
C ENDIF
C
C IF ( ( I.NE. J ) .AND. ( ISECT( I ) .EQ. 1 ) .AND.
C ( ISECT( J ) .EQ. 1 ) ) THEN
C COMPUTE DISTANCE BETWEEN ELEMENTS I AND J
C
C SD = SQRT( S(J)*S(J) + S(I)*S(I) - 2.0 * S(J)*S(I) *
C COS( DTR * ( BR(I)-BR(J) ) ) )
C COMPUTE MUTUAL IMPEDANCE BETWEEN I AND J. USE SYMMETRY FOR J AND I.
C
C Z( I, J ) = MUT1 ( A(I), A(J), SD )
C Z( J, I ) = Z( I, J )
C
C ENDIF
C 10 CONTINUE
C 20 CONTINUE
C
C RETURN
C END
C *****
C ZBASE - COMPUTES THE BASE SELF IMPEDANCES AND DRIVING POINT
C IMPEDANCES FOR ALL TOWERS. BOTH LOSSLESS AND LOSSY CASES ARE
C CONSIDERED.
C
C SUBROUTINE ZBASE

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C      IMPLICIT REAL (R), COMPLEX (Z)
C
C      COMMON / ARRAY / S(10), BR(10), FR(10), PH(10)
C      COMMON / PARAM / NT, PKW, PRKW, K
C      COMMON / TOWER / ISPECT(10), A(10), B(10), C(10), D(10), R(10)
C      COMMON / IMPED / Z(10,10), ZS(10), ZSL(10), ZR(10), ZRL(10)
C      COMMON / CONST / PI, DTR
C      COMMON / INPUT / IN, TO
C
C      COMPUTE BASE SELF IMPEDANCES OF EACH TOWER
C      THE ZS(I) ARE BASE IMPEDANCES NEGLECTING LOSSES
C      THE ZSL(I) ARE BASE IMPEDANCES INCLUDING LOSSES
C
C      WRITE ( IO, 2000 )
C
C      DO 22, J = 1, NT
C
C      TR = SIN ( DTR * A(J) ) ** 2.0
C
C      ZS(J) = Z(J,J) / TR
C
C      IF ( A(J) .GE. 90.0 ) THEN
C        ZSL(J) = ( 1.0 + Z(J,J) ) / TR
C      ELSE
C        ZSL(J) = 1.0 + Z(J,J)
C      ENDIF
C
C      WRITE ( IO, 2001 ) J, REAL( ZSL(J) ), AIMAG( ZSL(J) )
C
C 22  CONTINUE
C
C      PRINT OUT MUTUAL IMPEDANCES BETWEEN TOWERS
C
C      WRITE ( IO, 2002 )
C
C      DO 23, J = 1, NT
C      DO 23, I = 1, NT - J + 1
C      IF ( J .EQ. 1 ) GOTO 23
C      WRITE ( IO, 2003 ) J, I, REAL( Z(J,I) ), AIMAG( Z(J,I) )
C 23  CONTINUE
C
C      WRITE ( IO, 2004 )
C
C      COMPUTE BASE IMPEDANCE ACCOUNTING FOR MUTUAL EFFECTS
C      ZB(I) IS THE BASE IMPEDANCE NEGLECTING LOSS
C      ZBL(I) IS THE BASE IMPEDANCE INCLUDING LOSS
C
C      DO 25, J = 1, NT
C
C      ZB(J) = ( 0.0, 0.0 )
C      ZBL(J) = ( 0.0, 0.0 )
C
C      DO 24, I = 1, NT
C
C      IF ( I .EQ. J ) GOTO 24
C
C      ZR(J) = ZR(J) + FR(I) / FR(J) * EXP((0., 1.) * DTR * (PH(I) - PH(J))) * Z(J, I)
C
C 24  CONTINUE
C
C      ZBL(J) = ZB(J) + ZSL(J)
C      ZRL(J) = ZS(J) + ZS(J)
C
C      WRITE ( IO, 2005 ) J, REAL( ZRL(J) ), AIMAG( ZRL(J) )
C
C 25  CONTINUE
C
C 2000 FORMAT ( /, ' CALCULATED TOWER BASE IMPEDANCE ' / )
C 2001 FORMAT ( ' TOWER # ', I2, ' SX, F10.5, SX, F10.5 )
C 2002 FORMAT ( /, ' MUTUAL IMPEDANCE BETWEEN TOWERS : ' / )
C 2003 FORMAT ( ' TOWER # ', I2, ' TO ', I2, ' SX, F10.5, SX, F10.5 )
C 2004 FORMAT ( /, ' DRIVING POINT IMPEDANCE ' / )
C 2005 FORMAT ( ' TOWER # ', I2, ' SX, F10.5, SX, F10.5 )
C
C      RETURN
C
C      END
C*****

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C      RAD1 - THIS FUNCTION EVALUATES THE RADIATION IMPEDANCE OF THE
C      TOWER REFERRED TO THE CURRENT MAXIMUM WHEN NO OTHER TOWERS ARE
C      PRESENT.
C      FUNCTION RAD1( H )
C      EXTERNAL SI, CI
C      REAL IM, L
C      COMPLEX RAD1
C
C      PI = 3.1415927
C      DTR = PI / 180.
C      ETA = 376.7343
C      C = 0.577215664901532
C
C      L = DTR * 2.0 * H
C
C      COMPUTE THE REAL PART OF THE IMPEDANCE
C
C      RE = ( ETA / ( 4.0 * PI ) ) * ( C + ALOG ( L ) - CI ( L ) ) +
C      & 0.5 * SIN ( L ) * ( SI ( 2.0 * L ) - 2.0 * SI ( L ) ) +
C      & 0.5 * COS ( L ) * ( C + ALOG ( L / 2.0 ) + CI ( 2.0 * L ) ) -
C      & 2.0 * CI ( L ) )
C
C      COMPUTE THE IMAGINARY PART OF THE IMPEDANCE
C
C      IM = ( ETA / ( 8.0 * PI ) ) * ( 2.0 * SI ( L ) + COS ( L ) *
C      & ( 2.0 * SI ( L ) - SI ( 2.0 * L ) ) - SIN ( L ) *
C      & ( 2.0 * CI ( L ) - CI ( 2.0 * L ) ) )
C
C      PAD1 = RE + ( 0.0, 1.0 ) * IM
C
C      END
C
C      *****
C      MUT1 - THIS FUNCTION COMPUTE THE MUTUAL IMPEDANCE BETWEEN TWO
C      TOWERS OF DIFFERENT HEIGHTS AND SEPARATED BY DISTANCE D.
C      ( FORMULAS GIVEN BY WERES REPORT )
C      FUNCTION MUT1 ( H1, H2, D )
C      COMPLEX MUT1
C      REAL IM
C      EXTERNAL CI, SI
C
C      PI = 3.1415927
C      DTR = PI / 180.0
C
C      G1 = DTR * H1
C      G2 = DTR * H2
C      S = DTR * D
C
C      U0 = SQRT ( S*S + G1*G1 ) - G1
C      U1 = SQRT ( S*S + (G2-G1)**2.0 ) + G2 - G1
C      V0 = SQRT ( S*S + G1*G1 ) + G1
C      V1 = SQRT ( S*S + (G2-G1)**2.0 ) - G1 + G1
C      W0 = V0
C      W1 = SQRT ( S*S + (G2+G1)**2.0 ) + G2 + G1
C      X0 = U0
C      X1 = SQRT ( S*S + (G2+G1)**2.0 ) - G2 - G1
C      Y0 = S
C      Y1 = SQRT ( S*S + G2*G2 ) + G2
C      S0 = Y0
C      S1 = SQRT ( S*S + G2*G2 ) - G2
C
C      CIU0 = CI( U0 )
C      CIU1 = CI( U1 )
C      CIV0 = CI( V0 )
C      CIV1 = CI( V1 )
C      CIW1 = CI( W1 )
C      CIX1 = CI( X1 )
C      CIY1 = CI( Y1 )
C      CIS1 = CI( S1 )
C      SIU0 = SI( U0 )
C      SIU1 = SI( U1 )
C      SIV0 = SI( V0 )
C      SIV1 = SI( V1 )

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      SIX1 = SI( X1 )
      SIV1 = SI( Y1 )
      SIS1 = SI( S1 )
C
      RE = ( 15.0 / ( SIN(G1)*SIN(G2) ) ) * ( COS(G2-G1) *
      & ( CIU1 - CIU0 + CIV1 - CIV0 + 2.0 * CIY0 - CIY1 - CIS1 )
      & + SIN(G2-G1) * ( SIU1 - SIU0 + SIV0 - SIV1 - SIY1 + SIS1 )
      & + COS(G2+G1) * ( CIW1 - CIV0 + CIX1 - CIU0 + 2.0 * CIY0
      & - CIY1 - CIS1 ) + SIN(G2+G1) * ( SIW1 - SIV0 + SIU0
      & - SIX1 - SIV1 + SIS1 ) )
C
      IM = ( 15.0 / ( SIN(G1)*SIN(G2) ) ) * ( COS(G2-G1) *
      & ( SIU0 - SIU1 + SIV0 - SIV1 + SIY1 - 2.0 * SIY0
      & + SIS1 ) + SIN(G1-G1) * ( CIU1 - CIU0 + CIV0
      & - CIV1 - CIV1 + CYS1 )
      & + COS(G2+G1) * ( SIV0 - SIV1 + SIU0 - SIX1 + SIY1
      & + SIS1 ) + SIN(G2+G1) * ( CIX1 - CIV0 + CIU0
      & - CIX1 - CIV1 + CYS1 ) )
C
      MUT1 = RE + ( 0.0, 1.0 ) * IM
C
      END
C
*****
CI - THIS FUNCTION EVALUATES THE COSINE INTEGRAL. THE INTEGRAL
CIN(X) IS FIRST EVALUATED USING AN IMSL ROUTINE. THEN APPROPRIATE
CORRECTION TERMS ARE ADDED TO GET THE CI(X) FUNCTION.
C
      FUNCTION CI ( ARG )
C
      REAL * 8 DCADRE, F1, A, B, AERR, REPR, ERROR
      INTEGER * 4 IER
      EXTERNAL F1
C
      C = 0.577215664901532
      A = 1.0 D -16
      B = ARG
      RERR = 1.0 D -16
      AERR = 1.0 D -16
C
      FUNCTION DCADRE INTEGRATES THE FUNCTION F1 BETWEEN A AND B
      WITH THE GIVEN VALUES OF ABSOLUTE AND RELATIVE ERROR.
C
      CI = DLOG(B) + C - DCADRE ( F1, A, B, AERR, RERR, ERROR, IER )
C
      END
C
*****
SI - THIS FUNCTION EVALUATES THE SINE INTEGRAL USING AN IMSL
INTEGRATION ROUTINE.
C
      FUNCTION SI ( ARG )
C
      REAL * 4 SI, ARG
      REAL * 8 DCADRE, F2, A, B, AERP, REPR, ERROR
      INTEGER * 4 IER
      EXTERNAL F2
C
      A = 1.0 D -16
      B = ARG
      REPR = 1.0 D -16
      AERR = 1.0 D -16
C
      THE FUNCTION DCADRE INTEGRATES FUNCTION F2 BETWEEN A AND B
      WITH THE GIVEN RELATIVE AND ABSOLUTE ERROR.
C
      SI = DCADRE( F2, A, B, AERR, RERR, ERROR, IER )
C
      END
C
*****
F1 - THIS FUNCTION IS USED BY FUNCTION CI TO COMPUTE THE COSINE
INTEGRAL
C
      FUNCTION F1 ( X )
      REAL * 8 F1, X

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      F1 = ( 1.0 - DCOS( Y ) ) / X
      END
C*****
C F2 - THIS FUNCTION IS USED BY FUNCTION S1 TO COMPUTE THE SINE
C INTEGRAL.
C
C   FUNCTION F2 ( X )
C   REAL*8 F2, X
C   F2 = DSIN( X ) / X
C   END
C*****
C PATTERN - COMPUTES THE THEORETICAL AND STANDARD PATTERN FOR THE
C VALUES OF THETA AND PHI SPECIFIED IN INPUT FILE
C
C   SUBROUTINE PATTERN
C   IMPLICIT REAL (K)
C
C   COMMON / PARM / NT, PKW, PRKW, K
C   COMMON / PAT / THETA0, DTHETA, NTHETA, PHI0, DPHI, LPHI
C   COMMON / AUG / NAUG, PHIP(30), SPAN(30), EAUG(30)
C   COMMON / INOUT / IN, IO
C
C   COMPUTE HEIGHT OF SMALLEST TOWER FOR Q VALUE COMPUTATION
C
C   CALL HMIN ( JM, HM )
C
C   LOOP OVER THETA, PHI AND COMPUTE PATTERN VALUES
C
C   WRITE ( IO, 2000 )
C
C   DO 10, THETA = THETA0, NTHETA * DTHETA, DTHETA
C
C   COMPUTE THE Q FACTOR FOR THETA
C
C   CALL OVAL ( THETA, JM, HM, Q )
C
C   DO 10, PHI = PHI0, NPHI * DPHI, DPHI
C
C   CALL E ( THETA, PHI, ET )
C
C   COMPUTE THE THEORETICAL VALUE OF THE ELECTRIC FIELD
C
C   ETH = K * ET
C
C   COMPUTE THE STANDARD PATTERN VALUE
C
C   ESTD = 1.05 * SQRT ( ETH*ETH + Q*Q )
C
C   COMPUTE AUGMENTED PATTERN VALUE IF AUGMENTATIONS SPECIFIED
C
C   IF ( NAUG .NE. 0 ) THEN
C     ENDIF
C
C   PRINT PATTERN INFORMATION
C
C   WRITE ( IO, 2001 ) THETA, PHI, ETH, ESTD
C   WRITE ( IO, * ) 'BP'
C   WRITE ( IO, 2001 ) PHI, ESTD
C   WRITE ( IO, * ) '4.0, 999999'
C
C 10 CONTINUE
C
C 2000 FORMAT ( ///, '          PATTERN FIELD INTENSITY VALUES ', /,
C   & '          ( MV/M AT 1 KM )', /,
C   & '          THETA          PHI          THEORETICAL          STANDARD', / )
C 2001 FORMAT ( 2X, F6.2, 6X, F6.2, 6X, F12.6, 6X, F12.6 )
C 2001 FORMAT ( 2X, F6.2, 2X, ' ', 2X, F12.6 )
C
C   RETURN
C
C   END
C*****
C HMIN - THIS SUBROUTINE DETERMINES THE SHORTEST TOWER USED IN THE
C ARRAY. THIS IS REQUIRED TO EVALUATE THE Q FACTOR.

```

```

STD06410
STD06420
STD06430
STD06440
STD06450
STD06460
STD06470
STD06480
STD06490
STD06500
STD06510
STD06520
STD06530
STD06540
STD06550
STD06560
STD06570
STD06580
STD06590
STD06600
STD06610
STD06620
STD06630
STD06640
STD06650
STD06660
STD06670
STD06680
STD06690
STD06700
STD06710
STD06720
STD06730
STD06740
STD06750
STD06760
STD06770
STD06780
STD06790
STD06800
STD06810
STD06820
STD06830
STD06840
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STD06880
STD06890
STD06900
STD06910
STD06920
STD06930
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STD06950
STD06960
STD06970
STD06980
STD06990
STD07000
STD07010
STD07020
STD07030
STD07040
STD07050
STD07060
STD07070
STD07080
STD07090
STD07100
STD07110
STD07120
STD07130
STD07140
STD07150
STD07160
STD07170
STD07180
STD07190
STD07200

```

C	SUBROUTINE HMIN (JM, HM)	STD07210
C	IMPLICIT REAL (K)	STD07220
C	COMMON / TOWER / ISECT(10), A(10), B(10), C(10), D(10), R(10)	STD07230
C	COMMON / PARM / NT, PKW, PRKW, K	STD07240
C	CHOOSE TOWER WITH SHORTEST ELECTRICAL HEIGHT.	STD07250
C	DO 5, J = 1, NT	STD07260
C	IF (ISECT(J) .EQ. 1) H = A(J)	STD07270
C	IF (ISECT(J) .EQ. 2) H = A(J) + B(J)	STD07280
C	IF (ISECT(J) .EQ. 3) H = C(J) + D(J)	STD07290
C	IF (J .EQ. 1) HM = H	STD07300
C	IF (H .LT. HM) THEN	STD07310
C	HM = H	STD07320
C	JH = J	STD07330
C	ENDIF	STD07340
C	CONTINUE	STD07350
C	RETURN	STD07360
C	END	STD07370
C	*****	STD07380
C	*****	STD07390
C	*****	STD07400
C	*****	STD07410
C	*****	STD07420
C	*****	STD07430
C	*****	STD07440
C	*****	STD07450
C	*****	STD07460
C	*****	STD07470
C	*****	STD07480
C	*****	STD07490
C	*****	STD07500
C	*****	STD07510
C	*****	STD07520
C	*****	STD07530
C	*****	STD07540
C	*****	STD07550
C	*****	STD07560
C	*****	STD07570
C	*****	STD07580
C	*****	STD07590
C	*****	STD07600
C	*****	STD07610
C	*****	STD07620
C	*****	STD07630
C	*****	STD07640
C	*****	STD07650
C	*****	STD07660
C	*****	STD07670
C	*****	STD07680
C	*****	STD07690
C	*****	STD07700
C	*****	STD07710
C	*****	STD07720
C	*****	STD07730
C	*****	STD07740
C	*****	STD07750
C	*****	STD07760
C	*****	STD07770
C	*****	STD07780
C	*****	STD07790
C	*****	STD07800
C	*****	STD07810
C	*****	STD07820
C	*****	STD07830
C	*****	STD07840
C	*****	STD07850
C	*****	STD07860
C	*****	STD07870
C	*****	STD07880
C	*****	STD07890
C	*****	STD07900
C	*****	STD07910

APPENDIX B. PCMLL2 SAMPLE INPUT, OUPUT, AND PROGRAM LISTING

The program PMCMLL2 computes the distance to a specified protected contour, given the unattenuated field strength at one kilometer, the conductivity profile, and the frequency. The program also accepts information on the location of the station, and the bearing in which the distance is to be computed, to provide the latitude and longitude of the protected contour point. Additionally, the program also accepts the location of a proposed transmitter site, and computes the distance and bearing of all protected contour points from this location.

The program utilizes the equivalent distance technique, as previously described for paths where the conductivity is nonhomogeneous along a given radial. The subroutine CONTUR already described is used to compute distances to contour points when the conductivity is homogeneous. The subroutine FCCGW, is used in this iterative technique to compute the field strengths at a given distance. The subroutine FCCGW is a very slightly modified version of the program accompanying the 1986 FCC report (See reference 2).

The program uses a simple card deck input which can easily be expanded and modified to accomodate individual needs. The deck should be set up in the following format.

- CM - Indicates that what ever else is on the line will be ignored. Useful for commenting data sets.
- CL - Indicates that the following cards will contain the desired contour levels. Up to 10 different contour levels can be specified in a single execution of the program. Use a value of 999999 to indicate that there are no more contour levels to be read.
- TL - Indicates that the next two cards will contain the latitude and longitude of the transmitter site. The cards will contain:
 - latitude degrees, minutes, seconds
 - longitude degrees, minutes, seconds
- PL - Indicates that the next two cards will contain the latitude and longitude of the proposed site. The cards will contain:

- latitude degrees, minutes, seconds
longitude degrees, minutes, seconds
- FR - Indicates that the next card will contain the frequency in kHz.
- BR - Indicates that the next card will specify the bearing from the transmitter site, the field at one kilometer from the antenna, the conductivity profile along a radial taken in the direction specified from the transmitter site will then be read. The first card will contain:
- bearing in degrees, field strength at one kilometer
- The conductivity will be specified beginning at the transmitter site in the following way:
- conductivity in mS/m, distance to next discontinuity in km
- Cards of this type continue to be read until a value of 999999 is encountered for the distance to the next discontinuity.
- EN - Indicates the end of the data set.

A sample data set is shown in Figure B1., and the resulting output is shown in Figure B2.

CM WCKL 560 KHZ, CATSKILL, NY DAYTIME PATTERN
 CM 3 ELEMENT DIRECTIONAL ARRAY
 CM
 CM ARRAY INPUT POWER : 1.00 KW
 CM
 CM K FACTOR = 316.568604
 CM THEORETICAL ERMS = 316.679199 MV/M @ 1KM (INCLUDING LOSS)
 TL
 42.0, 12.0, 0.0
 73.0, 50.0, 7.0
 PL
 41.0, 12.0, 24.0
 77.0, 2.0, 46.0
 FR
 560.0
 CL
 0.5
 999999
 BR
 0.00, 501.528809
 4.0, 999999
 BR
 10.00, 419.412598
 4.0, 999999
 BR
 20.00, 329.198975
 4.0, 999999
 BR
 30.00, 237.846268
 4.0, 999999
 BR
 40.00, 152.973434
 4.0, 999999
 BR
 50.00, 81.658371
 4.0, 999999
 BR
 60.00, 29.134094
 4.0, 999999
 BR
 70.00, 2.208762
 4.0, 999999
 BR
 80.00, 13.194660
 4.0, 999999
 BR
 90.00, 7.506421
 4.0, 999999
 BR
 100.00, 9.244386
 4.0, 999999
 BR
 110.00, 30.699310
 4.0, 999999
 BR
 120.00, 50.873703
 4.0, 999999
 BR
 130.00, 65.001801
 4.0, 999999
 BR
 140.00, 70.055801
 4.0, 999999
 BR
 150.00, 65.001801
 4.0, 999999

BR
 160.00, 50.873703
 4.0, 999999
 BR
 170.00, 30.699310
 4.0, 999999
 BR
 180.00, 9.244386
 4.0, 999999
 BR
 190.00, 7.506421
 4.0, 999999
 BR
 200.00, 13.194660
 4.0, 999999
 BR
 210.00, 2.208762
 4.0, 999999
 BR
 220.00, 29.134094
 4.0, 999999
 BR
 230.00, 81.658371
 4.0, 999999
 BR
 240.00, 152.973434
 4.0, 999999
 BR
 250.00, 237.846268
 4.0, 999999
 BR
 260.00, 329.198975
 4.0, 999999
 BR
 270.00, 419.412598
 4.0, 999999
 BR
 280.00, 501.528809
 4.0, 999999
 BR
 290.00, 570.036133
 4.0, 999999
 BR
 300.00, 621.126465
 4.0, 999999
 BR
 310.00, 652.523926
 4.0, 999999
 BR
 320.00, 663.099609
 4.0, 999999
 BR
 330.00, 652.523926
 4.0, 999999
 BR
 340.00, 621.126465
 4.0, 999999
 BR
 350.00, 570.036133
 4.0, 999999
 EN

Figure B1. Sample Input Deck

INTERFERING TRANSMITTER LOCATION
 LATITUDE : 42. 50. 000 N
 LONGITUDE : 73. 00. 000 W
 PROPOSED NEW TRANSMITTER LOCATION
 LATITUDE : 41. 12. 24.000 N
 LONGITUDE : 77. 00. 000 W
 FREQUENCY : 560.0 KHZ

PROTECTED CONTOUR LEVEL : 0.500000 MV/M

HEARING	FIELD @ KM	RADIUS	LATITUDE	LONGITUDE	DIST	HEARING
0.0	501.53	120.472	17. 11. 33.00	50. 24. 15.00	351.196	44.4895
10.0	419.31	112.481	11. 3. 34.04	35. 24. 27.61	360.298	51.289
20.0	329.30	101.820	53. 42. 30.20	17. 15. 45.71	363.318	56.760
30.0	237.85	88.690	10. 14. 49.00	20. 31. 44.75	364.178	60.158
40.0	152.97	72.798	12. 13. 26.32	38. 49. 59.53	374.720	61.316
50.0	81.60	51.551	11. 10. 58.83	42. 29. 32.05	389.991	66.029
60.0	29.13	29.065	0. 0. 40.55	24. 24. 47.21	397.512	67.584
70.0	13.14	15.384	50. 39. 33.25	39. 50. 18.73	403.741	69.244
80.0	7.51	8.884	10. 0. 40.55	42. 29. 32.05	497.564	69.745
90.0	4.70	5.117	50. 39. 33.25	39. 50. 18.73	511.745	72.459
100.0	3.00	3.100	10. 0. 40.55	42. 24. 47.21	513.929	74.719
120.0	1.70	1.717	50. 39. 33.25	39. 50. 18.73	517.821	76.477
140.0	0.97	0.919	10. 0. 40.55	42. 24. 47.21	527.245	77.564
160.0	0.60	0.600	50. 39. 33.25	39. 50. 18.73	530.443	78.847
180.0	0.37	0.370	10. 0. 40.55	42. 24. 47.21	540.095	79.604
200.0	0.24	0.240	50. 39. 33.25	39. 50. 18.73	548.131	79.913
220.0	0.14	0.140	10. 0. 40.55	42. 24. 47.21	557.448	80.732
240.0	0.09	0.090	50. 39. 33.25	39. 50. 18.73	567.065	81.433
260.0	0.07	0.070	10. 0. 40.55	42. 24. 47.21	576.981	82.121
280.0	0.05	0.050	50. 39. 33.25	39. 50. 18.73	587.194	82.751
300.0	0.04	0.040	10. 0. 40.55	42. 24. 47.21	597.705	83.322
320.0	0.03	0.030	50. 39. 33.25	39. 50. 18.73	608.512	83.833
340.0	0.02	0.020	10. 0. 40.55	42. 24. 47.21	619.612	84.284
360.0	0.01	0.010	50. 39. 33.25	39. 50. 18.73	630.902	84.674
380.0	0.01	0.010	10. 0. 40.55	42. 24. 47.21	642.378	85.004
400.0	0.01	0.010	50. 39. 33.25	39. 50. 18.73	653.936	85.274
420.0	0.01	0.010	10. 0. 40.55	42. 24. 47.21	665.571	85.484
440.0	0.01	0.010	50. 39. 33.25	39. 50. 18.73	677.278	85.634
460.0	0.01	0.010	10. 0. 40.55	42. 24. 47.21	689.051	85.724

Figure B2. Sample Output

```

C*****
C AM BROADCAST FIXED PATH FIELD STRENGTH CONTOUR PROGRAM
C THIS PROGRAM COMPUTES THE DISTANCE TO A SPECIFIED FIELD-STRENGTH
C CONTOUR OVER PATHS WITH VARYING CONDUCTIVITY. IT USES A NEWTONS
C METHOD APPROXIMATION TO SOLVE THE INVERSE PROPAGATION PROBLEM.
C THE GROUND WAVE FIELD STRENGTH VALUES ARE COMPUTED USING THE FCC
C EQUIVALENT DISTANCE TECHNIQUE. THE LATITUDE AND LONGITUDE OF THE
C FIELD STRENGTH POINT ARE PROVIDED.
C
C IMPLICIT REAL*8 ( A-H, O-Z )
C
C COMMON / SIGP0 / DISTA(36,10), DISTR(36,10), SIGMA(36,10)
C COMMON / PATRN / HFT(36), FIELD(36), FBR
C COMMON / TL0C / TLAT0G, TLAT0N, TLAT0S, TLNG0G, TLNG0N, TLNG0S,
C & FREQ
C COMMON / CLEVEL / CL(10), NCL
C COMMON / RADIS / RAD(36,10)
C COMMON / CONST / PI, DTR, SMTKM, SMTNM
C COMMON / INOUT / IN, IO
C
C DEFINE INPUT AND OUTPUT DEVICES
C
C   IN = 5
C   IO = 6
C
C INITIALIZE CONSTANTS
C
C   PI = 3.14159 26535 89793 23846 26433
C   DTR = PI / 180.00
C   SMTKM = 1.609344
C   SMTNM = 1. / 1.151626418
C
C READ NECESSARY PARAMETERS IN INPUT SUBROUTINE
C
C CALL INPUT
C
C LOOP OVER CONTOUR LEVELS
C   DO 100, II = 1, NCL - 1
C
C LOOP OVER VARIOUS BEARINGS
C   DO 90, JJ = 1, NBR
C
C     D = 0.0
C     I = 1
C
C IF PROPAGATING OVER WATER, CHANGE PERMITTIVITY TO 80.0
C 40 IF ( SIGMA(JJ,I) .EQ. 5000.0 ) THEN
C     EPSILN = 80.0
C ELSE
C     EPSILN = 15.0
C ENDIF
C
C COMPUTE DISTANCE TO CL(II) CONTOUR USING SIGMA(JJ,I)
C   CALL CONTUR( FIELD(JJ), SIGMA(JJ,I), EPSILN, FREQ, DIST, CL(II) )
C
C IF THE DISTANCE TO THE SPECIFIED CONTOUR IS GREATER THAN THE
C DISTANCE TO THE DISCONTINUITY IN CONDUCTIVITY, THEN COMPUTE
C THE FIELD STRENGTH AT THE DISCONTINUITY.
C
C   IF ( DIST .GT. DISTA(JJ,I) ) THEN
C
C     WRITE (IO,*) 'CONTOUR LEVEL = ', CL(II)
C     WRITE (IO,*) 'DISTANCE TO CONTOUR = ', DIST/1.609344
C     WRITE (IO,*) 'USING SIGMA = ', SIGMA(JJ,I)
C     WRITE (IO,*) 'DISCONTINUITY AT ', DISTA(JJ,I)/1.609344
C
C     CALL FCCGW( FIELD(JJ), SIGMA(JJ,I), EPSILN, FREQ,
C & DISTA(JJ,I), FS1, 0 )
C
C     WRITE (IO,*) 'FS AT DISTA USING ABOVE SIGMA = ', FS1
C
C     I = I + 1
C
C IF PROPAGATING OVER WATER CHANGE PERMITTIVITY TO 80.0

```

```

PCN00010
PCY00020
PCN00030
PCN00040
PCN00050
PCN00060
PCN00070
PCN00080
PCN00090
PCN00100
PCN00110
PCN00120
PCN00130
PCN00140
PCN00150
PCN00160
PCN00170
PCN00180
PCN00190
PCN00200
PCN00210
PCN00220
PCN00230
PCN00240
PCN00250
PCN00260
PCN00270
PCN00280
PCN00290
PCN00300
PCN00310
PCN00320
PCN00330
PCN00340
PCN00350
PCN00360
PCN00370
PCN00380
PCN00390
PCN00400
PCN00410
PCN00420
PCN00430
PCN00440
PCN00450
PCN00460
PCN00470
PCN00480
PCN00490
PCN00500
PCN00510
PCN00520
PCN00530
PCN00540
PCN00550
PCN00560
PCN00570
PCN00580
PCN00590
PCN00600
PCN00610
PCN00620
PCN00630
PCN00640
PCN00650
PCN00660
PCN00670
PCN00680
PCN00690
PCN00700
PCN00710
PCN00720
PCN00730
PCN00740
PCN00750
PCN00760
PCN00770
PCN00780
PCN00790
PCN00800

```

```

C      IF ( SIGMA(JJ,I) .EQ. 5000.0 ) THEN
C          EPSILN = 60.0
C      ELSE
C          EPSILN = 15.0
C      ENDIF

C      COMPUTE THE EFFECTIVE DISTANCE TO THE FIELD COMPUTED AT THE
C      DISCONTINUITY IN THE CONDUCTIVITY
C      WRITE (IO,*) 'FOR SIGMA = ',SIGMA(JJ,I)
C      CALL CONTOUR( FIELD(JJ), SIGMA(JJ,I), EPSILN, FREQ, DIST, FS1 )
C      NOTE THE DIFFERENCE BETWEEN EFFECTIVE AND ACTUAL DISTANCES
C      ADJUST DISTA VALUE AND D TERM ACCORDINGLY
C      D = D + ( DISTA(JJ,I-1) - DIST )
C      DISTA(JJ,I) = DISTA(JJ,I) - D
C      WRITE (IO,*) 'EFFECTIVE DISTANCE TO ABOVE FS = ',DIST/1.609344
C      WRITE (IO,*) 'D FACTOR = ',D/1.609344
C      WRITE (IO,*)
C      GOTO 40
C      ELSE
C      ADJUST DISTANCE BY D TERM
C      WRITE ( IO, * ) 'DISTANCE TO FS CONTOUR = ',DIST/1.609
C      WRITE ( IO, * ) 'USING SIGMA = ',SIGMA(JJ,I)
C      DIST = DIST + D
C      WRITE (IO,*) 'DISTANCE AFTER ADJUSTMENT = ',DIST/1.609
C      RAD(JJ,I) = DIST
C      ENDIF
C      RESTORE DATA TO DISTA ARRAY FOR NEXT CONTOUR LEVEL
C      DO 80 K = 1, 10
C      80  DISTA(JJ,K) = DISTB(JJ,K)
C      90  CONTINUE
C      100 CONTINUE
C      PROVIDE TEXT AND FILE OUTPUT OF DATA
C      CALL OUTPUT
C      END
C      *****
C      INPUT - THIS SUBROUTINE READS THE CARDS CONTAINING THE INFORMATION
C      REQUIRED FOR PCLL1.
C      SUBROUTINE INPUT
C      IMPLICIT REAL*8 ( A-H, O-Z )
C      CHARACTER * 2 CAPD
C      COMMON / SIGPRO / DISTA (36,10), DISTB (36,10), SIGMA(36,10)
C      COMMON / PATRN / HFT(36), FIELD(36), JHR
C      COMMON / TLOC / TLATDG, TLATMN, TLATSC, TLNGDG, TLNGMN, TLNGSC,
C      & FREQ
C      COMMON / PLUC / PLATDG, PLATMN, PLATSC, PLNGDG, PLNGMN, PLNGSC
C      COMMON / CLEVEL / CL(10), NCL
C      COMMON / INOUT / IN, IO
C      INITIALIZE VARIABLES
C      NBR = 0
C      NCL = 0
C      INITIALIZE ARRAYS
C      DO 5, J = 1, 36

```

```

PCM00810
PCM00820
PCM00830
PCM00840
PCM00850
PCM00860
PCM00870
PCM00880
PCM00890
PCM00900
PCM00910
PCM00920
PCM00930
PCM00940
PCM00950
PCM00960
PCM00970
PCM00980
PCM00990
PCM01000
PCM01010
PCM01020
PCM01030
PCM01040
PCM01050
PCM01060
PCM01070
PCM01080
PCM01090
PCM01100
PCM01110
PCM01120
PCM01130
PCM01140
PCM01150
PCM01160
PCM01170
PCM01180
PCM01190
PCM01200
PCM01210
PCM01220
PCM01230
PCM01240
PCM01250
PCM01260
PCM01270
PCM01280
PCM01290
PCM01300
PCM01310
PCM01320
PCM01330
PCM01340
PCM01350
PCM01360
PCM01370
PCM01380
PCM01390
PCM01400
PCM01410
PCM01420
PCM01430
PCM01440
PCM01450
PCM01460
PCM01470
PCM01480
PCM01490
PCM01500
PCM01510
PCM01520
PCM01530
PCM01540
PCM01550
PCM01560
PCM01570
PCM01580
PCM01590
PCM01600

```

```

      BFT(J) = 0.0
      FIELD(J) = 0.0
      DO 4, I = 1, 10
        DISTA(J,I) = 0.0
        DISTB(J,I) = 0.0
        SIGMA(J,I) = 0.0
4      CONTINUE
5      CONTINUE
C 10    READ ( IN, 1000 ) CARD
      IF ( CAPD .EQ. 'CM' ) GOTO 10
C
      IF ( CARD .EQ. 'CL' ) THEN
20      NCL = NCL + 1
        READ ( IN, * ) CL(NCL)
        IF ( CL(NCL) .EQ. 999999 ) THEN
          GOTO 10
        ELSE
          GOTO 20
        ENDIF
      ENDIF
C
      IF ( CAPD .EQ. 'TL' ) THEN
        READ ( IN, * ) TLATDG, TLATMH, TLATSC
        READ ( IN, * ) TLNGDG, TLNGMH, TLNGSC
      ENDIF
C
      IF ( CARD .EQ. 'PL' ) THEN
        READ ( IN, * ) PLATDG, PLATMH, PLATSC
        READ ( IN, * ) PLNGDG, PLNGMH, PLNGSC
      ENDIF
C
      IF ( CAPD .EQ. 'FR' ) THEN
        READ ( IN, * ) FREQ
        FREQ = FREQ / 1000.0
      ENDIF
C
      IF ( CARD .EQ. 'BR' ) THEN
        J = 1
        NBR = NBR + 1
30      READ ( IN, * ) BFT(NBR), FIELD(NBR)
        READ ( IN, * ) SIGMA(NBR,J), DISTA(NBR,J)
        DISTB(NBR,J) = DISTA(NBR,J)
        IF ( DISTA(NBR,J) .EQ. 999999 ) THEN
          GOTO 10
        ELSE
          J = J + 1
          GOTO 30
        ENDIF
      ENDIF
C
      IF ( CAPD .EQ. 'EN' ) THEN
        RETURN
      ELSE
        GOTO 10
      ENDIF
C
1000  FORMAT ( A2 )
C
      END
C*****
C
C  OUTPUT - THIS SUBROUTINE PROVIDES THE TEXT AND FILE OUTPUT FOR
C  PCMLL1.
C
C  SUBROUTINE OUTPUT
C
C  IMPLICIT REAL*8 ( A-H, O-Z )
C
C  COMMON / SIGPRO / DISTA(36,10), DISTB(36,10), SIGMA(36,10)
C  COMMON / PATRN / BFT(36), FIELD(36), NBR
C  COMMON / TLOC / TLATDG, TLATMH, TLATSC, TLNGDG, TLNGMH, TLNGSC,
C  & FREQ
C  COMMON / PLOC / PLATDG, PLATMH, PLATSC, PLNGDG, PLNGMH, PLNGSC
C  COMMON / CLEVEL / CL(10), NCL
C  COMMON / RADIUS / RAD(36,10)
C  COMMON / CONST / PI, DTR, SMTKM, SMTMH
C  COMMON / INOUT / IN, IO

```

```

PC*01610
PC*01620
PC*01630
PC*01640
PC*01650
PC*01660
PC*01670
PC*01680
PC*01690
PC*01700
PC*01710
PC*01720
PC*01730
PC*01740
PC*01750
PC*01760
PC*01770
PC*01780
PC*01790
PC*01800
PC*01810
PC*01820
PC*01830
PC*01840
PC*01850
PC*01860
PC*01870
PC*01880
PC*01890
PC*01900
PC*01910
PC*01920
PC*01930
PC*01940
PC*01950
PC*01960
PC*01970
PC*01980
PC*01990
PC*02000
PC*02010
PC*02020
PC*02030
PC*02040
PC*02050
PC*02060
PC*02070
PC*02080
PC*02090
PC*02100
PC*02110
PC*02120
PC*02130
PC*02140
PC*02150
PC*02160
PC*02170
PC*02180
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C WRITE TRANSMITTER LOCATION, FREQUENCY
C
1000 WRITE ( IO, 1000 ) TLATDG, TLATMN, TLATSC, TLNGDG, TLNGMN, TLNGSC
1000 FORMAT ( //, 16X, 'INTERFERING TRANSMITTER LOCATION', /
& 'LATITUDE : ', 5X, F5.0, 5X, F3.0, 5X, F6.3, 2X, 'N', /
& 'LONGITUDE : ', 5X, F5.0, 5X, F3.0, 5X, F6.3, 2X, 'W', /
1005 WRITE ( IO, 1005 ) PLATDG, PLATMN, PLATSC, PLNGDG, PLNGMN, PLNGSC
1005 FORMAT ( //, 16X, 'PROPOSED NEW TRANSMITTER LOCATION', /
& 'LATITUDE : ', 5X, F5.0, 5X, F3.0, 5X, F6.3, 2X, 'N', /
& 'LONGITUDE : ', 5X, F5.0, 5X, F3.0, 5X, F6.3, 2X, 'W', /
1010 WRITE ( IO, 1010 ) FREQ * 1000.0
1010 FORMAT ( //, 'FREQUENCY : ', 2X, F6.1, 2X, 'KHZ' )
C
C CONVERT LATITUDE AND LONGITUDE TO DECIMAL EQUIVALENT
C
TLAT = TLATDG + TLATMN / 60.0 + TLATSC / 3600.0
TLNG = TLNGDG + TLNGMN / 60.0 + TLNGSC / 3600.0
C
PLAT = PLATDG + PLATMN / 60.0 + PLATSC / 3600.0
PLNG = PLNGDG + PLNGMN / 60.0 + PLNGSC / 3600.0
C
C LOOP OVER CONTOUR LEVELS
C
DO 100, II = 1, NCL - 1
C
C WRITE OUT CONTOUR LEVEL
C
1020 WRITE ( IO, 1020 )
1020 FORMAT ( //, 80 ( '*' ), / )
1030 WRITE ( IO, 1030 ) CL(II)
1030 FORMAT ( ' PROTECTED CONTOUR LEVEL : ', F12.6, 2X, 'MV/M' )
1040 WRITE ( IO, 1040 )
1040 FORMAT ( //, ' BEARING', 5X, 'FIELD 2 KM', 5X, 'RADIUS', 9X,
& 'LATITUDE', 14X, 'LONGITUDE',
& '10X, 'DIST', 10X, 'BEARING' )
C
C LOOP OVER BEARINGS
C
DO 90, JJ = 1, NBR
C
C COMPUTE LATITUDE AND LONGITUDE OF CONTOUR POINT
C
B1 = 90.0 - TLAT
R1 = RAD(JJ,II) / SMTKM * SMTNM / 60.0
C
A1 = ACOS ( COS( DTR * B1 ) * COS( DTR * R1 ) + SIN( DTR * B1 ) *
& SIN( DTR * R1 ) * COS( DTR * BFT(JJ) ) ) / DTR
C
CLAT = 90.0 - A1
CLNG = - ASIN( SIN( DTR * R1 ) *
& SIN( DTR * BFT(JJ) ) / SIN( DTR * A1 ) ) / DTR + TLNG
C
C COMPUTE BEARING AND DISTANCE OF CONTOUR POINT FROM PROPOSED SITE
C
G = CLNG - PLNG
A2 = 90.0 - PLAT
B2 = 90.0 - CLAT
C2 = ACOS ( COS( DTR * A2 ) * COS( DTR * B2 ) +
& SIN( DTR * A2 ) * SIN( DTR * B2 ) * COS( DTR * G ) ) / DTR
C
RPLPC = C2 * 60.0 * SMTKM / SMTNM
THETA = - ASIN( SIN( DTR * G ) * SIN( DTR * A2 ) /
& SIN( DTR * C2 ) ) / DTR
H = SIGN( 1.00, THETA )
V = SIGN( 1.00, (CLAT-PLAT) )
IF ( U.GT. 0 ) THEN
IF ( V.GT. 0 ) THEN
RPLPC = THETA
ELSE
RPLPC = 180.0 - THETA
ENDIF
ELSE
IF ( V.GT. 0 ) THEN
RPLPC = THETA + 360.0
ELSE
RPLPC = 180.0 - THETA
ENDIF
ENDIF
C
C CONVERT TO DEGREES, MINUTES, SECONDS

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C      CLATDG = AINT ( CLAT )
C      CLATMN = AINT ( ( CLAT - CLATDG ) * 60.0 )
C      CLATSC = ( CLAT - CLATDG - CLATMN / 60.0 ) * 3600.0
C
C      CLNGDG = AINT ( CLNG )
C      CLNGMN = AINT ( ( CLNG - CLNGDG ) * 60.0 )
C      CLNGSC = ( CLNG - CLNGDG - CLNGMN / 60.0 ) * 3600.0
C
C      WRITE OUT DATA FOR EACH BEARING
C
C      WRITE ( 10, 1050 ) BFT(JJ), FIELD(JJ), PAD(JJ,11),
C      & CLATDG, CLATMN, CLATSC, CLNGDG, CLNGMN, CLNGSC,
C      & RPLPC, RPLPC
C
C 1050 FORMAT ( 3X, F5.1, 5X, F7.2, 5X, F8.3,
C      & 2 ( 5X, F4.0, 3X, F3.0, 3X, F5.2 ),
C      & 5X, F8.3, 5X, F8.3 )
C
C 90 CONTINUE
C 100 CONTINUE
C
C      END
C
C *****
C
C      CONTOUR - COMPUTES THE DISTANCE TO A SPECIFIED FIELD STRENGTH CONTOUR
C      USING AN ITERATIVE NEWTON TECHNIQUE WITH THE FCC GROUND WAVE
C      FIELD STRENGTH PROGRAM.
C
C      SUBROUTINE CONTOUR ( FIELD0, SIGMA, EPSILN, FREQ, DIST, FS )
C
C      IMPLICIT REAL*8 ( A-H, O-Z )
C
C      INITIALIZE COUNTER, AND SOLUTION TYPE FOR FCCGW
C      IST = 0 - NORMAL SOLUTION USING MOST EFFICIENT METHOD
C      IST = 1 - RESIDUE SUM SOLUTION TO GIVE ACCURACY IN BULGE REGION
C
C      I = 0
C      IST = 0
C
C      SET INITIAL GUESS FOR DISTANCE TO DESIRED CONTOUR.
C
C      DIST = 10
C
C      EVALUATE FIELD STRENGTH AT ASSUMED DISTANCE, AND A NEARBY POINT
C      SO THE SLOPE CAN BE COMPUTED.
C
C 10 CALL FCCGW ( FIELD0, SIGMA, EPSILN, FREQ, DIST, FS1, IST )
C      DIST1 = DIST + DIST * 0.00001
C      CALL FCCGW ( FIELD0, SIGMA, EPSILN, FREQ, DIST1, FS2, IST )
C
C      CHECK TO SEE IF THE FIELD STRENGTH VALUE IS WITHIN TOLERANCE.
C      IF IT IS, THEN RETURN TO CALLING PROGRAM WITH DISTANCE.
C      IF IT IS NOT, THEN COMPUTE NEW GUESS FOR DISTANCE.
C
C      I = I + 1
C
C      IF ( ABS ( FS1 - FS ) .GT. 1.E-9 ) THEN
C      DISTN = DIST - ( FS1 - FS ) * ( DIST * 0.00001 ) / ( FS2 - FS1 )
C
C      WRITE ( 6, * ) 1, FS1-FS, DISTN
C
C      AVOID CONVERGENCE PROBLEM WHEN DISTANCE IS CLOSE TO THE
C      Crossover POINT FROM CORRECTION TERM SOLUTION TO RESIDUE SUM
C      SOLUTION.
C
C      IF ( INT( DISTN * 1E8 ) .EQ. INT( DIST * 1E8 ) ) IST = 1
C      IST = 1
C
C      IF ( ( I / 3. ) .EQ. INT( I / 3. ) ) DISTN2 = DISTN
C
C      WE MUST AVOID GUESSING DISTANCES THAT ARE LESS THAN OR EQUAL TO ZERO.
C
C      IF ( DISTN .LE. 0.0 ) THEN
C      DIST = DIST / 2.0
C      ELSE
C      DIST = DISTN
C      ENDIF
C
C

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      GOTO 10
      ENDIF
      WRITE ( 6, * ) I, PSI-PS, DIST
      RETURN
      END
*****
FCCGW - CALCULATED THE GROUND-WAVE FIELD STRENGTH FOR VERTICALLY
POLARIZED MF WAVES.
      SUBROUTINE FCCGW ( FIELD0, SIGMA, EPSILN, FREQ, DIST, FIELD, IST )
      IMPLICIT REAL*8 ( A-H, O-Z )
      REAL*8 K
      DATA PI / 3.1415927 /
      DEFINE INPUT AND OUTPUT UNITS.
      COMMON / INOUT / IN, IO
      EVALUATE CONSTANTS REQUIRED BY MAJOR SUBROUTINES.
      CALL GWCHST(
      & SIGMA, EPSILN, FREQ, DIST,
      & P, B, K, CHI )
      SET DISTANCE BEYOND WHICH TO USE RESIDUE SERIES.
      FAR = 80. / FREQ ** ( 1 / 3.0 )
      CALCULATE GROUND-WAVE ATTENUATION BY APPROPRIATE METHOD.
      IST ALLOWS SOLUTION TO BE FORCED TO RESIDUE SERIES METHOD TO
      PREVENT CONVERGENCE PROBLEMS IN SUBROUTINE CONTOUR.
      IF ( ( DIST .LE. FAR ) .AND. ( IST .EQ. 0 ) ) THEN
        CALL SURFAC ( P, B, K, A )
      ELSE
        PSI = 0.5 * B
        CALL RESIDU ( CHI, K, PSI, A )
      ENDIF
      MULTIPLY INVERSE-DISTANCE FIELD THE ATTENUATION.
      (160.93440 MV/M AT 1 KM = 100 MV/M AT 1 MILE)
      FIELD = A * FIELD0 / DIST
      RETURN
      END
*****
GWCHST - SET GROUNDWAVE CONSTANTS. THE INDEPENDENT VARIABLES
PASSED AS ARGUMENTS OF THIS SUBROUTINE ARE:
      SIGMA, THE GROUND CONDUCTIVITY IN MILLISIEMENS/METER
      EPSILN, THE RELATIVE DIELECTRIC CONSTANT (1.0 FOR AIR)
      FREQ, THE FREQUENCY IN MHZ
      DIST, RADIO PATH LENGTH IN KILOMETERS
      QUANTITIES CALCULATED IN THIS SUBROUTINE AND RETURNED ARE:
      P, AMPLITUDE OF THE COMPLEX NUMERICAL DISTANCE INTRODUCED
      FOR THE SOLUTION OF RADIO PROPAGATION PROBLEMS BY ARNOLD
      SOMMERFELD IN 1909.
      B, PHASE ANGLE OF THE NUMERICAL DISTANCE
      K, A DIMENSIONLESS PARAMETER PROPORTIONAL TO THE CUBE-
      ROOT OF THE RATIO OF WAVELENGTH TO THE EFFECTIVE EARTH
      RADIUS, AND DEPENDENT ALSO UPON THE GROUND CONSTANTS
      CHI, A DIMENSIONLESS PARAMETER PROPORTIONAL TO THE CUBE-
      ROOT OF THE EFFECTIVE EARTH RADIUS MEASURED IN WAVELENGTHS,

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C      AND PROPORTIONAL ALSO TO THE RADIO PATH DISTANCE MEASURED
C      AS THE ANGLE SUBTENDED FROM THE CENTER OF THE EARTH.
C
C      USE OF THE SYMBOLS P AND B FOR THE AMPLITUDE AND PHASE OF
C      THE NUMERICAL DISTANCE FOLLOWS NORTON, PROC. IRE 1941.
C
C      THE SYMBOL K IS USED TO DENOTE EXACTLY THE SAME QUANTITY IN
C      NBS TECH NOTE 101, IN NORTON'S 1941 IRE PAPER AND IN THE
C      1949 BOOK BY BREMMER. SEE REFERENCE BELOW.
C
C      CHI IS USED IN EVALUATING THE RESIDUE SERIES. THE GREEK LETTER
C      OF THAT NAME WAS USED FOR THIS QUANTITY BY BREMMER IN HIS 1949
C      BOOK.
C
C      REFERENCES:
C
C      Y. A. NORTON, THE CALCULATION OF GROUND-WAVE FIELD
C      INTENSITY OVER A FINITELY CONDUCTING SPHERICAL EARTH,
C      PROC. IRE, DEC 1941, PAGES 623-639.
C
C      H. BREMMER, TERRESTRIAL RADIO WAVES, ELSEVIER
C      PUBLISHING CO., 1949.
C
C      *****
C
C      SUBROUTINE GACST(
C      & SIGMA, EPSILN, FREQ, DIST,
C      & P, R, K, CHI )
C
C      IMPLICIT REAL*8 ( A-H, O-Z )
C      REAL*8 K
C
C      DATA PI / 3.1415927 /
C      RADIUS OF EARTH IN KILOMETERS.
C      DATA EARTH / 6370. /
C      ASSUMED EFFECTIVE EARTH RADIUS FACTOR.
C      DATA FACTOR / 1.3333333333333333 /
C
C      SPEED OF LIGHT IN AIR (REFRACTIVE INDEX 1.00031) KM/SEC.
C      DATA SPEED / 299700. /
C
C      BEGIN EXECUTION. DETERMINE EFFECTIVE EARTH RADIUS FROM
C      GIVEN EARTH RADIUS FACTOR.
C
C      EFFRAD = FACTOR * EARTH
C
C      FIND WAVELENGTH. DISTANCE AND THE EFFECTIVE EARTH RADIUS WILL
C      BE DIVIDED BY WAVELENGTH TO PRODUCE DIMENSIONLESS QUANTITIES.
C
C      WAVLGT = SPEED / ( 1E6 * FREQ )
C
C      INTERMEDIATE VARIABLES X, B1 AND B2 ARE DERIVED FROM THE
C      GROUND CONSTANTS BY FORMULAS THAT APPEAR IN NORTON, PROC.
C      IRE, 1941.
C
C      X = 17.97 * SIGMA / FREQ
C      R1 = ATAN2( EPSILN - 1, X )
C      R2 = DATAN2( EPSILN, X )
C
C      CALCULATION OF NUMERICAL DISTANCE, P, AND ITS PHASE ANGLE, R
C
C      P = PI * ( DIST / WAVLGT ) * COS( R2 ) ** 2
C      & / ( X * COS( R1 ) )
C      R = 2 * R2 - R1
C
C      CALCULATION OF K. SEE NORTON, PROC. IRE, DEC 1941, PAGE 628.
C
C      K = ( WAVLGT /
C      & ( 2 * PI * EFFRAD ) ) ** ( 1./3. )
C      & * SORT( X * COS( R1 ) ) / COS( R2 )
C
C      CALCULATION OF CHI. CONF. BREMMER, TERRESTRIAL RADIO WAVES,
C      PAGE 40, EQUATION (III, 31).
C
C      CHI = DIST / EFFRAD *
C      & ( 2 * PI * EFFRAD / WAVLGT ) ** ( 1./3. )
C
C      RETURN
C      END

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C *****
C SUPFAC - CALCULATION OF SURFACE WAVE ATTENUATION. THE
C FLAT EARTH VALUE IS FOUND BY THE USUAL FORMULA DUE TO A.
C SOMMERFELD. CORRECTIONS FOR CURVED EARTH ARE THEN APPLIED.
C THE CURVED EARTH CORRECTIONS ARE FROM H. BRENNER, APPLICATIONS
C OF THE OPERATIONAL CALCULUS TO GROUND-WAVE PROPAGATION,
C PARTICULARLY FOR LONG WAVES, IER TRANSACTIONS ON ANTENNAS
C AND PROPAGATION, JULY 1958.
C
C INPUTS ARE THE NUMERICAL DISTANCE, P, ITS PHASE, R, AND
C THE PARAMETER K. THE PARAMETER K (SO DENOTED BY HORTON
C AND IN NBS TECH NOTE 101) CARRIES INFORMATION CONCERNING
C THE EFFECTIVE EARTH RADIUS SO THAT SPHERICAL EARTH
C CORRECTIONS CAN BE APPLIED.
C
C OUTPUT IS THE ATTENUATION FACTOR, A, THE MAGNITUDE OF THE
C RATIO OF THE GROUND-WAVE FIELD TO THE FIELD PRODUCED BY
C THE SAME ANTENNA OVER A PERFECTLY CONDUCTING FLAT EARTH.
C *****
C SUBROUTINE SUPFAC( P, R, K, A )
C
C   IMPLICIT REAL*8 ( A-Z, O-Y ),
C     COMPLEX*16 ( Z )
C   COMPLEX*16 DEL, RHO, ERFC, SPS1, SPS2
C   REAL*8 K
C
C   DATA PI / 3.1415927 /
C
C   SPHERICAL EARTH CORRECTION FORMULAS. USE WHEN NUMERICAL
C   DISTANCE IS NOT CLOSE TO 0. NOTICE FORMULAS ARE IN CASCADE
C   SO THAT THE LAST AUTOMATICALLY USES THOSE PREVIOUS. THEY
C   ARE WRITTEN THIS WAY SO THAT THEY CAN BE EXAMINED SEPARATELY,
C   BUT DIRECT REFERENCE IN THE PROGRAM IS MADE ONLY TO THE
C   LAST FORMULA.
C
C   ZA DENOTES THE SOMMERFELD FLAT-EARTH ATTENUATION WHICH MUST
C   BE CALCULATED SEPARATELY.
C
C     ZADJ1( DEL, RHO, ZA ) = ZA
C     & + DEL ** 3 *
C     & 1./2 * ( ( 1 + 2 * RHO ) * ZA
C     & - 1 - ( 0,1 ) * SORT( PI * RHO ) )
C
C     ZADJ2( DEL, RHO, ZA ) = ZADJ1( DEL, RHO, ZA )
C     & + DEL ** 6 *
C     & ( ( 1./2 * RHO ** 2 - 1 ) * ZA
C     & + ( 0,1 ) * SORT( PI * RHO ) * ( 1 - RHO )
C     & + 1 - 2 * RHO + 5./6 * RHO ** 2 )
C
C   BEGIN EXECUTION. CONVERT INPUT VARIABLES P, R TO COMPLEX FORM.
C   THE RESULTING COMPLEX VARIABLE, RHO, WILL ALWAYS BE IN THE
C   UPPER HALF-PLANE.
C
C     RHO = P * ( COS( B ) + ( 0,1 ) * SIN( B ) )
C
C   THE COMPLEX PARAMETER DEL IS DETERMINED BY K AND THE ANGLE R.
C
C     DEL = K *
C     & ( COS ( 3*PI/4 - R/2 ) + ( 0,1 ) * SIN ( 3*PI/4 - R/2 ) )
C
C   FIND COMPLEX ATTENUATION FOR FLAT EARTH
C
C     CALL SOMMER( RHO, ZA )
C
C   ADJUST FOR SPHERICAL EARTH. THE FUNCTIONS SRS1 AND SRS2 ARE
C   POWER SERIES CORRESPONDING TO ZADJ1 AND ZADJ2 RESPECTIVELY.
C   WHEN RHO IS SMALL AND CONSEQUENTLY ZA IS NEAR UNITY, THE POWER
C   SERIES MUST BE USED. ZADJ1 AND ZADJ2 ARE NOT ACCURATE UNDER
C   THESE CONDITIONS BECAUSE THE FORMULAS INVOLVE THE DIFFERENCE
C   BETWEEN ZA AND A NUMBER NEAR UNITY.
C
C     IF ( ABS( RHO ) .GT. 0.5 ) THEN
C       ZADJ = ZADJ2( DEL, RHO, ZA )
C     ELSE
C       Z1 = ( 0,1 ) * SORT( RHO )
C       Z3 = ( DEL * Z1 ) ** 3

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      ZADJ = ZA - Z3 * ( SRS1( Z1 )
&          - Z3 * ( SRS2( Z1 ) ) )
& ENDIF
C RETURN THE AMPLITUDE
      A = ABS( ZADJ )
      RETURN
& END
C *****
C SOMMER - CALCULATION OF THE SURFACE WAVE ATTENUATION
C FACTOR FOR GIVEN NUMERICAL DISTANCE. NUMERICAL
C DISTANCE IS THE PARAMETER INTRODUCED BY A. SOMMERFELD
C IN 1909 WHEN HE SHOWED HOW TO FIND THE FIELD OF A SHORT
C DIPOLE RADIATING OVER A FINITELY CONDUCTING PLANE EARTH.
C INPUT IS THE NUMERICAL DISTANCE IN COMPLEX FORM, PHO.
C OUTPUT ZA IS THE COMPLEX SURFACE WAVE ATTENUATION FACTOR
C FOR A FLAT EARTH.
C *****
C SUBROUTINE SOMMER( RHO, ZA )
C   IMPLICIT REAL*8 ( A-H, O-P, U, V, Y ),
&   COMPLEX*16 ( S, T, W, Z )
&   COMPLEX*16 ERFC, RHO, RHOROT
C   DATA PI / 3.1415927 /
C   COEFFICIENTS C1, D1 ETC. TO APPROXIMATE W-FUNCTION OF
C   LARGE MODULUS (W. ABRAMOWITZ AND I. STEGUN, HANDBOOK OF
C   MATHEMATICAL FUNCTIONS, NATIONAL BUREAU OF STANDARDS,
C   1964, PAGE 328).
C   DATA C1, C2, C3 / 0.4613135, 0.09999216, 0.002883894 /
C   DATA D1, D2, D3 / 0.1901635, 1.7844927, 5.5253437 /
C   APPROXIMATION FORMULA FOR W-FUNCTION OF LARGE MODULUS.
C   ERROR LESS THAN 2 PARTS IN 1E6 PROVIDED THE ABSOLUTE
C   VALUE OF X EXCEEDS 3.9, OR Y > 3.
C   W( Z ) = ( 0,1 ) * Z *
&   ( C1 / ( Z ** 2 - D1 )
&   + C2 / ( Z ** 2 - D2 )
&   + C3 / ( Z ** 2 - D3 ) )
C   BEGIN EXECUTION. THE NUMERICAL DISTANCE, AS A COMPLEX
C   VARIABLE, SHOULD ALWAYS BE FOUND IN THE UPPER HALF-PLANE.
C   CALCULATE ITS SQUARE ROOT, RHOROT, WHICH WILL BE LOCATED
C   IN THE FIRST QUADRANT.
C   IF ( .IMAG( RHO ) .LT. 0 ) STOP
&   'ERROR: COMPLEX NUMERICAL DISTANCE IN LOWER HALF-PLANE'
&   RHOROT = SQRT( RHO )
C   DETERMINE MOST APPROPRIATE METHOD.
C   IF ( REAL( RHOROT ) .GT. 3.9
&   .OR. .IMAG( RHOROT ) .GT. 3.0 ) GOTO 300
&   IF ( ABS( RHOROT ) .GT. 1 ) GOTO 200
C   POWER SERIES FOR SMALL PHO. FOR ABS( RHO ) = 1 OR LESS.
C   THE I-TH TERM WILL BE LESS THAN 1E-35 IN MAGNITUDE AFTER
C   33 TERMS.
C 100 CONTINUE
&   TERM = 1
&   SUM = 1
&   DO 110 I = 1, 33
&     TERM = -2 * RHO * TERM / ( 2 * I - 1 )
&     SUM = SUM + TERM
&     IF ( ABS( TERM ) .LT. ABS( SUM ) / 1E5 )
&       GO TO 120
&   CONTINUE
C 110 CONTINUE
C 120 ZA = SUM + ( 0,1 ) * SQRT( PI * PHO ) * EXP( -RHO )

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      RETURN
      WHEN RHOFOT IS OUTSIDE THE UNIT CIRCLE BUT BELOW THE HORIZONTAL
      LINE Y = 3.0 AND LEFT OF X = 3.0, CALCULATE THE COMPLEX SURFACE
      ATTENUATION IN TERMS OF THE COMPLEMENTARY ERROR FUNCTION
      THE LATTER IS TO BE EVALUATED AT  $-(0.1) * \text{RHOFOT}$ , AND THE SERIES
      OF SALZER PROVIDES A CLOSE APPROXIMATION.
200  CONTINUE
      ERFC = SALZER(  $-(0.1) * \text{RHOFOT}$  )
      ZA = 1 + (0.1) * SORT( PI * RHO ) * EXP( -RHO ) * ERFC
      RETURN
      FOR ARGUMENTS WITH LARGE ABSOLUTE MAGNITUDE, COMPUTE THE
      COMPLEX ATTENUATION FUNCTION IN TERMS OF THE W-FUNCTION
      DESCRIBED IN ABRAMOWITZ.
300  CONTINUE
      ZA = 1 + (0.1) * SORT( PI * RHO ) * W( RHOFOT )
      RETURN
      END
      *****
      SALZER - COMPUTE THE COMPLEMENTARY ERROR FUNCTION OF
      THE COMPLEX ARGUMENT, Z, USING THE METHOD DESCRIBED BY
      H. F. SALZER, FORMULAS FOR CALCULATING THE ERROR FUNCTION
      OF A COMPLEX VARIABLE, MATH. TABLES AND OTHER AIDS TO
      COMPUTATION (JOURNAL PUBLISHED BY NATIONAL RESEARCH COUNCIL),
      VOL. V, 1951. THE FORMULAS ALSO APPEAR IN ABRAMOWITZ AND
      STEGUN, PAGE 299.
      *****
      FUNCTION SALZER( Z )
      IMPLICIT REAL*8 ( A-H, O-R, T-Y ),
      & COMPLEX*16 ( S, Z )
      & COMPLEX*16 TEST
      COMMON / INOUT / IN, IO
      DATA PI / 3.1415927 /
      COEFFICIENTS P, AND A1, A1, ETC. TO APPROXIMATE ERROR
      FUNCTION OF REAL ARGUMENT (C. HASTINGS, APPROXIMATIONS
      FOR DIGITAL COMPUTERS, PRINCETON UNIV. PRESS, 1955 )
      DATA P, A1, A2, A3, A4, A5 / 0.3275911, 0.2548296,
      & -0.2844967, 1.4214137, -1.4531520, 1.0614054 /
      APPROXIMATION OF COMPLEMENTARY ERROR FUNCTION FOR REAL,
      NON-NEGATIVE ARGUMENTS.
      T( X ) = 1 / ( 1 + P * X )
      REERFC( X ) =
      & ( A1 + T( X ) *
      & ( A2 + T( X ) *
      & ( A3 + T( X ) *
      & ( A4 + T( X ) *
      & ( A5 ) ) ) ) )
      & * EXP( - X ** 2 )
      FUNCTIONS FOR APPROXIMATING ERF OF COMPLEX ARGUMENTS.
      F( X, Y, N ) = 2 * X
      & - 2 * X * COSH( N * Y ) * COS( 2 * X * Y )
      & + N * SINH( N * Y ) * SIN( 2 * X * Y )
      G( X, Y, N ) =
      & 2 * X * COSH( N * Y ) * SIN( 2 * X * Y )
      & + N * SINH( N * Y ) * COS( 2 * X * Y )
      BEGIN EXECUTION.
      X = REAL( Z )
      Y = IMAG( Z )
      SUM = 0
      IF ( Y .EQ. 0 ) GOTO 20

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PCM07210
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      N = VTN( ABS( R0./Y ), 50.00 )
      DO 10 I = 1, N
        SUM = SUM + EXP( -.25 * ( I ** 2 ) )
        * ( F( X, Y, I ) + ( 0,1 ) * G( X, Y, I ) )
        / ( I ** 2 + 4.0 * X ** 2 )
      &
      IF ( I.GT. 1 .AND.
      & ABS( SUM - TEST ) .LT. ABS( TEST ) / 1E5 )
      & GOTO 20
      TEST = SUM
10    CONTINUE
      WRITE ( IO, * )
      & 'UNEXPECTED ERROR: SALZER SERIES FAILED TO CONVERGE'
C
20    CONTINUE
      SALZER = - 2 * EXP( - X ** 2 ) * SUM / PI
      IF ( X.NE. 0. )
      & SALZER = SALZER - EXP( - X ** 2 ) *
      & ( 1 - COS( 2 * X * Y ) + ( 0,1 ) * SIN( 2 * X * Y ) ) /
      & ( 2 * PI * X )
C
      WHEN CALLED TO AID EVALUATION OF THE SOMMERFELD COMPLEX
      SURFACE WAVE ATTENUATION, THE VARIABLE Z WILL BE IN THE
      4TH QUADRANT. ADDITIONAL PROVISION IS MADE BELOW TO
      RETURN THE VALUE OF THE COMPLEMENTARY ERROR FUNCTION
      INDEPENDENT OF WHAT QUADRANT Z IS IN.
C
      IF ( X.GE. 0. ) THEN
        SALZER = SALZER + REERFC( X )
      ELSE
        SALZER = SALZER + 2 - REERFC( -X )
      ENDIF
      RETURN
C
      END
C
*****
SRS1 - COMPUTE THE FIRST OF TWO POWER SERIES ASSOCIATED
WITH CORRECTION TERMS TO THE COMPLEX SURFACE WAVE
ATTENUATION. AN EFFECTIVE EARTH RADIUS FACTOR IS NEEDED
TO APPLY THE CORRECTION, AND THE APPROPRIATE FACTOR IS
APPLIED TO THE SUM OF THE POWER SERIES ADDRESSED BY THIS
SUBPROGRAM.
C
      THE INPUT IS A SINGLE COMPLEX VALUE, AND THE OUTPUT IS
      ALSO COMPLEX.
C
*****
      FUNCTION SRS1( Z )
C
      IMPLICIT REAL*8 ( A-H, O-R, U-Y ),
      & COMPLEX*16 ( S, T, Z )
      COMPLEX*16 ODD, EVEN
C
      COMMON / INOUT / IN, IO
      DATA PI / 3.1415927 /
C
      ODD = 4 * Z / ( 3 * SORT( PI ) )
      EVEN = 1
      SUM = 1 + 2 * ODD
      DO 100 I = 2, 50
        IF ( MOD( I, 2 ) .EQ. 0 ) THEN
          EVEN = 2 * EVEN * Z ** 2 / ( I + 2 )
          TERM = EVEN
        ELSE
          ODD = 2 * ODD * Z ** 2 / ( I + 2 )
          TERM = ODD
        ENDIF
        SUM = SUM + ( I + 1 ) * TERM
        IF ( I.GT. 2 .AND.
        & ABS( SUM - TEST ) .LT. ABS( TEST ) / 1E6 )
        & GOTO 110
        TEST = SUM
100    CONTINUE
      WRITE ( IO, * ) 'SLOW CONVERGENCE IN SERIES 1'
C
110    SRS1 = SUM * SORT( PI ) / 2
      RETURN
C

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C      FLD
C*****
C      SRS2 - COMPUTE THE SECOND OF TWO POWER SERIES ASSOCIATED
C      WITH CORRECTION TERMS TO THE COMPLEX SURFACE WAVE
C      ATTENUATION. AN EFFECTIVE EARTH RADIUS FACTOR IS NEEDED
C      TO APPLY THE CORRECTION, AND THE APPROPRIATE FACTOR IS
C      APPLIED TO THE SUM OF THE POWER SERIES ADDRESSED BY THIS
C      SUBPROGRAM.
C      THE INPUT IS A SINGLE COMPLEX VALUE, AND THE OUTPUT IS
C      ALSO COMPLEX.
C*****
C      FUNCTION SPS2( Z )
C      IMPLICIT REAL*8 ( A-H, O-R, U-Y ),
C      & COMPLEX*16 ( S, T, Z )
C      COMPLEX*16 ODD, EVEN
C      COMMON / INQUI / IN, IO
C      DATA PI / 3.1415927 /
C      EVEN = 1 / ( 15 * SQRT( PI ) )
C      ODD = Z / 6
C      SUM = 7 * EVEN + 2 * 8 * ODD
C      DO 200 I = 2, 50
C        IF ( MOD( I, 2 ) .EQ. 0 ) THEN
C          EVEN = 2 * EVEN * Z ** 2 / ( I + 5 )
C          TERM = EVEN
C        ELSE
C          ODD = 2 * ODD * Z ** 2 / ( I + 5 )
C          TERM = ODD
C        ENDIF
C        SUM = SUM + ( I + 1 ) * ( I + 7 ) * TERM
C        IF ( I .GT. 2 .AND.
C      & ABS( SUM - TEST ) .LT. ABS( TEST ) / 1E6 )
C      & GOTO 210
C      TEST = SUM
C 200 CONTINUE
C      WRITE ( IO, * ) 'SLOW CONVERGENCE IN SERIES 2'
C 210 SRS2 = SUM * SQRT( PI ) / 8
C      RETURN
C      FLD
C*****
C      RESIDU - CALCULATION OF DIFFRACTION LOSS OVER SMOOTH
C      FINITELY CONDUCTING EARTH USING RESIDUE SERIES.
C      THE COMPUTATIONS FOLLOW THE METHOD OUTLINED BY H. BREMMER,
C      TERRESTRIAL RADIO WAVES, ELSEVIER PUBLISHING CO., 1949.
C*****
C      SUBROUTINE RESIDU( CHI, K, PSI,
C      & ATTENU )
C      IMPLICIT REAL*8 ( A, B, E-H, O-P, R, S, U-Y ),
C      & COMPLEX*16 ( C, D, O, T, Z )
C      REAL*8 K, CHI
C      INTEGER S
C      PARAMETER ( NTERMS = 30 )
C      TAU(S) DENOTES THE POINTS IN THE COMPLEX PLANE AT WHICH
C      RESIDUES OF THE DIFFRACTION FIELD INTEGRAND ARE TO BE
C      EVALUATED.
C      DIMENSION TAU( NTERMS )
C      NO NEED TO RECALCULATE RESIDUE POINTS IF NO CHANGE IN DEL.
C      COMMON / SAVE / DEL, OSOR, TAU, NPTS
C      SAVE / SAVE /
C      NTERMS REFERS TO HOW MANY TERMS MAY BE INCLUDED IN THE

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RESIDUE SERIES. THE NUMBER ACTUALLY INCLUDED WILL BE DETERMINED BY A CONVERGENCE CRITERION REPRESENTED BY THE PARAMETER PRECIS.

DATA PRECIS / 1E-4 /

FINE DETERMINES THE SIZE OF THE THE ELEMENT OF INTEGRATION USED TO LOCATE RESIDUE POINTS.

DATA FINE / 0.05 /
DATA PI / 3.1415927 /

TAU0 AND TAU1 DENOTE REFERENCE POINTS WHERE RESIDUES WOULD BE EVALUATED IN CERTAIN LIMITING CASES. THESE REFERENCE POINTS ARE ALL IN THE FIRST QUADRANT ON THE LINE OF SLOPE 60 DEGREES. THE AMPLITUDES OF THE COMPLEX NUMBERS REPRESENTING REFERENCE POINTS TAU0 AND TAU1 ARE DETERMINED FROM THE AMPLITUDES OF CORRESPONDING ROOTS OF THE AIRY FUNCTION AND ITS DERIVATIVE.

THE REQUIRED ROOTS OF THE AIRY FUNCTION AND ITS DERIVATIVE ARE FOUND BY FUNCTION AIRY0 AND FUNCTION AIRY1 RESPECTIVELY.

FUNCTION PRODUCING TAU0 OR TAU1 FROM THE AMPLITUDE AIRY0 OR AIRY1:

TFN(AIRY) = AIRY / 2 ** (1./3) * EXP((0,1) * PI/3)

FUNCTIONS FOR FINDING POINTS TAU, AT WHICH RESIDUES ARE EVALUATED, FROM THE REFERENCE POINTS TAU0. THESE FORMULAS ARE USED WHEN DEL IS SMALL.

C3 (TA) = -2./3 * TA
C5 (TA) = -4./5 * TA ** 2
C6 (TA) = 14./9 * TA
C7 (TA) = - (5 + 8 * TA ** 3) / 7
C8 (TA) = 58./15 * TA ** 2
C9 (TA) = -TA * (2296./567 + 16/9 * TA ** 3)
C10(TA) = 47./35 + 4656/525 * TA ** 3

TAUFN0(TA, DEL) =
& (TA + DEL *
& (- 1 + DEL *
& (0 + DEL *
& (C3 (TA) + DEL *
& (C5 (TA) + DEL *
& (1./2 + DEL *
& (C6 (TA) + DEL *
& (C7 (TA) + DEL *
& (C8 (TA) + DEL *
& (C9 (TA) + DEL *
& (C10(TA))))))))))

FUNCTIONS FOR FINDING TAU FROM TAU1. USED WHEN DEL IS LARGE AFTER SETTING Q = 1/DEL.

D1(TA) = - 1 / (2 * TA)
D2(TA) = - 1 / (2 * TA ** 3)
D3(TA) = - 1 / (TA ** 2) *
D4(TA) = 1 / 12 + 1 / (16 * TA ** 3))
D5(TA) = - 7./96 + 5 / (128 * TA ** 3))
D6(TA) = - 1 / (TA ** 3) *
D7(TA) = - 21 / (1024 * TA ** 3)))
D8(TA) = - 1 / (TA ** 6) *
D9(TA) = 1 / (TA ** 3) *
D10(TA) = 1 / (TA ** 3) *
D11(TA) = 1 / (TA ** 3) *
D12(TA) = 1 / (TA ** 3) *
D13(TA) = 1 / (TA ** 3) *
D14(TA) = 1 / (TA ** 3) *
D15(TA) = 1 / (TA ** 3) *
D16(TA) = 1 / (TA ** 3) *
D17(TA) = 1 / (TA ** 3) *
D18(TA) = 1 / (TA ** 3) *
D19(TA) = 1 / (TA ** 3) *
D20(TA) = 1 / (TA ** 3) *
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D25(TA) = 1 / (TA ** 3) *
D26(TA) = 1 / (TA ** 3) *
D27(TA) = 1 / (TA ** 3) *
D28(TA) = 1 / (TA ** 3) *
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D30(TA) = 1 / (TA ** 3) *
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D90(TA) = 1 / (TA ** 3) *
D91(TA) = 1 / (TA ** 3) *
D92(TA) = 1 / (TA ** 3) *
D93(TA) = 1 / (TA ** 3) *
D94(TA) = 1 / (TA ** 3) *
D95(TA) = 1 / (TA ** 3) *
D96(TA) = 1 / (TA ** 3) *
D97(TA) = 1 / (TA ** 3) *
D98(TA) = 1 / (TA ** 3) *
D99(TA) = 1 / (TA ** 3) *
D100(TA) = 1 / (TA ** 3) *

TAUFN1(TA, Q) =
& (D1(TA) + Q *
& (D2(TA) + Q *
& (D3(TA) + Q *
& (D4(TA) + Q *
& (D5(TA) + Q *
& (D6(TA) + Q *
& (D7(TA) + Q *
& (D8(TA) + Q *
& (D9(TA) + Q *
& (D10(TA) + Q *
& (D11(TA) + Q *
& (D12(TA) + Q *
& (D13(TA) + Q *
& (D14(TA) + Q *
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& (D16(TA) + Q *
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& (D46(TA) + Q *
& (D47(TA) + Q *
& (D48(TA) + Q *
& (D49(TA) + Q *
& (D50(TA) + Q *
& (D51(TA) + Q *
& (D52(TA) + Q *
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& (D57(TA) + Q *
& (D58(TA) + Q *
& (D59(TA) + Q *
& (D60(TA) + Q *
& (D61(TA) + Q *
& (D62(TA) + Q *
& (D63(TA) + Q *
& (D64(TA) + Q *
& (D65(TA) + Q *
& (D66(TA) + Q *
& (D67(TA) + Q *
& (D68(TA) + Q *
& (D69(TA) + Q *
& (D70(TA) + Q *
& (D71(TA) + Q *
& (D72(TA) + Q *
& (D73(TA) + Q *
& (D74(TA) + Q *
& (D75(TA) + Q *
& (D76(TA) + Q *
& (D77(TA) + Q *
& (D78(TA) + Q *
& (D79(TA) + Q *
& (D80(TA) + Q *
& (D81(TA) + Q *
& (D82(TA) + Q *
& (D83(TA) + Q *
& (D84(TA) + Q *
& (D85(TA) + Q *
& (D86(TA) + Q *
& (D87(TA) + Q *
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& (D91(TA) + Q *
& (D92(TA) + Q *
& (D93(TA) + Q *
& (D94(TA) + Q *
& (D95(TA) + Q *
& (D96(TA) + Q *
& (D97(TA) + Q *
& (D98(TA) + Q *
& (D99(TA) + Q *
& (D100(TA) + Q *

PCM09610
PCM09620
PCM09630
PCM09640
PCM09650
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PCM09750
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PCM10000
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PCM10190
PCM10200
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PCM10240
PCM10250
PCM10260
PCM10270
PCM10280
PCM10290
PCM10300
PCM10310
PCM10320
PCM10330
PCM10340
PCM10350
PCM10360
PCM10370
PCM10380
PCM10390
PCM10400

```

&      ( D2( TA ) + 0 *
&      ( D3( TA ) + 0 *
&      ( D4( TA ) + 0 *
&      ( D5( TA ) + 0 *
&      ( D6( TA ) + 0 *
&      ( D7( TA ) + 0 *
&      ( D8( TA ) )))))))
C
C      FORMULA FOR FINDING TAU BY INTEGRATION
C
      DELTAU( TA, DEL, DELDEL ) =
&      DELDEL / ( 2 * TA * DEL ** 2 - 1 )
C
C      BEGIN EXECUTION. DEFINE QUANTITIES THAT WILL BE USED
C      REPEATEDLY IN THE RESIDUE SUMMING LOOP.
      DELNEW = K * EXP( (0,1) * ( 3*PI/4 - PSI ) )
      IF ( DELNEW.NE. DEL ) THEN
        DEL = DELNEW
        QSQR = ( 1 / DEL ) ** 2
        NPTS = 0
      ENDIF
C
C      CLEAR THE VARIABLE USED TO ACCUMULATE THE RESIDUE SUM, AND
C      INITIALIZE VARIABLE USED TO TEST CONVEGENCE.
      ZS = 0
      TEST = 0
C
C      BEGIN CALCULATION OF SUM OF RESIDUES. IF THE S-TH RESIDUE
C      POINT HAS BEEN LOCATED BY A PREVIOUS CALL TO THIS SUBROUTINE,
C      JUMP PAST THE CALCULATION OF TAU(S).
      DO 100 S = 1, MTERMS
        IF ( S.LE. NPTS ) GOTO 90
C
C      FIND TAU(S) FROM TAU0 IF K IS SMALL, OR FROM TAU1 IF K IS
C      LARGE. FOR INTERMEDIATE VALUES OF K, ACCURACY REQUIRES AN
C      INTEGRATION PROCEDURE.
      TAU0 = TFN( AIRY0( S ) )
      TAU1 = TFN( AIRY1( S ) )
      IF ( ABS( TAU0 * K ** 2 ) .LT. 0.25 ) THEN
        TAU( S ) = TAU0( TAU0, DEL )
      ELSE IF ( ABS( TAU1 * K ** 2 ) .GT. 1.0 ) THEN
        TAU( S ) = TAU1( TAU1, 1. / DEL )
      ELSE
        T = TAU0
        N = MAX( ABS( DEL ) / FINE, 2.00 )
        DELDEL = DEL / N
C
C      INTEGRATE ALONG A DIAGONAL PATH IN THE COMPLEX DEL-PLANE
C      USING A FOURTH-ORDER RUNGE-KUTTA METHOD. THE VARIABLE OF
C      INTEGRATION, DEL1, RUNS FROM 0 TO DEL.
      DEL1 = 0
      DO 80 I = 1, N
        TK1 = DELTAU( T, DEL1, DELDEL )
        DEL1 = DEL1 + DELDEL/2
        TK2 = DELTAU( T + TK1/2, DEL1, DELDEL )
        TK3 = DELTAU( T + TK2/2, DEL1, DELDEL )
        DEL1 = DEL1 + DELDEL/2
        TK4 = DELTAU( T + TK3, DEL1, DELDEL )
        T = T + ( TK1 + 2 * TK2 + 2 * TK3 + TK4 ) / 6
      80 CONTINUE
      TAU( S ) = T
      ENDIF
C
C      REMEMBER HOW MANY RESIDUE POINTS HAVE BEEN LOCATED. THIS
C      PERMITS PEUSE OF THIS SUBROUTINE WITHOUT RECALCULATION OF
C      TAU( S ) SO LONG AS THE VALUE OF DEL REMAINS UNCHANGED, THAT
C      IS FOR CHANGES IN DISTANCE ONLY.
      NPTS = S
C
C      EVALUATE RESIDUE AT TAU( S ) .
      90 CONTINUE
      Z = EXP( (0,1) * TAU( S ) * CHI ) /
&      ( 2 * TAU( S ) - QSQR )

```

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PCM10410
PCM10420
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PCM10500
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PCM10590
PCM10600
PCM10610
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PCM10690
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PCM11070
PCM11080
PCM11090
PCM11100
PCM11110
PCM11120
PCM11130
PCM11140
PCM11150
PCM11160
PCM11170
PCM11180
PCM11190
PCM11200

```

```

C ACCUMULATE AND LOOP TO CALCULATE NEXT RESIDUE UNTIL
C TEST INDICATES CONVERGENCE IS SATISFACTORY.
      ZS = ZS + Z
      IF ( ABS( ZS - TEST ) .LT. PRECIS * ABS( TEST ) )
        & GOTO 200
      TEST = ZS
100 CONTINUE
C EXIT WITH DIFFRACTION LOSS DETERMINED FROM SUM OF RESIDUES.
200 CONTINUE
      ATTEM = ABS( ZS ) * SORT( 2 * PI * CHI )
      RETURN
      END

```

```

C *****
C AIRY0 - LOCATE THE ZEROS OF THE AIRY FUNCTION
C THE ZEROS OF INTEREST ARE LOCATED ON THE NEGATIVE REAL
C AXIS, AND THIS SUBPROGRAM RETURNS POSITIVE VALUES EQUAL
C TO MINUS THE X-COORDINATE OF THESE ZEROS.
C THE FIRST 10 VALUES OF AIRY0 ARE THOSE TABULATED IN
C ABRAMOWITZ AND STEGUN, HANDBOOK OF FUNCTIONS, PAGE 478;
C VALUES FOR INDICES LARGER THAN 10 ARE CALCULATED USING
C EQUATIONS (10.4.94) ETC. ON PAGE 450 OF THE SAME REFERENCE.
C *****

```

```

C FUNCTION AIRY0( S )
C IMPLICIT REAL*8 ( A-H, O-Z )
C INTEGER S
C DIMENSION AO( 10 )
C DATA PI / 3.1415927 /
C DATA AO /
C & 2.3381074, 4.0879494, 5.5205598, 6.7867081,
C & 7.9441336, 9.0226508, 10.0401743, 11.0085243,
C & 11.9360156, 12.8287767 /
C FORMULAS USED TO CALCULATE THE AMPLITUDES AO( S ) FOR
C FOR INDICES GREATER THAN 10.
C X( S ) = 3 * PI * ( 4 * S - 1 ) / 8
C F( Y ) = Y ** ( 2./3 ) * ( 1 + 5./48 * ( 1/Y ) ** 2 )
C IF ( S .LE. 10 ) THEN
C   AIRY0 = AO( S )
C ELSE
C   AIRY0 = F( X( S ) )
C ENDIF
C RETURN
C END

```

```

C *****
C AIRY1 - LOCATE THE ZEROS OF THE DERIVATIVE OF THE AIRY
C FUNCTION.
C THE ZEROS OF INTEREST ARE LOCATED ON THE NEGATIVE REAL
C AXIS, AND THIS SUBPROGRAM RETURNS POSITIVE VALUES EQUAL
C TO MINUS THE X-COORDINATES OF THESE ZEROS.
C THE FIRST 10 VALUES OF AIRY1 ARE THOSE TABULATED IN
C ABRAMOWITZ AND STEGUN, HANDBOOK OF FUNCTIONS, PAGE 478;
C VALUES FOR INDICES LARGER THAN 10 ARE CALCULATED USING
C EQUATIONS (10.4.94) ETC. ON PAGE 450 OF THE SAME REFERENCE.
C *****
C FUNCTION AIRY1( S )
C IMPLICIT REAL*8 ( A-H, O-Z )
C INTEGER S

```

PCN11210
 PCN11220
 PCN11230
 PCN11240
 PCN11250
 PCN11260
 PCN11270
 PCN11280
 PCN11290
 PCN11300
 PCN11310
 PCN11320
 PCN11330
 PCN11340
 PCN11350
 PCN11360
 PCN11370
 PCN11380
 PCN11390
 PCN11400
 PCN11410
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 PCN11490
 PCN11500
 PCN11510
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 PCN11900
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 PCN11920
 PCN11930
 PCN11940
 PCN11950
 PCN11960
 PCN11970
 PCN11980
 PCN11990
 PCN12000

```

C      DIMENSION A1( 10 )
      DATA PI / 3.1415927 /
C
C      DATA A1 /
      &      1.0187930, 3.2481976, 4.8200992, 6.1633074,
      &      7.3721773, 8.4884867, 9.5354491, 10.5276604,
      &      11.4750566, 12.3847984 /
C
C      FORMULAS USED TO CALCULATE THE AMPLITUDE A1( S ) FOR
C      INDICES GREATER THAN 10.
      Y( S ) = 3 * PI * ( 4 * S - 3 ) / 8
      G( X ) = X ** (2./3) * ( 1 - 7./48 * (1/X) ** 2 )
C
C      IF ( S .LE. 10 ) THEN
      AIRY1 = A1( S )
      ELSE
      AIRY1 = G( Y( S ) )
      ENDIF
      RETURN
C
      END

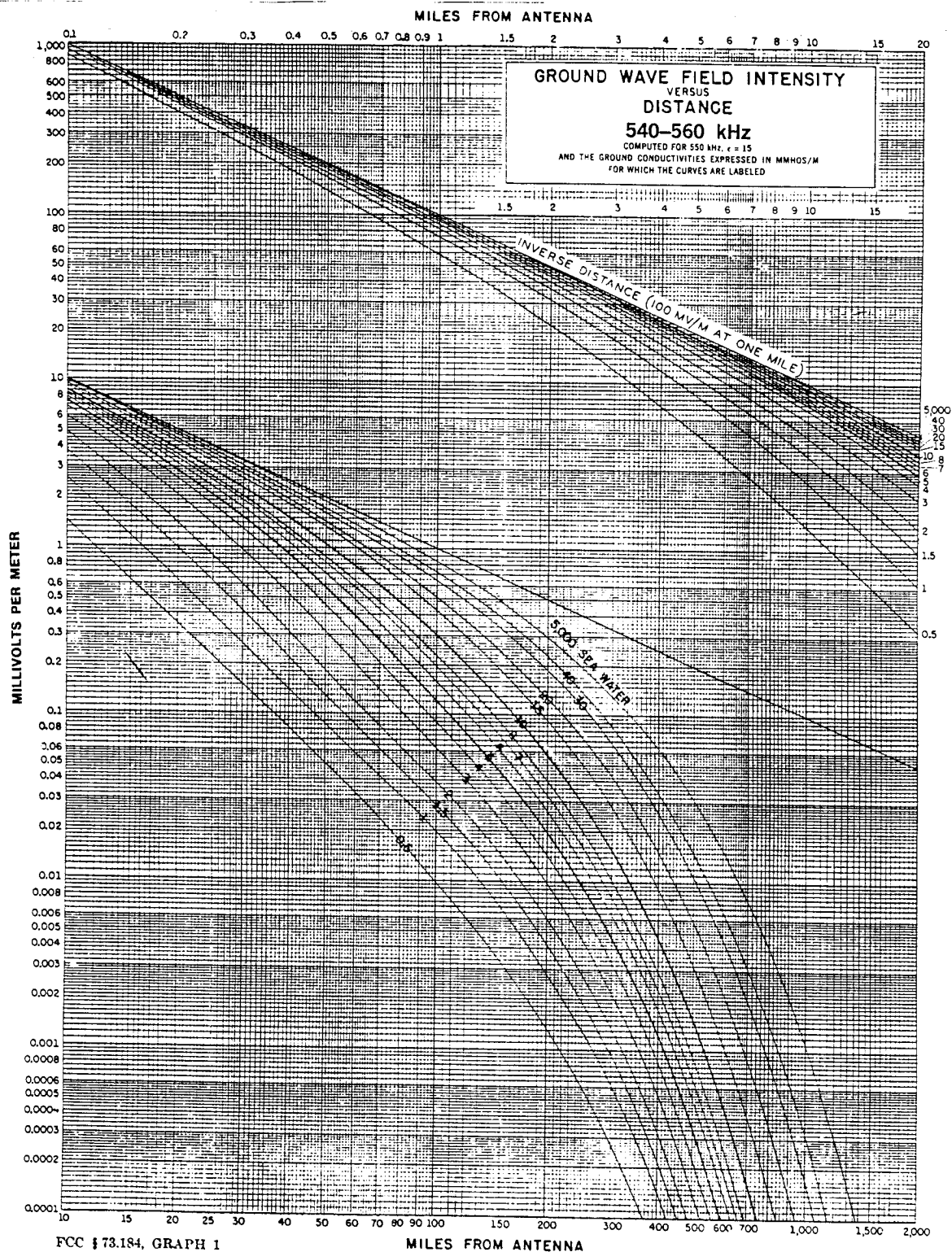
```

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PCN12040
PCN12050
PCN12060
PCN12070
PCN12080
PCN12090
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PCN12110
PCN12120
PCN12130
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PCN12150
PCN12160
PCN12170
PCN12180
PCN12190
PCN12200
PCN12210
PCN12220

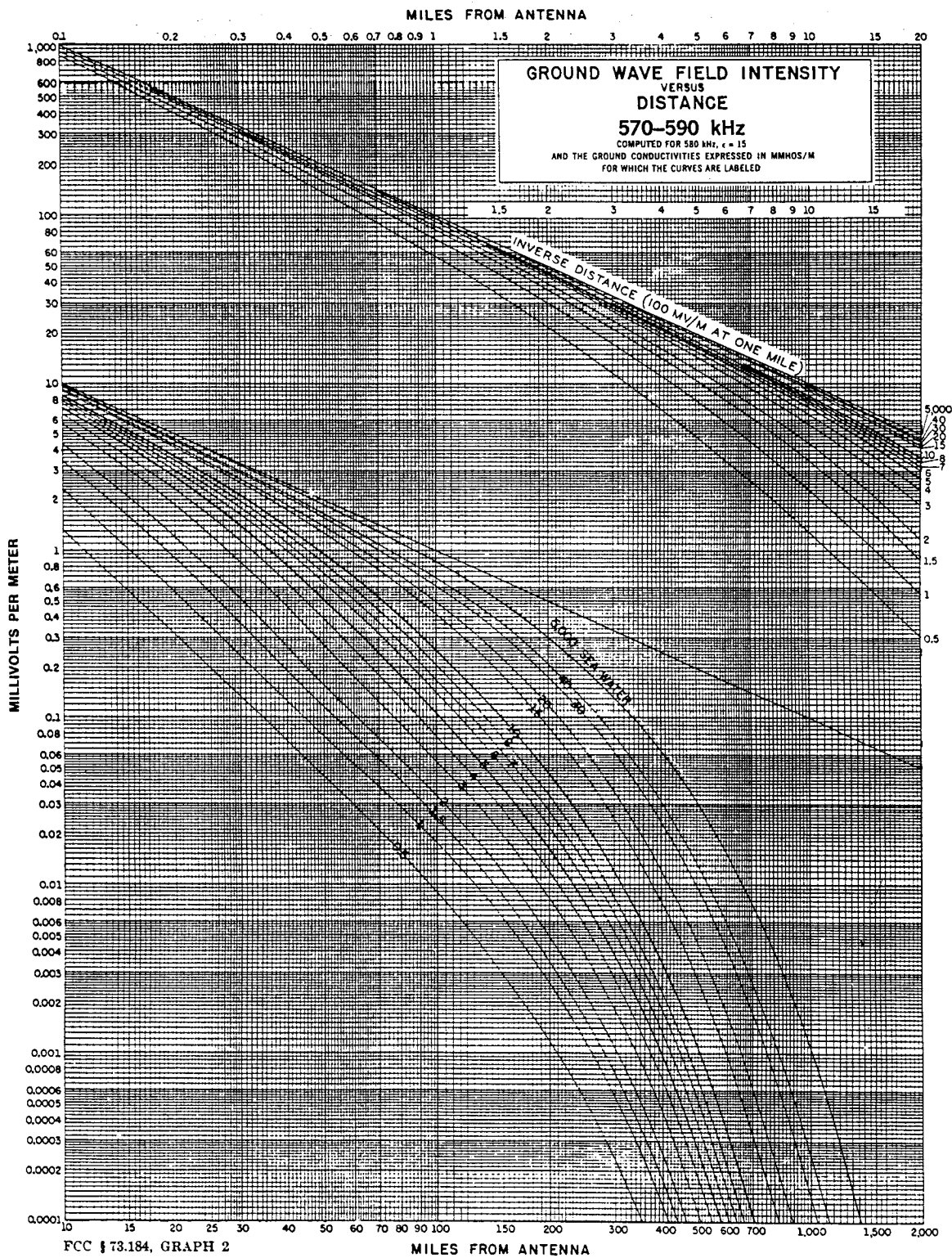
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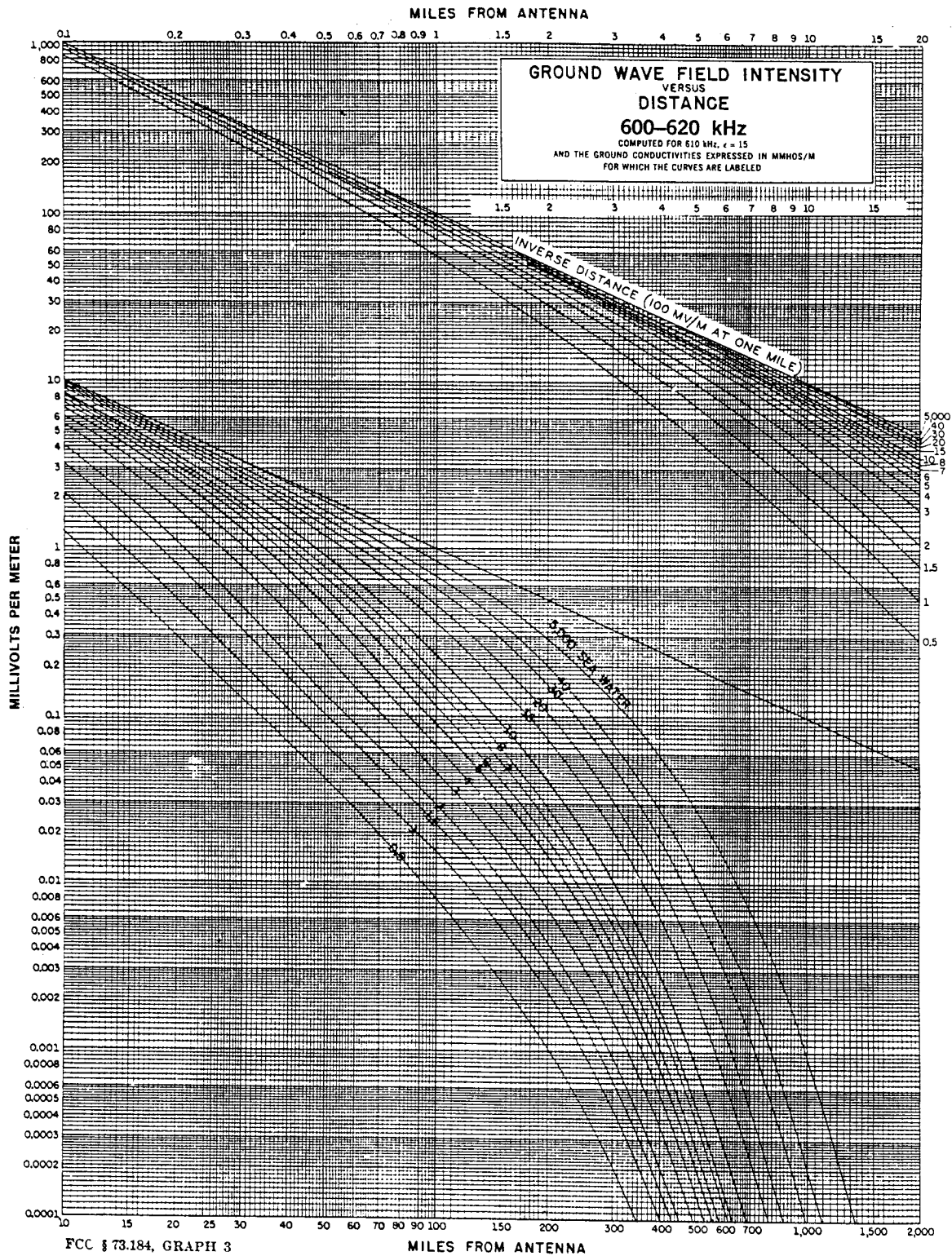
APPENDIX C. FCC GROUND WAVE FIELD STRENGTH CHARTS

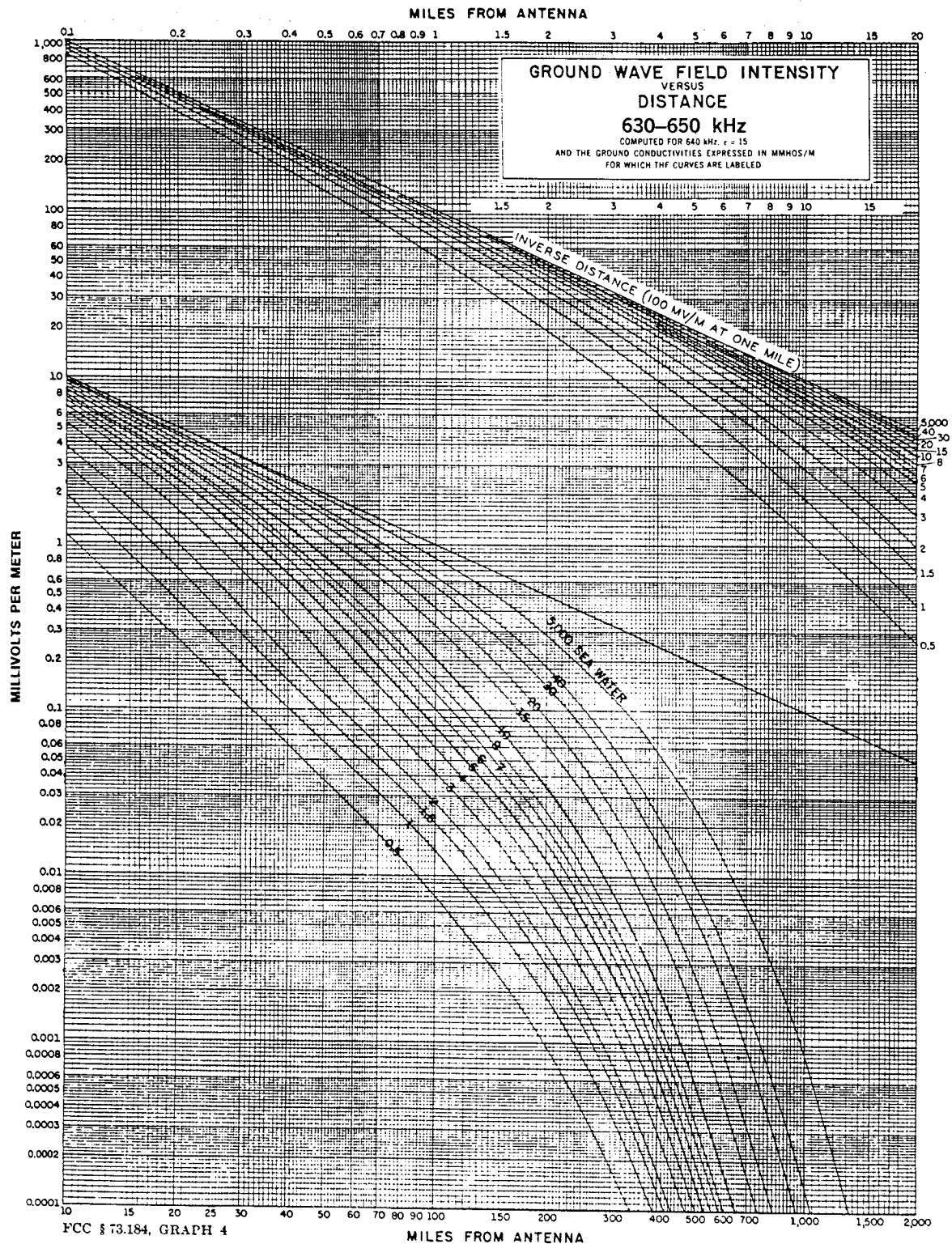


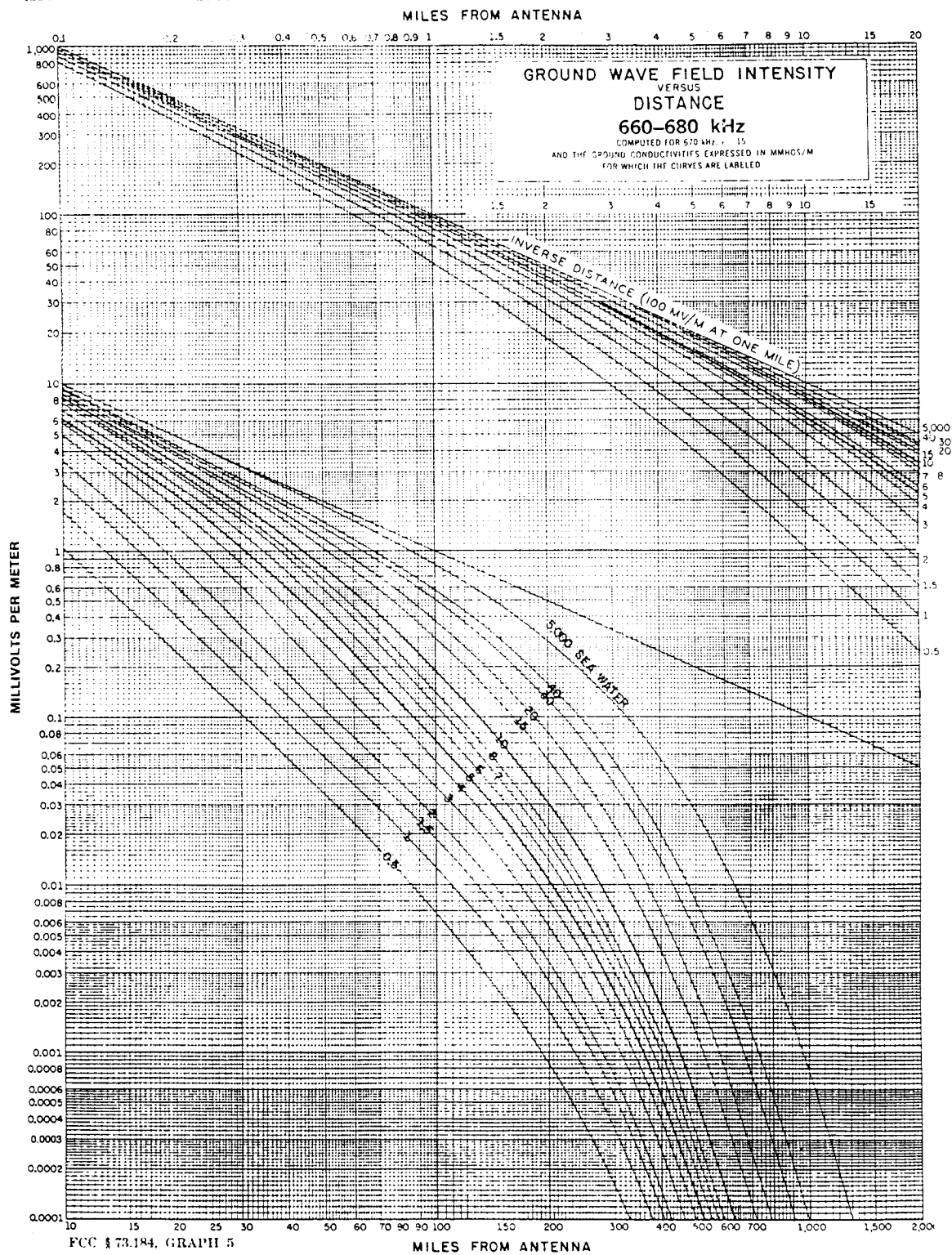
FCC §73.184, GRAPH 1

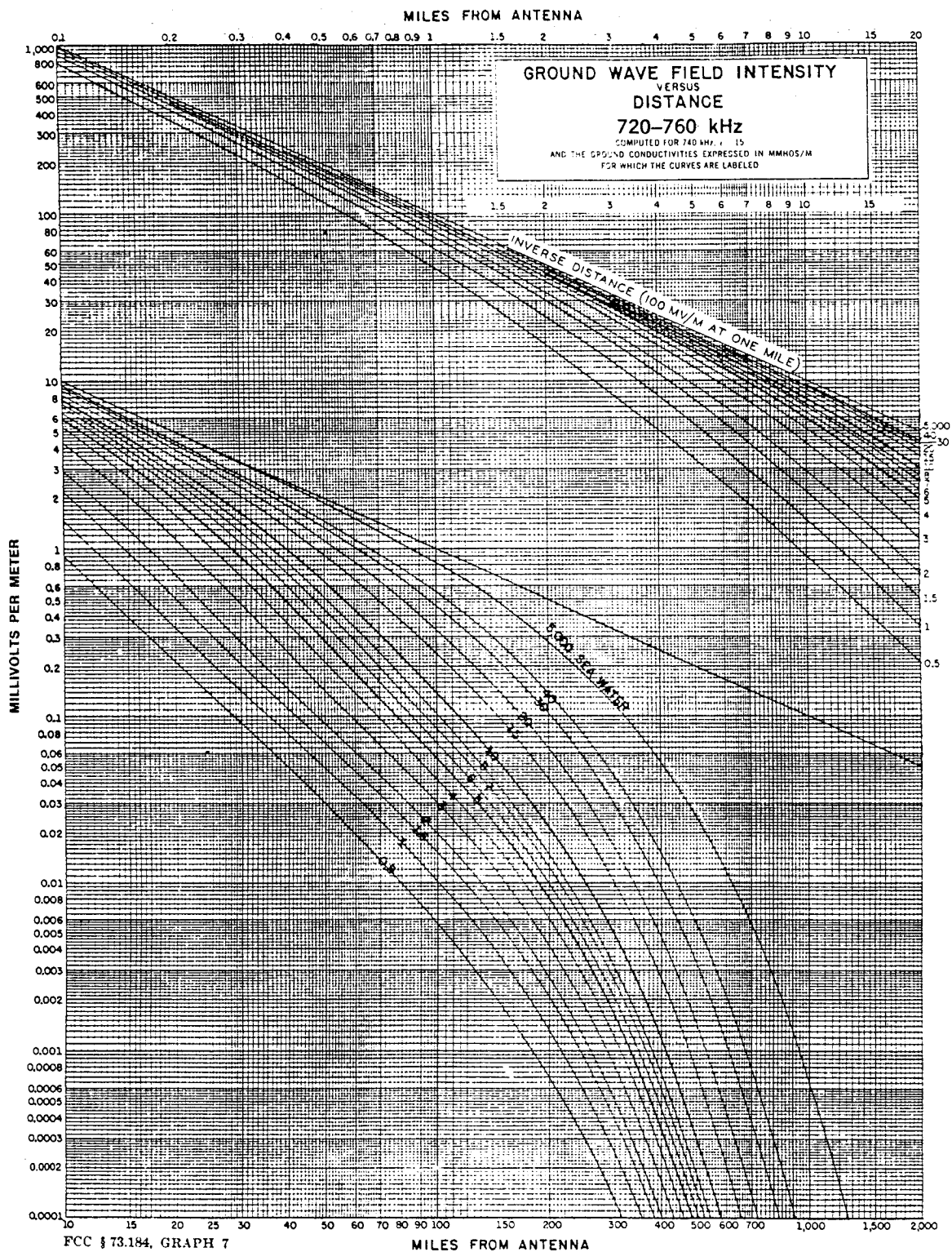
MILES FROM ANTENNA

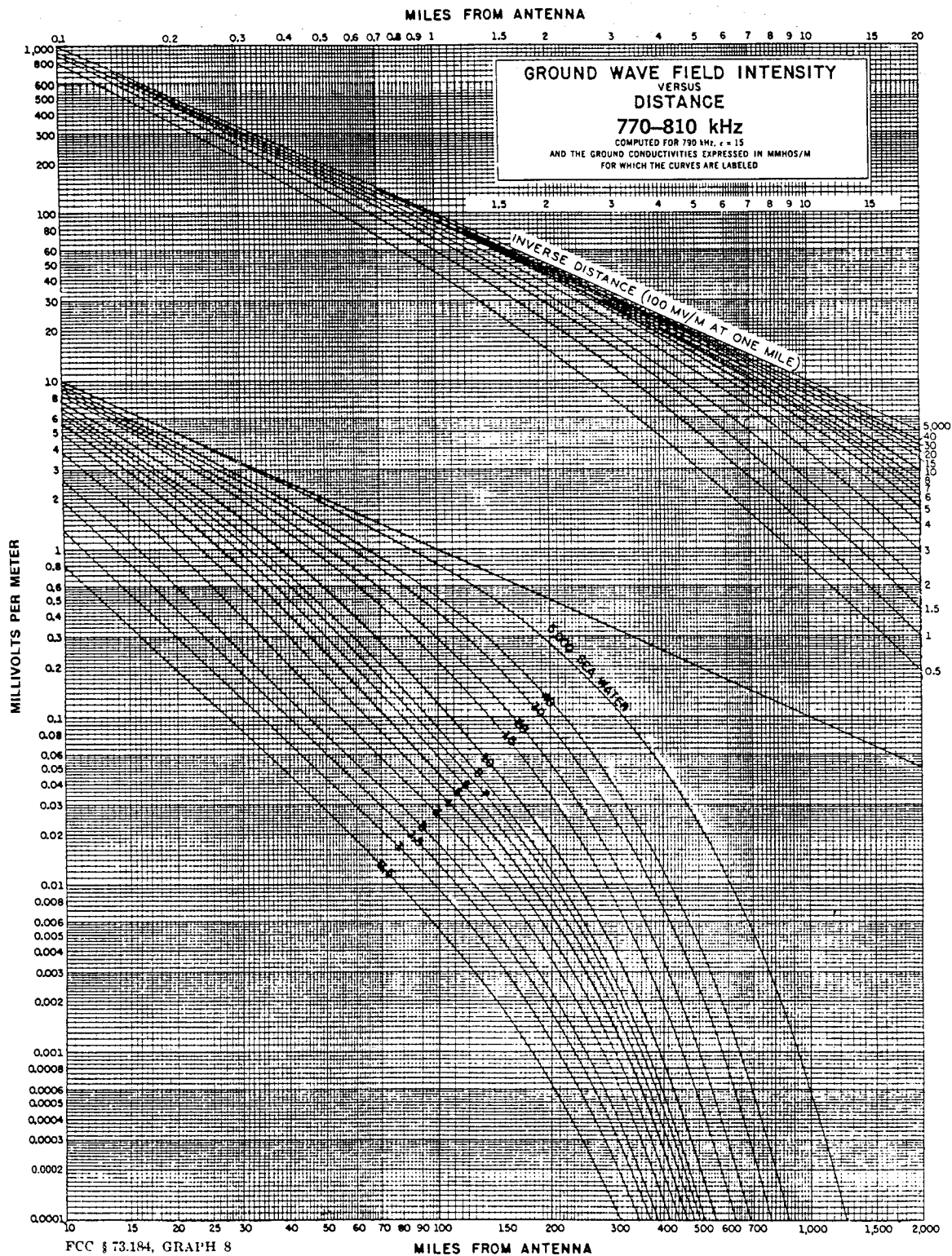


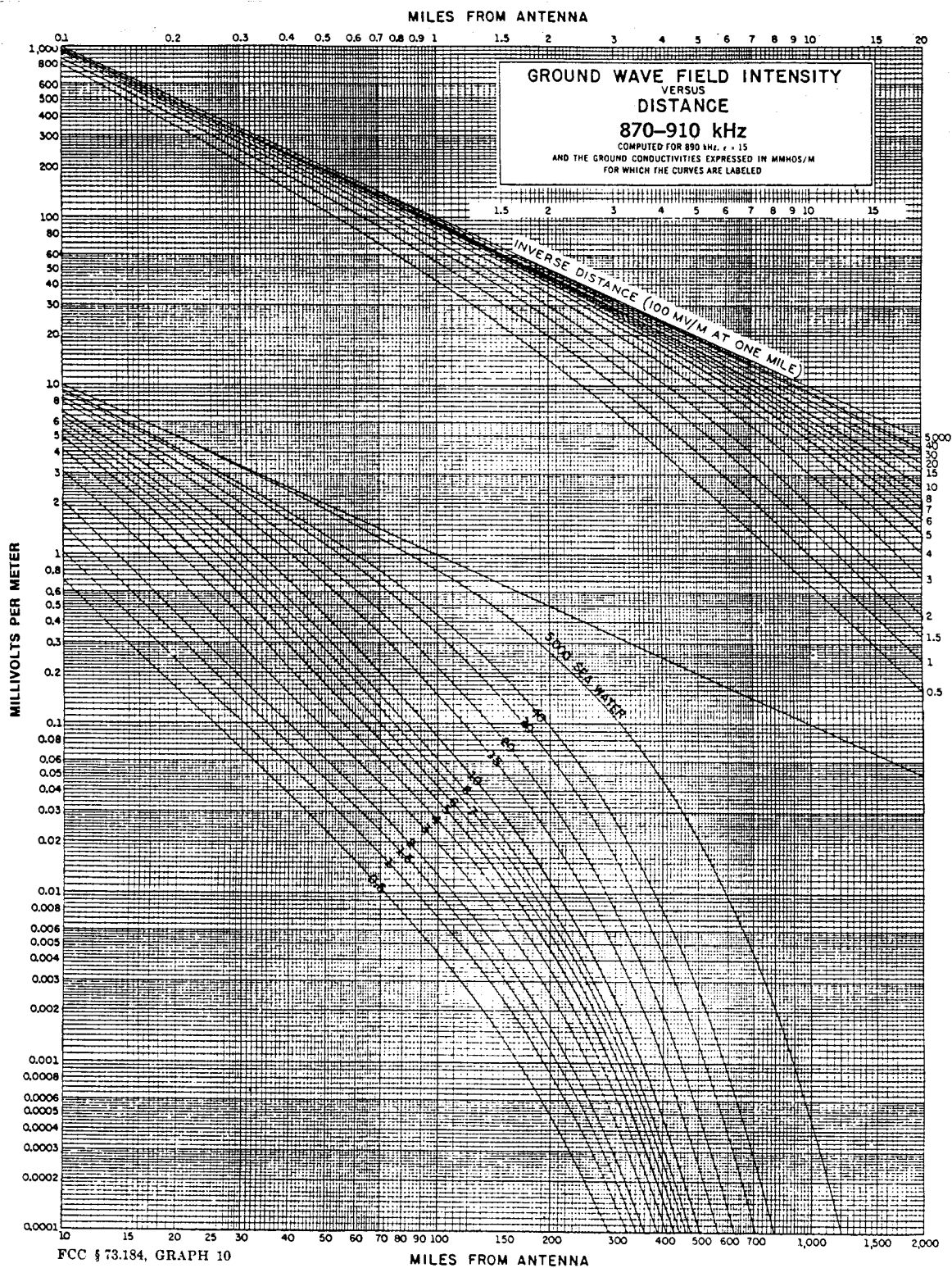


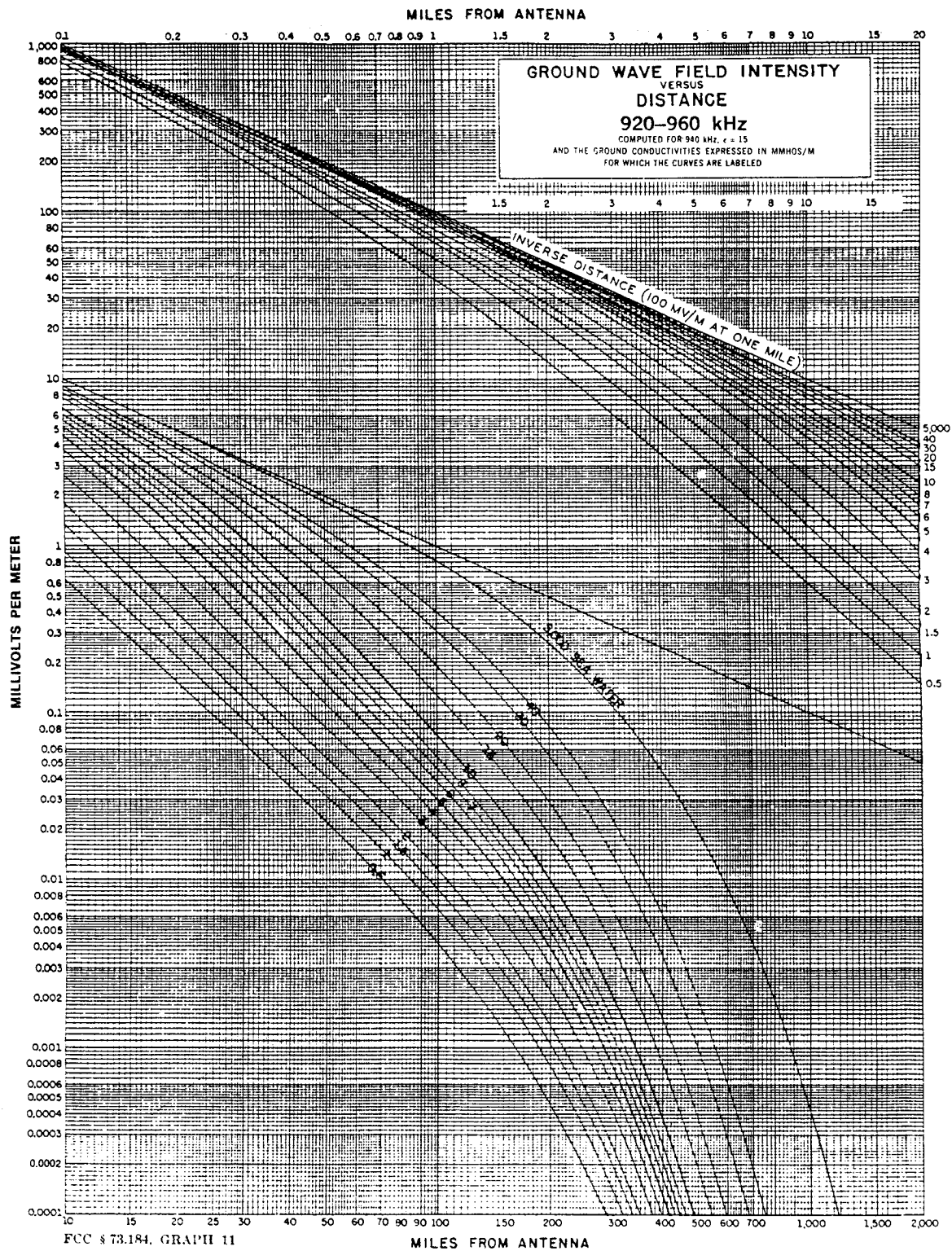


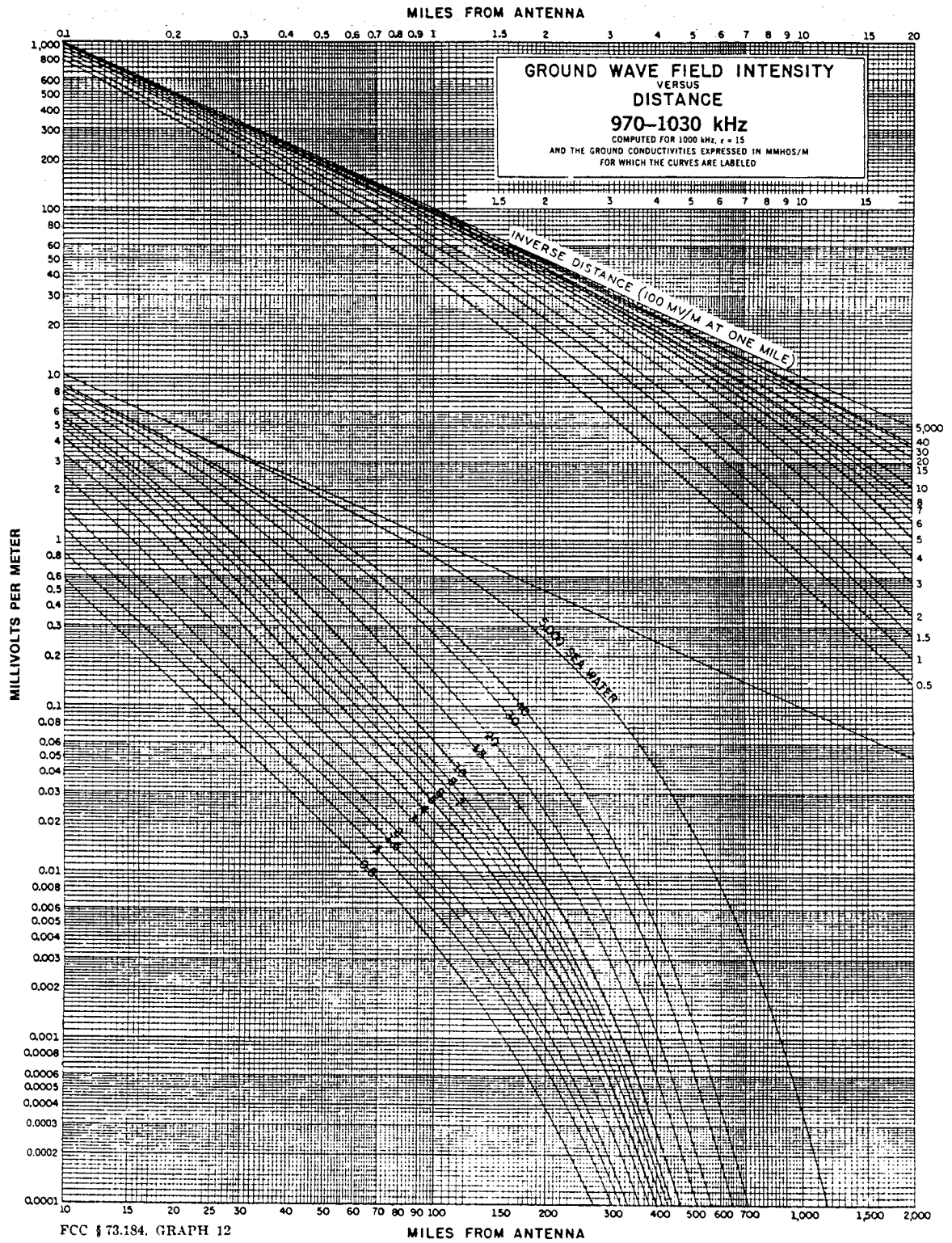


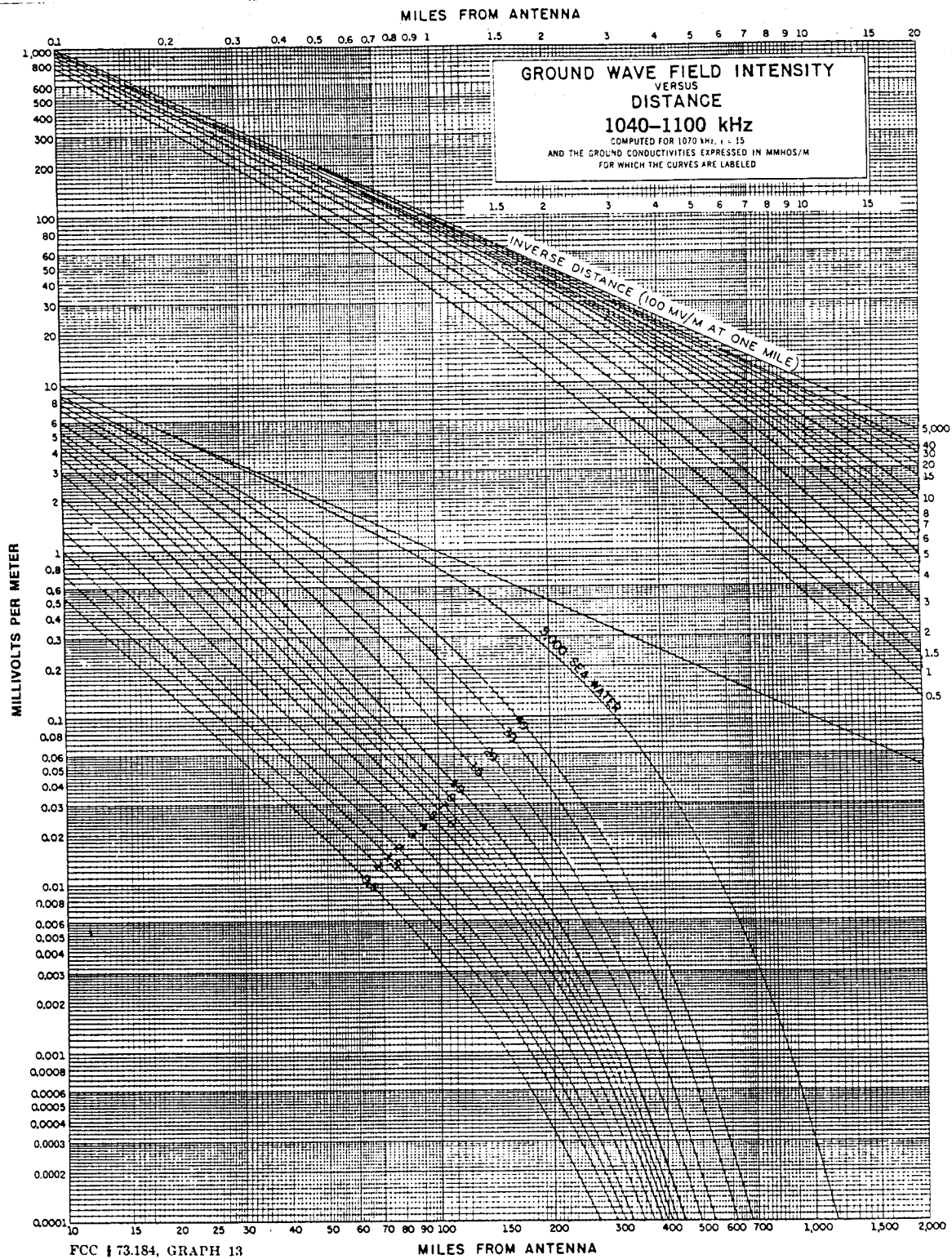


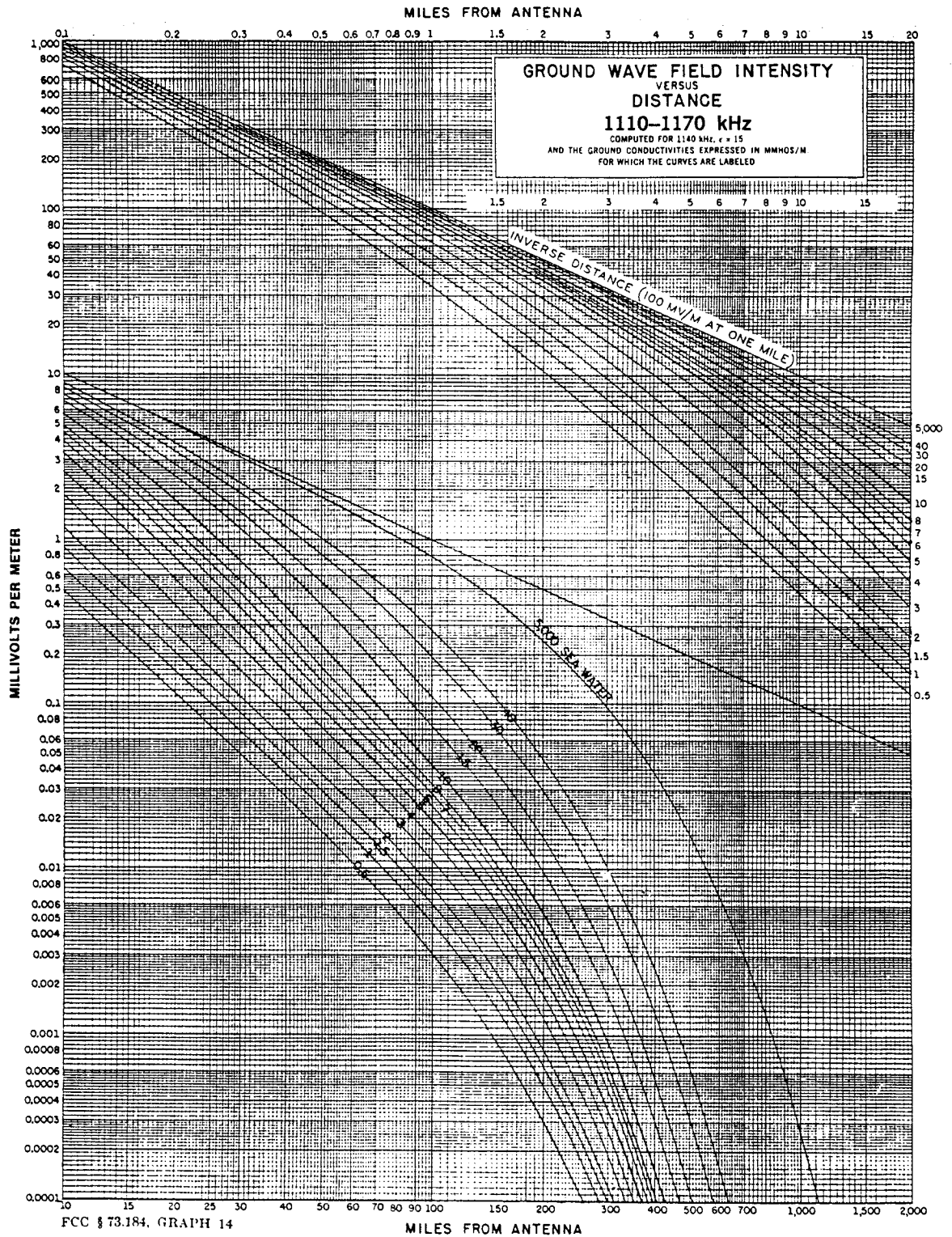


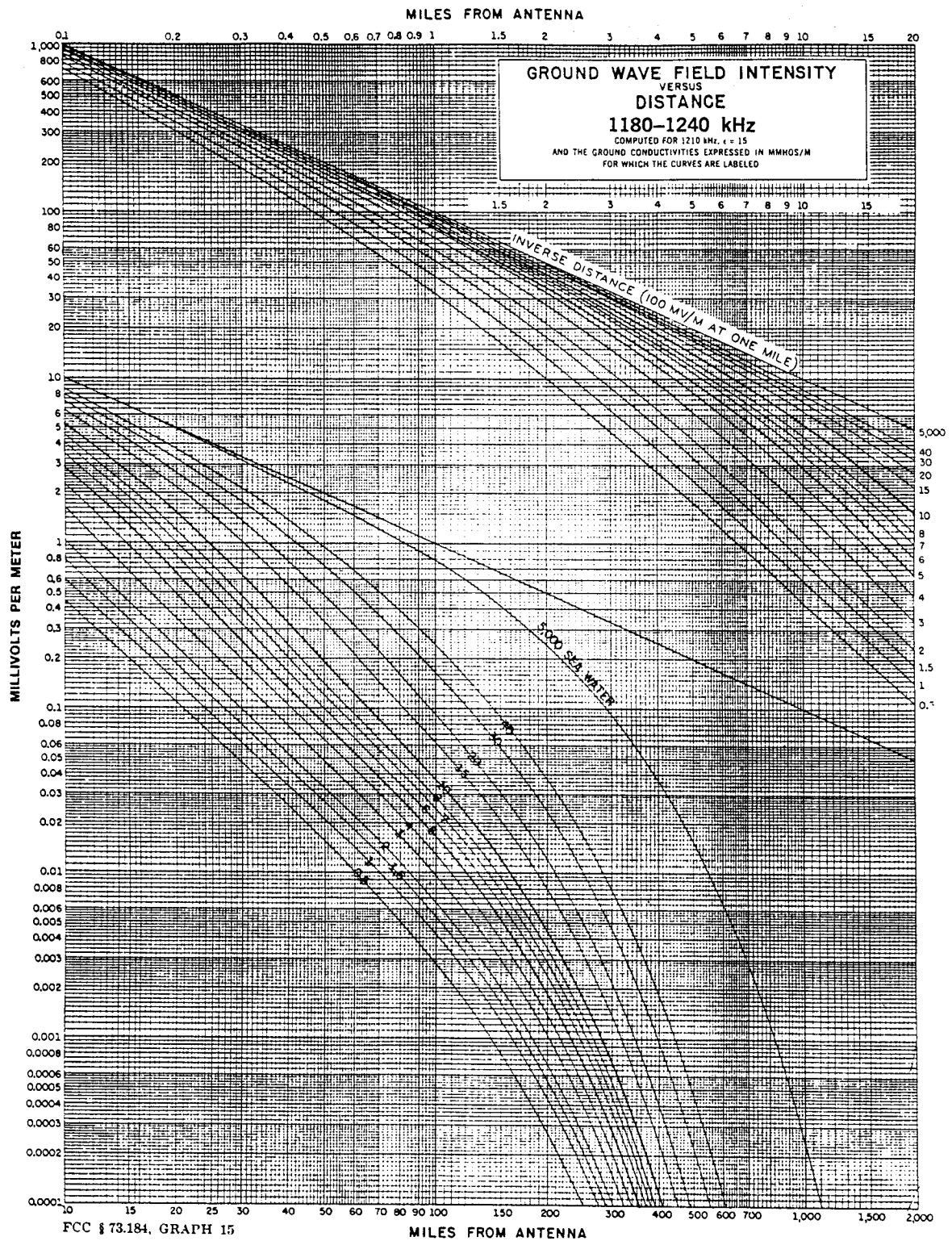


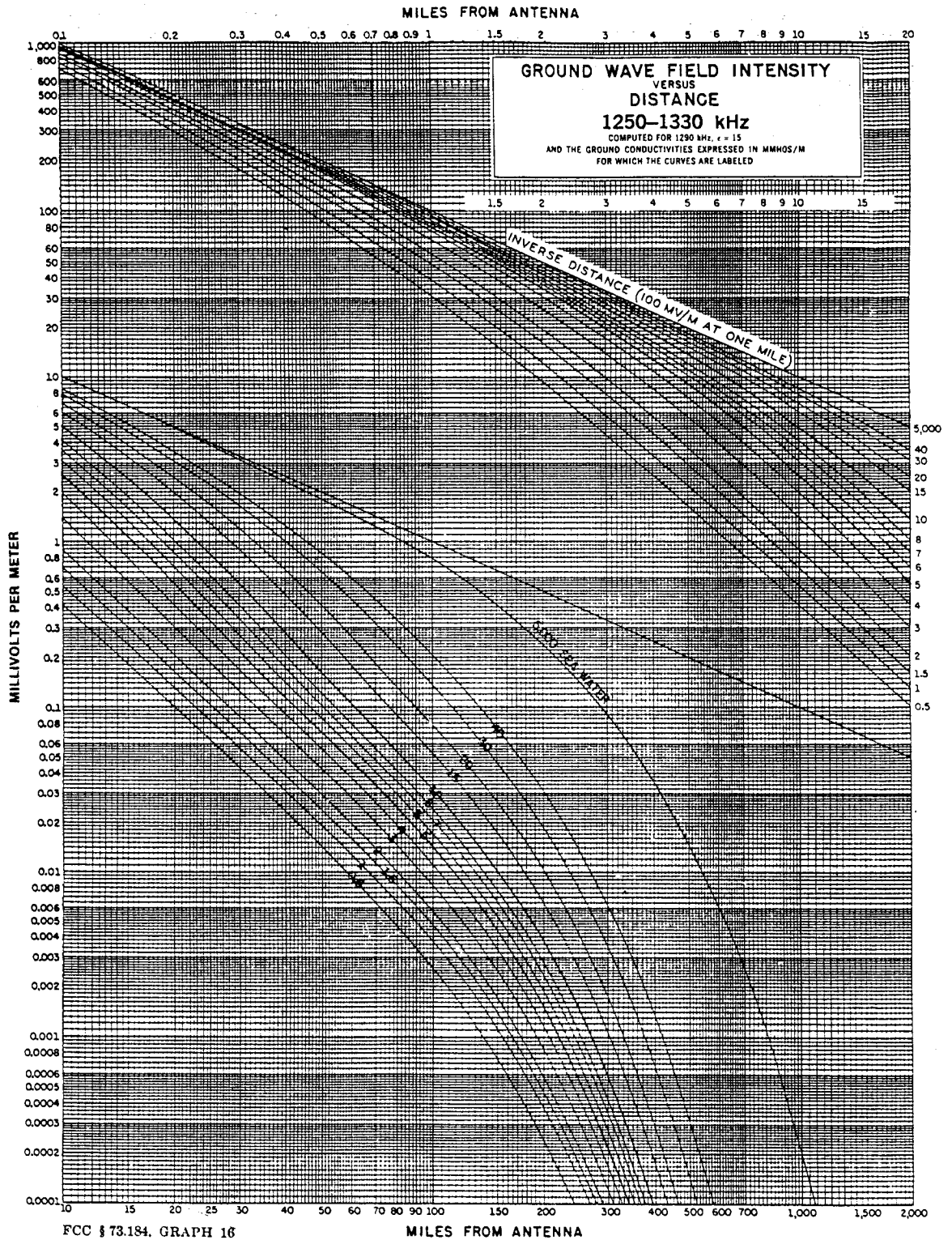


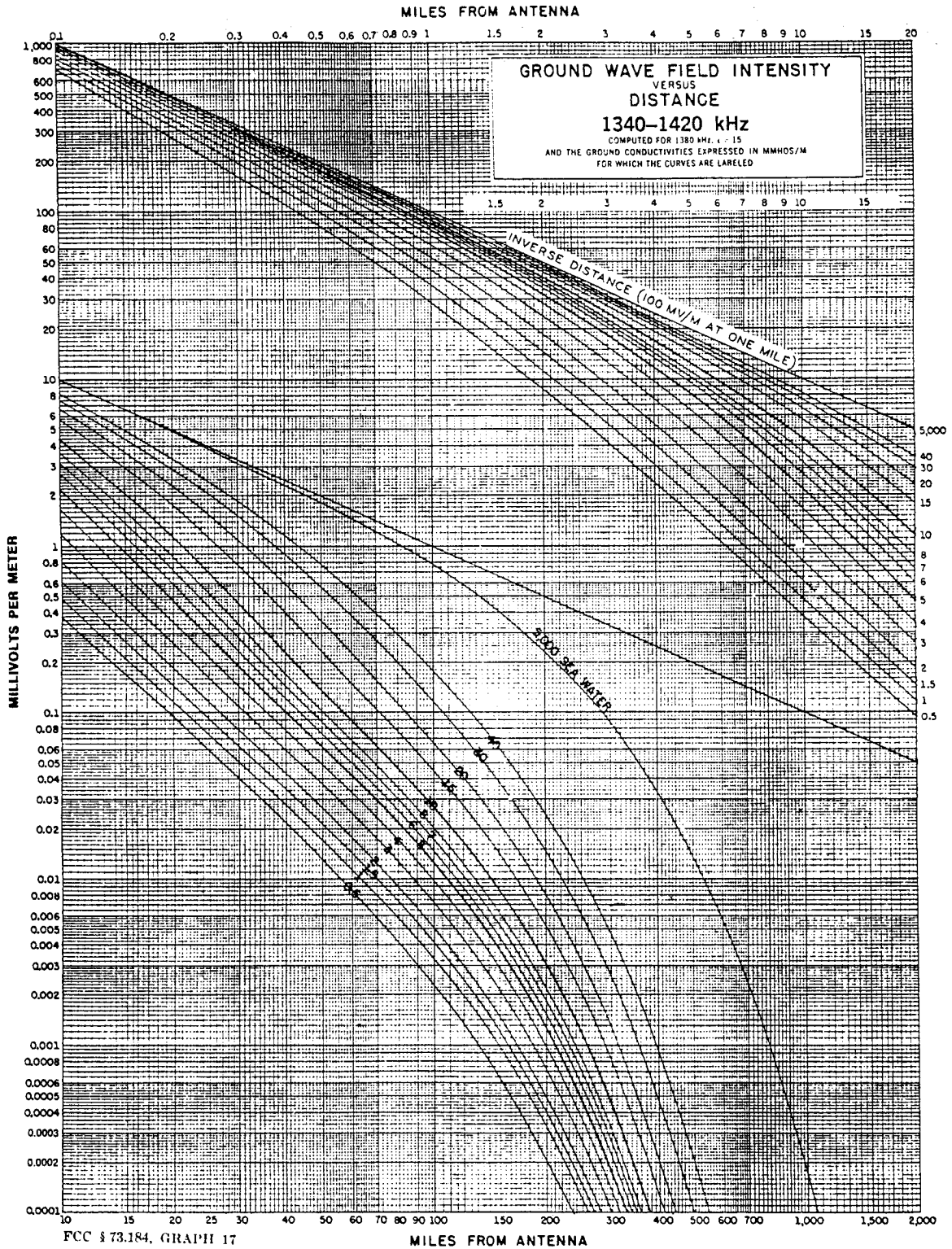


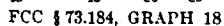


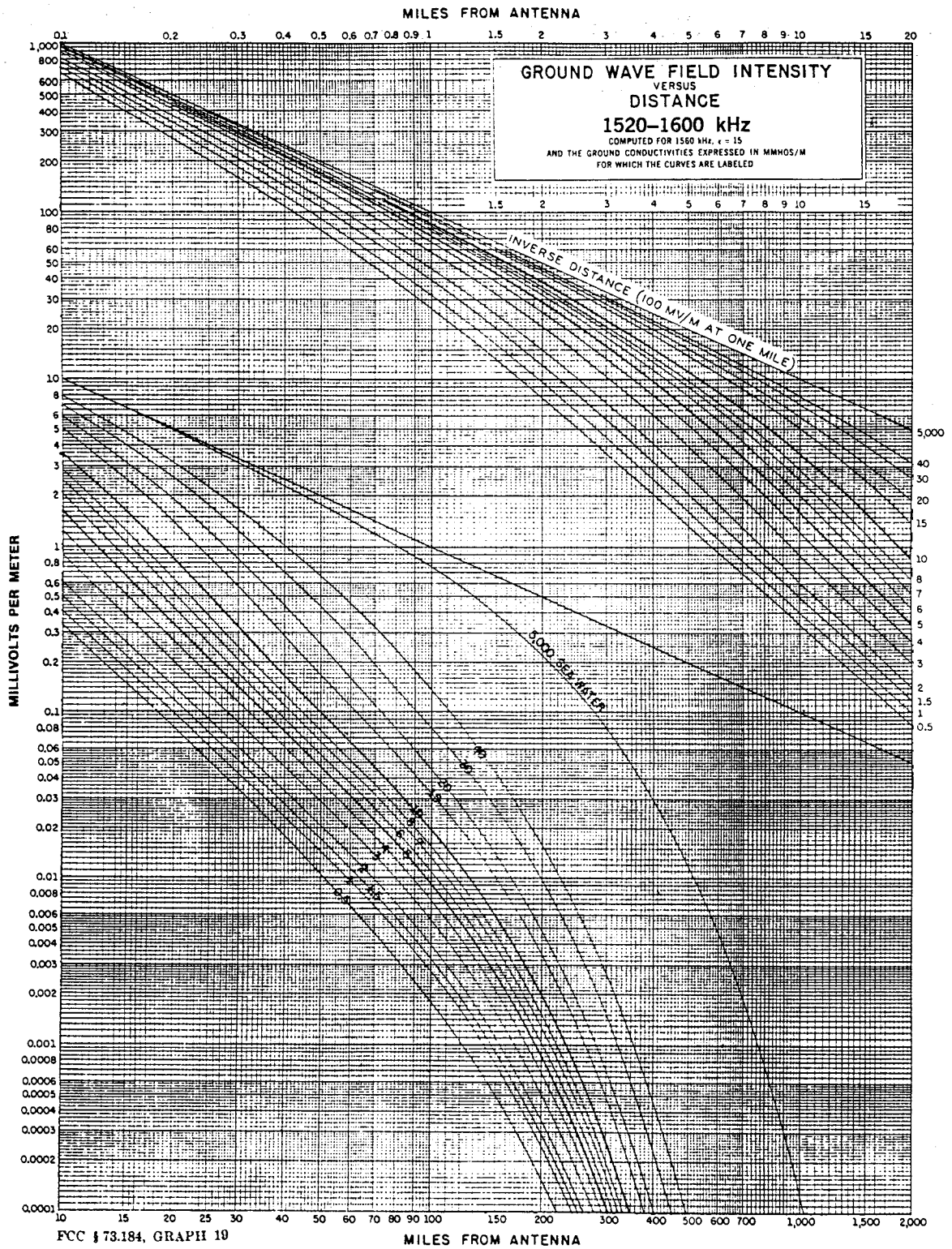


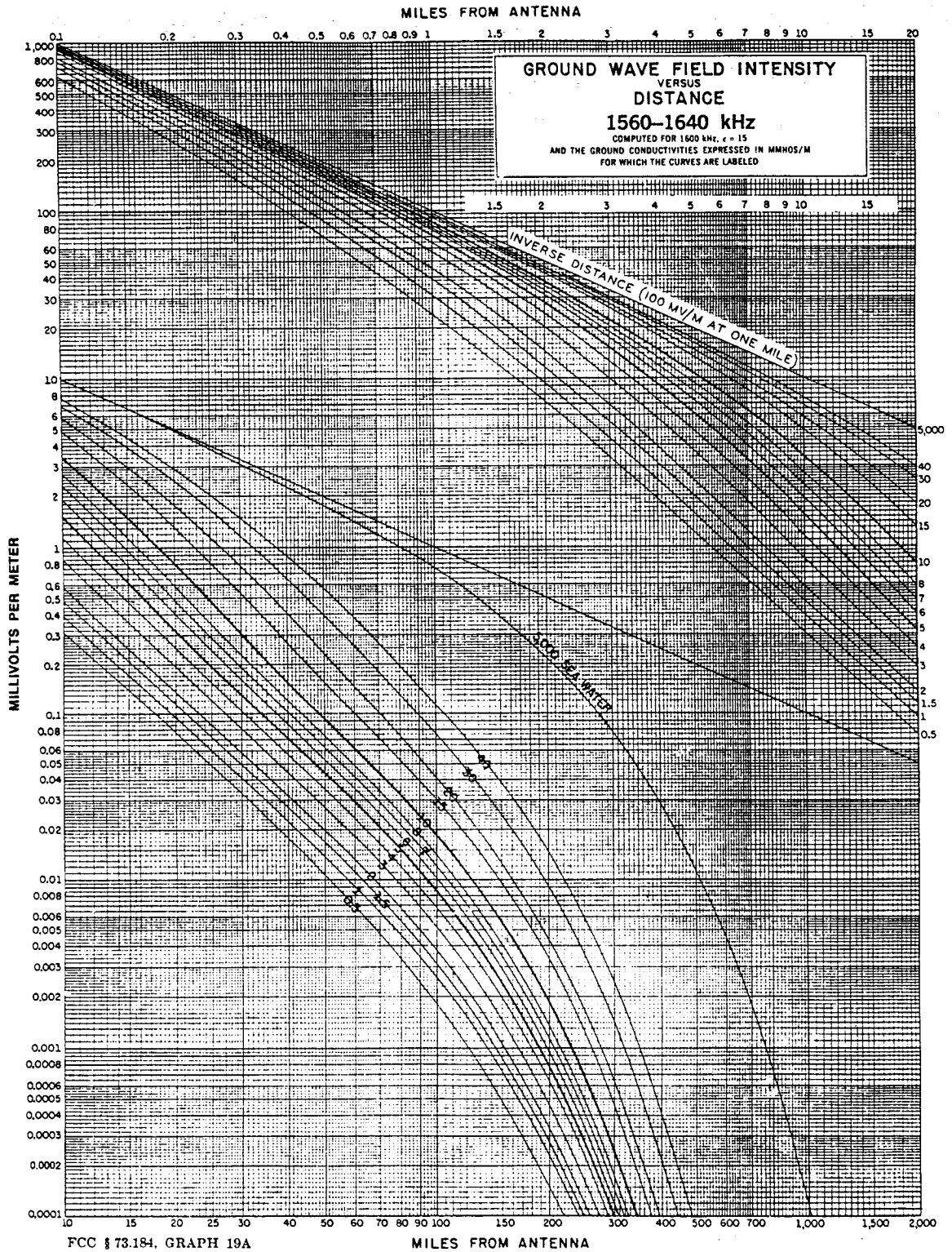












APPENDIX D. EXAMPLE PROBLEM

This appendix contains the protected contour information of those stations not culled from the data base listing for feasibility studies of both the Smethport, and Williamsport, Pennsylvania service areas. These results were obtained using program PCMLL2, and the assumptions stated in Chapter IV. Figures D1 through D5 show the actual program output for the 0.5 mV/m contour for the stations listed on page 29. Figure D6 shows the location of the protected contours sketched in Figure D7 on a map of Pennsylvania. Figure D8 shows the maximum pattern envelope resulting from these contours.

FILE: WGR OUTPUT B WGR CONVERSATIONAL MONITOR SYSTEM

 LATITUDE : 12. 36. 40.000
 LONGITUDE : 78. 50. 39.000

LATITUDE : 41. 12. 36.000
 LONGITUDE : 77. 50. 39.000

FREQUENCY : 500.0 MHz

PROTECTED COPTURE LEVEL : 0.500000 WGR

BEARING	FIELD	RA	RADIUS	LATITUDE	LONGITUDE	DIST	BEARING
120.0	1472.55	186.000	11. 54. 52.40	76. 53. 48.62	79.692	9.016	
130.0	1472.55	186.000	11. 30. 35.60	77. 7. 40.94	53.003	352.570	
140.0	1472.55	186.000	11. 28. 37.16	77. 24. 31.50	42.065	314.675	
150.0	1472.55	186.000	11. 18. 47.85	77. 43. 49.63	58.422	281.484	
160.0	1472.55	186.000	11. 11. 34.62	78. 5. 1.57	86.834	268.652	
170.0	1472.55	186.000	11. 9. 0.54	78. 16. 8.87	102.555	266.084	
180.0	1472.55	186.000	11. 7. 2.88	78. 27. 30.76	118.641	264.842	
			11. 5. 40.63	78. 39. 39.00	134.861	264.484	
					151.068	264.678	

FILE: WFRB OUTPUT B WFRB CONVERSATIONAL MONITOR SYSTEM

 LATITUDE : 30. 11. 57.000
 LONGITUDE : 78. 50. 39.000

LATITUDE : 41. 12. 36.000
 LONGITUDE : 77. 50. 39.000

FREQUENCY : 500.0 MHz

PROTECTED COPTURE LEVEL : 0.500000 WFRB

BEARING	FIELD	RA	RADIUS	LATITUDE	LONGITUDE	DIST	BEARING
0.0	1529.00	185.070	11. 20. 53.37	78. 57. 57.00	161.199	274.967	
10.0	1529.00	185.070	11. 19. 20.09	78. 34. 51.86	128.944	275.214	
20.0	1529.00	185.070	11. 11. 43.31	78. 12. 31.96	97.344	272.150	
30.0	1529.00	185.070	11. 7. 12.06	77. 51. 40.74	68.923	261.605	
40.0	1529.00	185.070	10. 57. 0.99	77. 32. 58.18	50.925	235.789	
50.0	1529.00	185.070	10. 34. 24.82	77. 16. 59.31	55.411	200.970	
60.0	1529.00	185.070	10. 30. 3.56	77. 4. 11.08	78.530	181.477	
70.0	1529.00	185.070	11. 11. 6.64	76. 55. 1.50	108.294	174.294	
80.0	1529.00	185.070	11. 11. 11.41	76. 49. 39.20	140.591	172.527	

Figure D1. Results for WGR and WFRB

FILE: WFIL OUTPUT B V-32P CONVERSATIONAL MONITOR SYSTEM

 LATITUDE : 40. 15. 59.000 0
 LONGITUDE : 75. 38. 47.000 0

 LATITUDE : 41. 12. 59.000 0
 LONGITUDE : 77. 2. 46.000 0

FREQUENCY : 500.0 KHZ

 PROTECTED CARRIER LEVEL : 0.500000 1V/W

BEARING	FIELD ° RM	RADIUS	LATITUDE	LONGITUDE	DIST	BEARING
0.0	504.36	121.133	11. 29.84	15. 47.00	149.203	91.336
10.0	397.32	110.030	11. 28.15	15. 6.54	169.052	95.670
20.0	357.69	105.385	11. 26.12	15. 1.14	186.965	98.148
30.0	357.69	105.385	11. 25.59	15. 0.55	204.890	99.793
40.0	356.88	105.287	11. 20.13	15. 31.78	221.250	101.987
50.0	361.49	105.889	11. 28.22	15. 12.81	237.119	104.426
60.0	424.87	113.072	11. 8.53	15. 11.57	256.251	106.160
70.0	571.15	127.270	11. 6.73	15. 50.52	280.519	107.708
80.0	767.62	142.821	11. 39.36	15. 15.36	306.519	110.080
90.0	971.55	156.273	11. 6.76	15. 34.06	329.877	111.382
100.0	1151.73	166.616	11. 25.08	15. 30.00	348.855	113.350
110.0	1288.71	173.740	11. 58.81	15. 31.65	362.609	117.722
120.0	1371.73	177.815	11. 10.90	15. 25.79	370.802	126.301
130.0	1395.81	179.065	11. 15.50	15. 25.74	373.327	130.945
140.0	1359.74	177.230	11. 16.50	15. 49.95	370.156	135.536
150.0	1265.39	172.521	11. 5.10	15. 3.36	361.111	139.960
160.0	1118.67	154.807	11. 18.76	15. 48.83	349.911	144.073
170.0	931.86	153.619	11. 18.76	15. 48.83	327.343	147.689
180.0	726.53	139.616	11. 13.49	15. 18.53	303.588	150.583
190.0	536.39	124.140	11. 32.60	15. 45.00	277.891	152.497
200.0	405.34	110.438	11. 59.65	15. 45.00	254.470	153.788
210.0	327.67	105.438	11. 41.37	15. 11.03	236.470	155.529
220.0	327.67	105.438	11. 37.07	15. 31.71	220.880	157.855
230.0	350.92	105.202	11. 15.87	15. 10.40	204.115	159.778
240.0	360.85	105.763	11. 1.21	15. 56.68	186.581	161.334
250.0	413.46	111.833	11. 57.23	15. 31.91	168.702	164.055
260.0	533.28	123.769	11. 18.58	15. 30.55	148.609	168.494
270.0	686.46	136.775	11. 1.21	15. 14.91	125.207	173.273
280.0	837.91	147.710	11. 3.48	15. 43.69	99.012	176.740
290.0	959.75	155.551	11. 55.15	15. 36.44	71.456	178.465
300.0	1035.91	160.111	11. 29.11	15. 37.65	45.766	185.638
310.0	1058.31	161.407	11. 24.10	15. 12.70	32.960	188.283
320.0	1028.62	159.162	11. 30.58	15. 16.53	46.733	192.207
330.0	938.85	154.243	11. 47.51	15. 10.04	72.601	198.521
340.0	808.70	145.753	11. 17.51	15. 35.99	100.080	202.301
350.0	654.76	131.273	11. 18.82	15. 31.53	126.077	206.354

Figure D2. Results for WFIL

FILE: WCKL 301100 W VA/NO OPERATIONAL MONITOR SYSTEM

LATITUDE : 12. 50. 0.000 0.000 0.000

LONGITUDE : 73. 35. 0.000 0.000 0.000

LATITUDE : 12. 50. 0.000 0.000 0.000

LONGITUDE : 73. 35. 0.000 0.000 0.000

FREQUENCY : 500.0 MHz

PROTECTED COUNTRY LEVEL : 0.000000 0.000

BEARING	FILE NO	RANGE	RANGE	LATITUDE	LONGITUDE	DIST	BEARING
0.0	501.53	120.072	24. 17. 13.00	73. 50. 7.00	351.196	49.895	
10.0	419.41	112.821	33. 11. 35.20	73. 35. 30.68	360.201	53.289	
20.0	359.99	101.820	33. 3. 34.64	73. 24. 21.04	363.998	56.760	
30.0	237.89	66.600	33. 3. 34.64	73. 17. 27.61	362.118	60.158	
40.0	152.97	72.798	33. 3. 34.64	73. 15. 45.71	354.178	64.316	
50.0	91.56	53.581	33. 3. 34.64	73. 20. 5.49	339.720	66.029	
60.0	29.13	29.365	33. 3. 34.64	73. 31. 43.75	317.298	67.991	
70.0	13.21	15.301	33. 3. 34.64	73. 49. 10.16	289.912	68.584	
80.0	7.51	8.015	33. 3. 34.64	73. 38. 59.50	203.741	69.270	
90.0	3.70	4.008	33. 3. 34.64	73. 43. 50.53	297.564	69.244	
100.0	1.94	2.017	33. 3. 34.64	73. 42. 38.73	297.869	69.745	
110.0	0.70	0.700	33. 3. 34.64	73. 29. 35.05	315.929	72.459	
120.0	0.30	0.300	33. 3. 34.64	73. 24. 46.42	313.976	74.719	
130.0	0.00	0.000	33. 3. 34.64	73. 23. 7.21	307.831	76.477	
140.0	0.00	0.000	33. 3. 34.64	73. 27. 56.88	299.285	77.564	
150.0	0.00	0.000	33. 3. 34.64	73. 32. 53.10	290.443	76.847	
160.0	0.00	0.000	33. 3. 34.64	73. 39. 19.41	284.095	74.569	
170.0	0.00	0.000	33. 3. 34.64	73. 46. 7.00	284.732	70.604	
180.0	0.00	0.000	33. 3. 34.64	73. 50. 12.29	284.131	70.035	
190.0	0.00	0.000	33. 3. 34.64	73. 51. 58.25	278.448	70.913	
200.0	0.00	0.000	33. 3. 34.64	73. 53. 34.24	287.448	68.732	
210.0	0.00	0.000	33. 3. 34.64	73. 50. 40.65	263.335	71.447	
220.0	0.00	0.000	33. 3. 34.64	73. 19. 40.65	238.335	72.333	
230.0	0.00	0.000	33. 3. 34.64	73. 16. 47.86	216.814	70.921	
240.0	0.00	0.000	33. 3. 34.64	73. 15. 47.86	199.874	67.255	
250.0	0.00	0.000	33. 3. 34.64	73. 12. 2.08	189.805	61.651	
260.0	0.00	0.000	33. 3. 34.64	73. 12. 2.08	188.363	55.052	
270.0	0.00	0.000	33. 3. 34.64	73. 17. 1.02	195.812	48.823	
280.0	0.00	0.000	33. 3. 34.64	73. 17. 41.42	210.709	44.056	
290.0	0.00	0.000	33. 3. 34.64	73. 13. 51.95	230.702	51.416	
300.0	0.00	0.000	33. 3. 34.64	73. 5. 51.86	253.416	40.036	
310.0	0.00	0.000	33. 3. 34.64	73. 54. 14.27	276.864	40.343	
320.0	0.00	0.000	33. 3. 34.64	73. 39. 46.57	299.488	41.948	
330.0	0.00	0.000	33. 3. 34.64	73. 23. 20.68	337.578	46.728	

Figure D3. Results for WCKL

FILE: MELA OUTPUT P VZEP CONVENTIONAL MONITOR SYSTEM PAGE 00001

LATITUDE : 10.56
 LONGITUDE : 16.29
 LATITUDE : 41.12
 LONGITUDE : 77.2
 FREQUENCY : 550.0 KHZ

 PROJECTED COUNTRY LEVEL : 0.500000 MVA

BEARING	FIELD IN KM	RADIUS	LATITUDE	LONGITUDE	DIST	BEARING
0.0	202.57	57.089	41.30	23.00	57.220	54.438
10.0	239.14	66.626	41.35	2.27	71.590	54.365
20.0	373.77	74.396	41.37	2.43	85.415	57.607
30.0	412.48	79.775	41.36	35.83	97.576	62.671
40.0	482.86	82.741	41.33	31.62	107.305	68.778
50.0	487.37	83.073	41.32	33.47	114.089	75.473
60.0	455.83	80.777	41.27	38.0	117.678	82.454
70.0	396.14	76.149	41.13	35.97	115.762	89.479
80.0	321.16	69.679	41.5	14.96	96.344	96.344
90.0	246.16	62.168	41.69	56.59	102.840	102.840
100.0	185.04	54.840	41.60	49.62	108.781	108.781
110.0	132.35	46.617	41.50	56.0	114.001	114.001
120.0	135.77	47.781	41.40	53.83	123.602	123.602
130.0	142.24	48.662	41.30	50.0	128.352	128.352
140.0	141.27	48.500	41.20	53.83	132.969	132.969
150.0	136.77	47.781	41.10	50.0	138.052	138.052
160.0	147.38	54.840	41.0	53.83	144.298	144.298
170.0	135.04	54.840	41.0	50.0	151.568	151.568
180.0	246.16	62.168	41.0	56.0	159.466	159.466
190.0	321.16	74.396	41.0	53.83	167.665	167.665
200.0	366.14	78.141	41.0	50.0	175.910	175.910
210.0	455.83	83.073	41.0	53.83	184.067	184.067
220.0	487.37	82.741	41.0	50.0	191.848	191.848
230.0	482.86	79.775	41.0	53.83	194.950	194.950
240.0	442.49	74.396	41.0	50.0	204.765	204.765
250.0	373.77	66.626	41.0	53.83	207.606	207.606
260.0	289.14	57.089	41.0	50.0	203.433	203.433
270.0	126.57	46.617	41.0	53.83	175.251	175.251
280.0	71.28	34.544	41.0	50.0	158.863	158.863
290.0	49.40	28.083	41.0	53.83	130.159	130.159
300.0	51.01	28.083	41.0	50.0	113.533	113.533
310.0	49.40	27.532	41.0	53.83	25.210	25.210
320.0	47.79	27.532	41.0	50.0	28.589	28.589
330.0	71.28	34.544	41.0	53.83	32.782	32.782
340.0	126.57	46.617	41.0	50.0	35.921	35.921
350.0			41.0	53.83	44.155	44.155

Figure D4. Results for WHLM

LAUFTEMPERATUR:	100,000
LAUFGEHÄRTE:	0,000

[illegible]

1990-1991

[illegible][illegible]

Figure D5. Results for WSYR

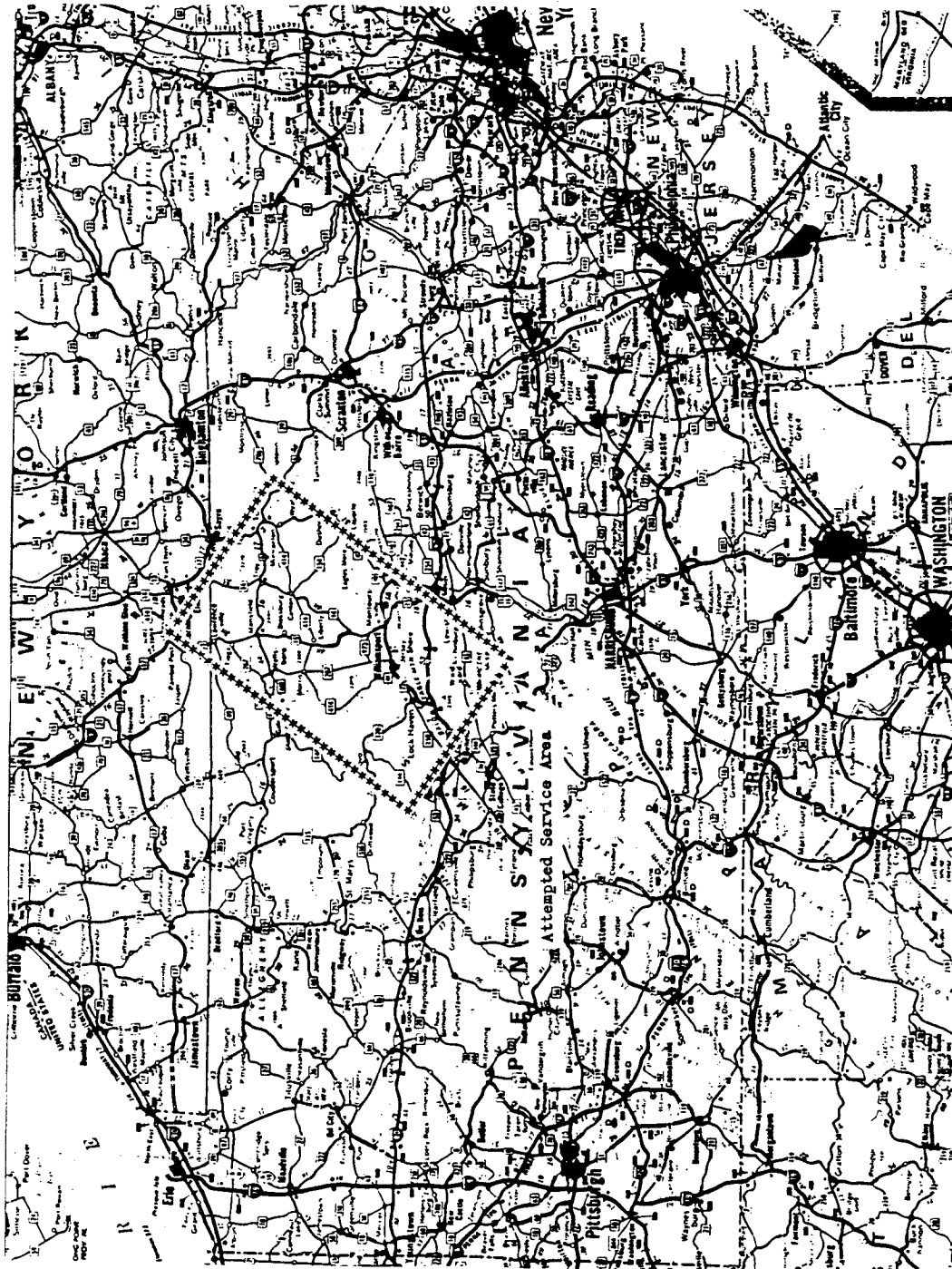


Figure D6. Attempted Service Area

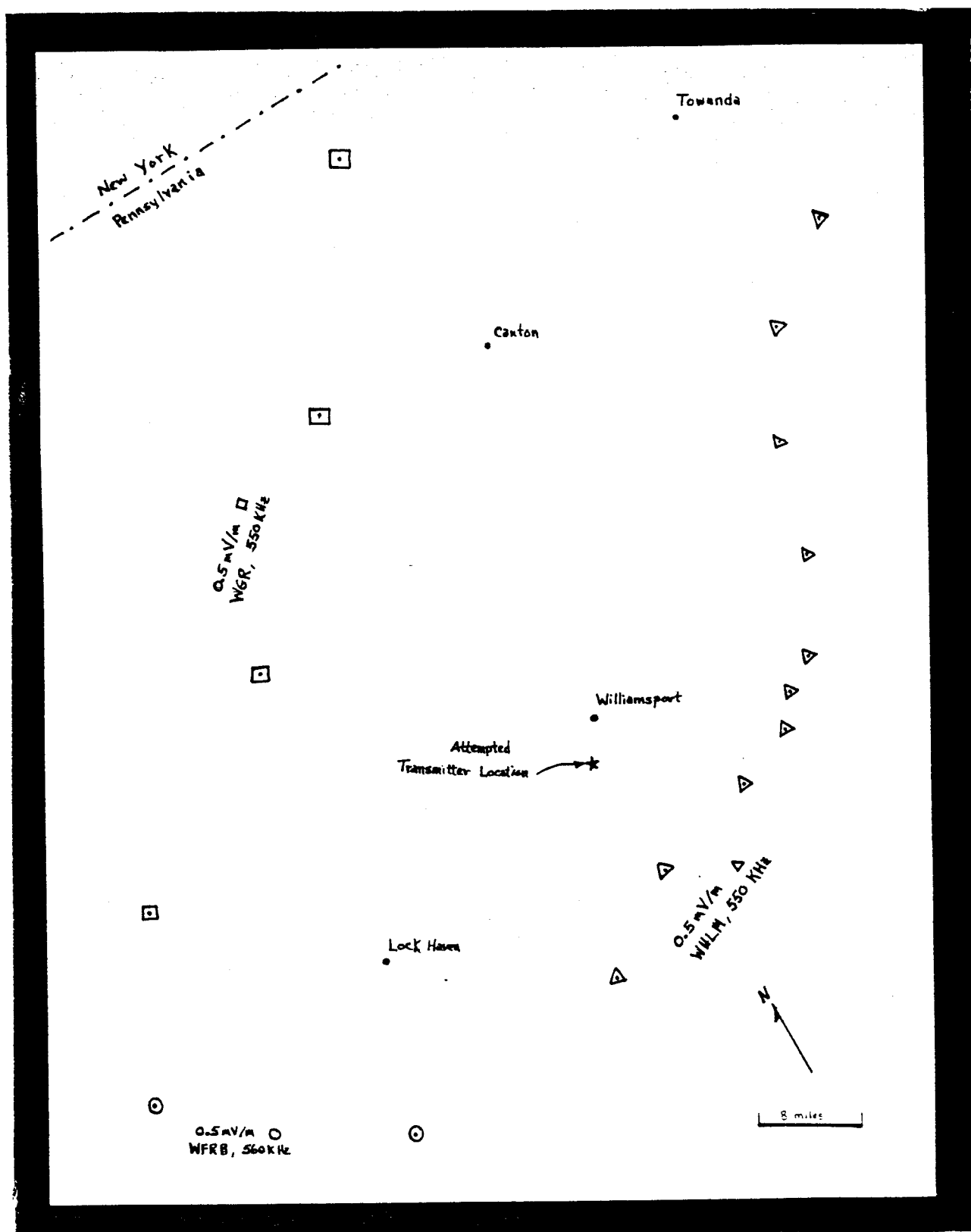
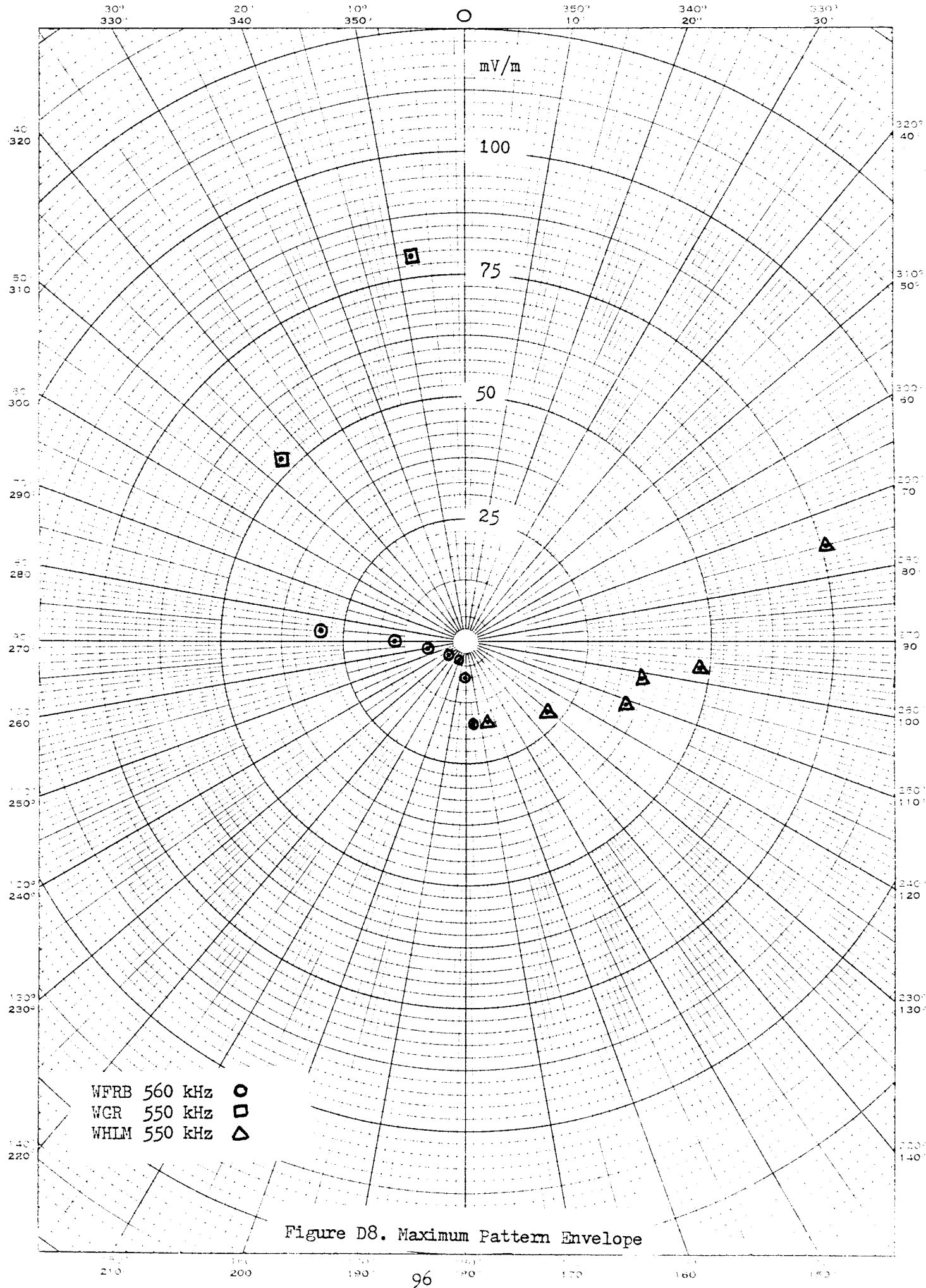


Figure D7. Protected Contour Points



APPROVAL OF EXAMINING COMMITTEE

Wils L. Cooley, Ph.D.

Mark A. Jerabek, Ph.D.

Date

James F. Corum, Ph.D., Chairman